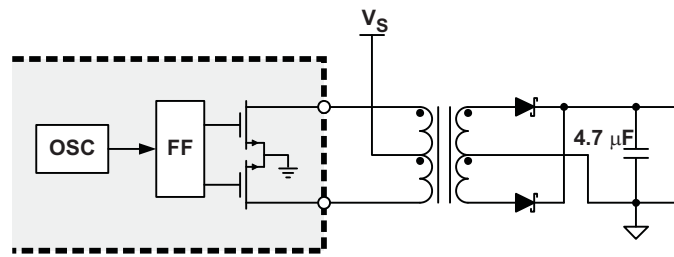




The integrated transformer driver consists of an oscillator followed by a flip-flop stage generating two complementary 50% duty-cycle square waves. These two signals drive the ground-referenced N-channel power switches. Internal circuitry ensures break-before-make action between the two switches.



**Figure 2. Simplified Transformer Driver**

A pair of Schottky diodes and a bulk capacitor build a full-wave rectifier, providing the input voltage for the TPS76350 linear voltage regulator.

This low-dropout (LDO) voltage regulator offers the benefits of low dropout voltage, low-power operation, and miniaturized packaging. The regulator features low dropout voltages and quiescent currents compared to conventional LDO regulators. Offered in a 5-terminal, small-outline, integrated-circuit SOT-23 package, the TPS76350 is ideal for cost-sensitive designs and for applications where board space is at a premium.

A combination of new circuit design and process innovation has enabled the usual PNP pass transistor to be replaced by a PMOS pass element. Because the PMOS pass element behaves as a low-value resistor, the dropout voltage is very low (typically 300 mV at 150 mA of load current) and is directly proportional to the load current. Because the PMOS pass element is a voltage-driven device, the quiescent current is very low (140 µA maximum) and is stable over the entire range of output load current (0 mA to 150 mA).

The TPS76350 also features a logic-enabled sleep mode to shut down the regulator, reducing quiescent current to 1 µA maximum at  $T_j = 25^\circ\text{C}$ . The TPS76350 output provides a regulated 5-V power supply to the ISO3086T bus-side circuitry ( $V_{CC2}$ ) as shown in [Figure 1](#).

Although not required, a 0.047 µF or larger ceramic bypass input capacitor, connected between IN and GND and located close to the TPS763xx, is recommended to improve transient response and noise rejection. A higher-value electrolytic input capacitor may be necessary if large, fast-rise-time load transients are anticipated and the device is located several inches from the power source.

Like all low-dropout regulators, the TPS76350 requires an output capacitor connected between OUT and GND to stabilize the internal loop control. The minimum recommended capacitance value is 4.7 µF and the equivalent series resistance (ESR) must be between 0.3 Ω and 10 Ω. Capacitor values of 4.7 µF or larger are acceptable, provided the ESR is less than 10 Ω. Solid tantalum electrolytic, aluminum electrolytic, and multilayer ceramic capacitors are all suitable, provided they meet the requirements described previously. Most of the commercially available 4.7-µF surface-mount solid tantalum capacitors, including devices from Sprague, Kemet, and Nichico, meet the ESR requirements stated previously.

### 3 Common Variations

When the ISO3086T is powered by 3.3 V on the logic side, a 1CT : 2.2CT transformer (T1) is required to step up the 3.3-V input supply ( $V_S$ ) to 6 V at the transformer secondary, thereby ensuring enough headroom for the TPS76350 LDO to provide a regulated 5-V output. If the ISO3086T is powered by 5 V on the logic side, a 1CT : 1.5CT transformer is needed.

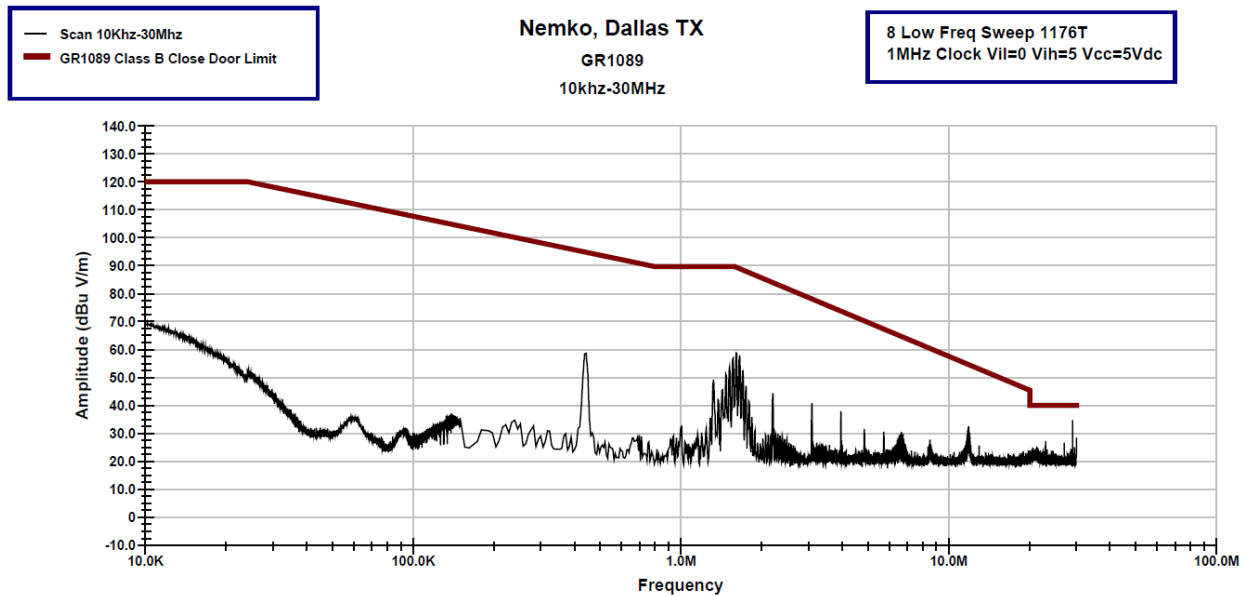
The external transformer used with the ISO3086T must have a center-tapped primary winding. The turns ratio of the transformer must provide the minimum required output voltage at the maximum anticipated load current with the minimum input voltage. [Table 1](#) lists suitable transformers for the circuit in [Figure 1](#).

**Table 1. Recommended Isolation Transformers**

Manufacturer	Primary Voltage (3.3 V)	Primary Voltage (5 V)
Coilcraft	DA2304-AL	DA2303-AL
Murata	782482/35C	782485/55C

#### 4 Circuit Performance

Figure 3 shows a low-frequency sweep for measuring the radiated emissions to demonstrate compliance with the EN55022 standard, which specifies the limits and methods of measurement of radio disturbance characteristics of information technology equipment.



**Figure 3. Radiated Emissions (dB $\mu$ V/m) from 10 kHz to 30 MHz**

Figure 4 shows the circuit efficiency over the load current. While load currents in the lower 20 mA can be expected for low data rates and no common-mode voltage, higher load currents around 100 mA are required for high data rates and high common-mode voltages.

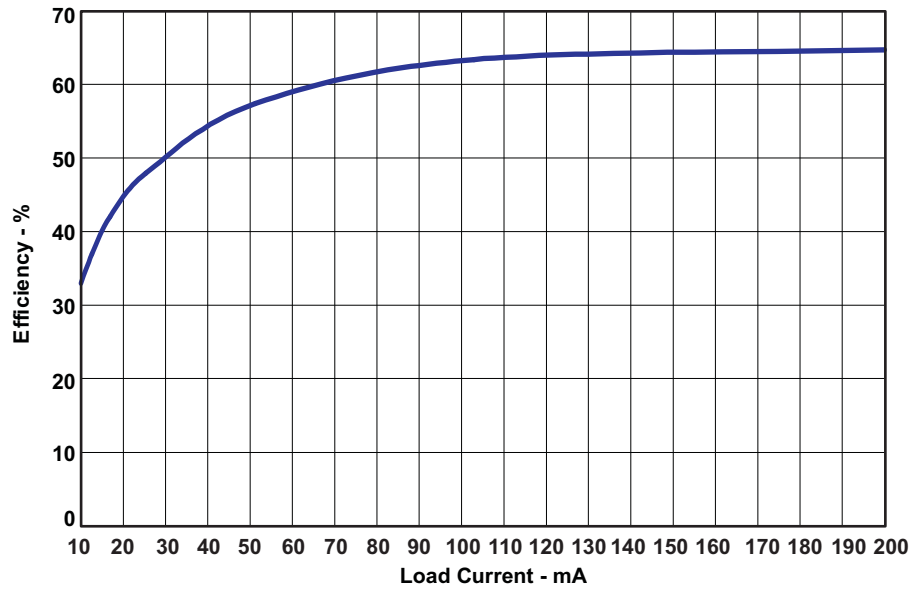


Figure 4. Efficiency Over Load Current

## 5 Circuit Layout and BOM

Figure 5 shows the silk screen as well as the top and bottom layers of the two-layer board, with the actual dc-dc converter capturing an area of 20 mm x 14 mm only.

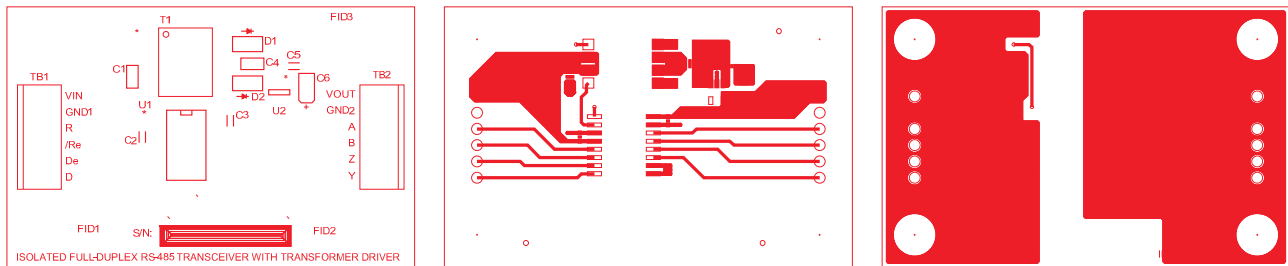


Figure 5. Board Layout: Silk Screen (Left), Top Layer (Middle), Bottom Layer (Right)

Figure 6 shows a picture of the actual board.

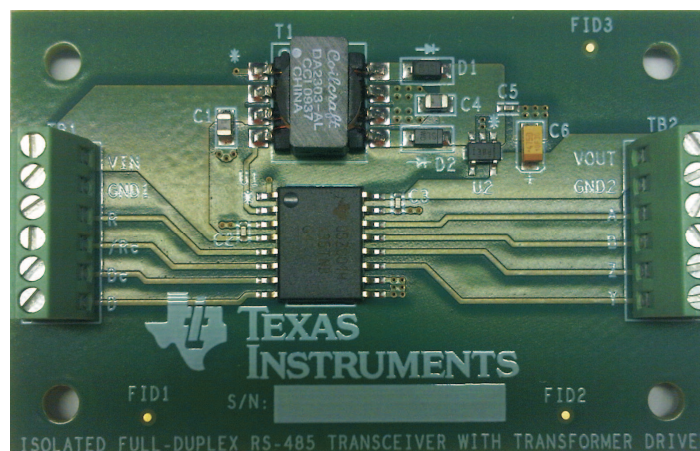


Figure 6. Reference Design Board: Top View

Table 2 shows the final bill of materials.

**Table 2. Bill of Materials**

<b>Designator</b>	<b>Part No.</b>	<b>MFR</b>	<b>Value</b>	<b>Package/Size</b>
C1, C4	GRM21BR61C106KE15L	Murata	10- $\mu$ F/16-V/X5R (ceramic)	0805 (2 x 1,25 mm <sup>2</sup> )
C2, C3, C5	C0402C104K4RACTU	KEMET	0.1- $\mu$ F/16-V/X7R (ceramic)	0402 (1 x 0,5 mm <sup>2</sup> )
C6	T491A475K016AT	KEMET	4.7- $\mu$ F/16-V (tantalum)	3,2 x 1,6 mm <sup>2</sup>
D1, D2	MBR0520L	Fairchild	0.4-W Schottky diode	SOD123
J1, J2	MPT 0,5 / 6-2,54	Phoenix-Contact	6-position terminal block	6,2 x 15, mm <sup>2</sup>
T1	DA2304-AL	Coilcraft	1:2.2	10 x 12 mm <sup>2</sup>
U1	ISO3086T	TI	Half-duplex XCVR	DW-16
U2	TPS76350	TI	5-V LDO	DBV-5

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