

Sample &

Buy





SBVS054J-NOVEMBER 2004-REVISED APRIL 2015

Support &

Community

29

TPS730 Low-Noise, High PSRR, RF, 200-mA Low-Dropout Linear Regulators

Technical

Documents

1 Features

- 200-mA RF Low-Dropout Regulator With Enable
- Available in Fixed Voltages from 1.8 V to 3.3 V and Adjustable Voltages (1.22 V to 5.5 V)
- High PSRR (68 dB at 100 Hz)
- Low Noise (33 μV_{RMS}, TPS73018)
- Fast Start-Up Time (50 µs)
- Stable With a 2.2-µF Ceramic Capacitor
- Excellent Load/Line Transient Response
- Very Low Dropout Voltage (120 mV at 200 mA)
- 5- and 6-Pin SOT-23 (DBV), and Wafer Chip Scale (YZQ) Packages

2 Applications

- RF: VCOs, Receivers, ADCs
- Audio
- Cellular and Cordless Telephones
- Bluetooth[®], Wireless LAN
- Handheld Organizers, PDAs

3 Description

Tools &

Software

The TPS730 family of low-dropout (LDO) low-power linear voltage regulators features high power-supply rejection ratio (PSRR), low noise, fast start-up, and excellent line and load transient responses in a small SOT-23 package. NanoStar[™] packaging gives an ultrasmall footprint as well as an ultralow profile and package weight, making it ideal for portable applications such as handsets and PDAs. Each device in the family is stable, with a small, 2.2-µF ceramic capacitor on the output. The TPS730 family uses an advanced, proprietary BiCMOS fabrication process to yield low dropout voltages (for example, 120 mV at 200 mA, TPS73030). Each device achieves fast start-up times (approximately 50 µs with a 0.001-µF bypass capacitor) while consuming low quiescent current (170 µA typical). Moreover, when the device is placed in standby mode, the supply current is reduced to less than 1 µA. The TPS73018 exhibits approximately 33 μV_{RMS} of output voltage noise at 1.8 V output with a 0.01-µF bypass capacitor. Applications with analog components that are noise-sensitive, such as portable RF electronics, benefit from the high PSRR and low-noise features as well as the fast response time.

Device Information⁽¹⁾

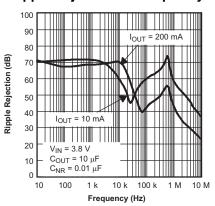
PART NUMBER	PACKAGE	BODY SIZE (NOM)
	SOT-23 (5)	2.90 mm × 1.60 mm
TPS730	SOT-23 (6)	2.90 mm × 1.60 mm
	DSBGA (5)	1.35 mm × 1.00 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

V_{IN} OUT V_{IN} V_{IN} V_{IN} V_{IN} V_{OUT} V

Simplified Schematic

Ripple Rejection vs Frequency



Features 1

Applications 1 Description 1

Revision History..... 2 Pin Configuration and Functions 3

Specifications...... 4 6.1 Absolute Maximum Ratings 4 ESD Ratings 4

Recommended Operating Conditions 4

Thermal Information 4

6.5 Electrical Characteristics 5 6.6 Typical Characteristics 6

7.3 Feature Description...... 10 7.4 Device Functional Modes...... 10 Application and Implementation 11

1

2

3 4

5 6

7

8

6.2

6.3

6.4

2

Table of Contents

	8.1	Application Information	11
	8.2	Typical Application	13
	8.3	Do's and Don'ts	14
9	Pow	er Supply Recommendations	15
10	Lay	out	15
	10.1		
	10.2	Layout Example	15
	10.3	Thermal Considerations	16
	10.4	Power Dissipation	16
11	Dev	ice and Documentation Support	18
	11.1	Device Support	18
	11.2	Documentation Support	18
	11.3	Trademarks	18
	11.4	Electrostatic Discharge Caution	18
	11.5	Glossary	18
12	Mec	hanical, Packaging, and Orderable	
		rmation	18
	12.1		
		Information	19

4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision I (February, 2011) to Revision J

•	Added ESD Ratings table, Feature Description section, Device Functional Modes, Application and Implementation section, Power Supply Recommendations section, Layout section, Device and Documentation Support section, and	
	Mechanical, Packaging, and Orderable Information section	1
•	Changed fourth bullet of Features list to low noise	1
•	Changed front-page figure	1
•	Added Pin Configuration and Functions section	3
•	Changed "free-air temperature" to "junction temperature" in Absolute Maximum Ratings condition statement	4
•	Deleted Dissipation Ratings table; added Thermal Information table	4
•	Added condition statement to Typical Characteristics	6
•	Moved Ordering Information to Device Nomenclature section	18

Changes from Revision H (October, 2007) to Revision I

•	Corrected units in y-axis of Figure 5	6

XAS STRUMENTS

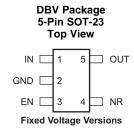
www.ti.com

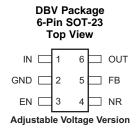
Page

Page

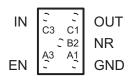


5 Pin Configuration and Functions





YZQ Package 5-Pin DSBGA Top View



Pin Functions

	PIN				
NAME	N	ю.	I/O	DESCRIPTION	
NAWE	SOT-23	DSBGA			
EN	3	A3	I	Enable pin. Driving the enable pin (EN) high turns on the regulator. Driving this pin low puts the regulator into shutdown mode. EN can be connected to IN if not used.	
FB	5	N/A	I	Feedback pin. This terminal is the feedback input pin for the adjustable device. Fixed-voltage versions in the DBV package do not have this pin.	
GND	2	A1	_	Regulator ground.	
IN	1	C3	I	Input to the device.	
OUT	6	C1	0	Output of the regulator.	
NR	4	B2	_	Noise Reduction pin. Connecting an external capacitor to this pin filters noise generated by the internal bandgap. This configuration improves power-supply rejection and reduces output noise.	



6 Specifications

6.1 Absolute Maximum Ratings

over operating junction temperature range (unless otherwise noted)⁽¹⁾

			MIN	MAX	UNIT
	Input range, V _{IN}		-0.3	6	
Voltage	Enable range, V _E	Enable range, V _{EN}		6	V
	Output range, V _C	UT	-0.3	6	
Current	Peak output, I _{OUT(max)}		Interna	Internally limited	
Continuous total power dissi	Continuous total power dissipation		See Therm	al Information	
	Junction, T _J	DBV package	-40	150	
Temperature		YZQ package	-40	125	°C
	Storage, T _{stg}		-65	150	

(1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

6.2 ESD Ratings

			VALUE	UNIT
		Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins ⁽¹⁾	±2000	
V _(ESD)	Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101, all $\ensuremath{\text{pins}}^{(2)}$	±500	V

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

over operating junction temperature range (unless otherwise noted).

		MIN	NOM	MAX	UNIT
V _{IN}	Input supply voltage	2.7		5.5	V
V _{EN}	Enable supply voltage	0		V _{IN}	V
V _{OUT}	Output voltage	V _{FB}		5	V
I _{OUT}	Output current	0		200	mA
TJ	Operating junction temperature	-40		125	°C
C _{IN}	Input capacitor	0.1	1		μF
C _{OUT}	Output capacitor	2.2 ⁽¹⁾	10		μF
C _{NR}	Noise reduction capacitor	0	10		nF
C _{FF}	Feed-forward capacitor		15		pF
R ₂	Lower feedback resistor		30.1		kΩ

(1) If C_{FF} is not used or V_{OUT(nom)} < 1.8 V, the minimum recommended C_{OUT} = 4.7 μ F.

6.4 Thermal Information

		TPS7		
	THERMAL METRIC ⁽¹⁾	DBV (SOT-23)	YZQ (DSBGA)	UNIT
		6 PINS	5 PINS	
R_{\thetaJA}	Junction-to-ambient thermal resistance	225.1	178.5	
R _{0JC(top)}	Junction-to-case (top) thermal resistance	78.4	1.4	
$R_{\theta JB}$	Junction-to-board thermal resistance	54.7	62.1	°C/W
ΨJT	Junction-to-top characterization parameter	3.3	0.9	
ψ_{JB}	Junction-to-board characterization parameter	53.8	62.1	

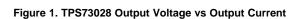
(1) For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, SPRA953.

6.5 Electrical Characteristics

Over recommended operating temperature range $T_J = -40$ to $+125^{\circ}$ C, $V_{EN} = V_{IN}$, $V_{IN} = V_{OUT(nom)} + 1$ V⁽¹⁾, $I_{OUT} = 1$ mA, $C_{OUT} = 10 \mu$ F, $C_{NR} = 0.01 \mu$ F (unless otherwise noted). Typical values are at 25°C.

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{IN}	Input voltage range	1)		2.7		5.5	V
I _{OUT}	Continuous output c	urrent		0		200	mA
V _{FB}	Internal reference (T	PS73001)		1.201	1.225	1.25	V
V _{OUT}	Output voltage range	TPS73001		V _{FB}	5	.5 – V _{DO}	V
	Output voltage accu	racy	0 μA ≤ I _{OUT} ≤ 200 mA, 2.75 V ≤ V _{IN} ≤ 5.5 V	-2%	V _{OUT(nom)}	2%	V
$\Delta V_{OUT(\Delta VIN)}$	Line regulation ⁽¹⁾		V_{OUT} + 1 V ≤ V_{IN} ≤ 5.5 V		0.05		%/V
$\Delta V_{OUT(\Delta IOUT)}$	Load regulation		0 µA ≤ I_{OUT} ≤ 200 mA, T_J = 25°C		5		mV
V _{DO}	Dropout voltage ⁽²⁾ $(V_{IN} = V_{OUT(nom)} - 0$.1 V)	I _{OUT} = 200 mA		120	210	mV
I _{CL}	Output current limit		V _{OUT} = 0 V	285		600	mA
I _{GND}	Ground pin current		0 μA < I _{OUT} < 200 mA		170	250	μA
I _{SHUTDOWN}	Shutdown current ⁽³⁾		$V_{EN} = 0 \text{ V}, 2.7 \text{ V} \le V_{IN} \le 5.5 \text{ V}$		0.07	1	μA
I _{FB}	FB pin current		V _{FB} = 1.8 V			1	μA
PSRR	Power-supply rejection ratio	TPS73028	f = 100 Hz, I _{OUT} = 200 mA, T _J = 25°C		68		dB
V _n	Output noise voltage	TPS73018	BW = 200 Hz to 100 kHz, I_{OUT} = 200 mA, C_{NR} = 0.01 μ F		33		μV_{RMS}
t _{STR}	Start-up time	TPS73018	R_L = 14 Ω, C_{OUT} = 1 μF, C_{NR} = 0.001 μF		50		μs
V _{EN(high)}	High-level enable in	put voltage	2.7 V ≤ V _{IN} ≤ 5.5 V	1.7		V _{IN}	V
V _{EN(low)}	Low-level enable inp	out voltage	$2.7 \text{ V} \leq \text{V}_{\text{IN}} \leq 5.5 \text{ V}$	0		0.7	V
I _{EN}	EN pin current		V _{EN} = 0 V	-1		1	μA
	UVLO		Threshold, V _{CC} rising	2.25		2.65	V
	UVLU		Hysteresis		100		mV

2.795 0 50 100 IOUT (mA)



V_{IN} = 3.8 V

T_J = 25°C

C_{OUT} = 10 μF

150

200

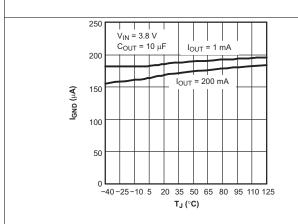
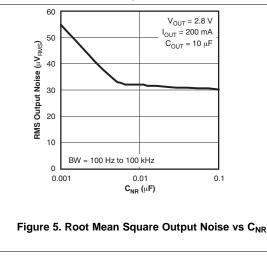
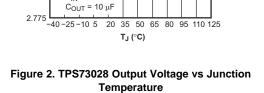


Figure 3. TPS73028 Ground Current vs Junction Temperature





I_{OUT} = 1 mA

I_{OUT} = 200 mA

V_{IN} = 3.8 V

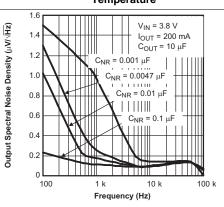
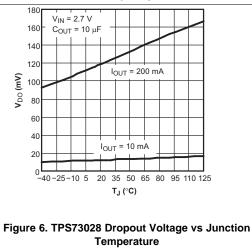


Figure 4. TPS73028 Output Spectral Noise Density vs Frequency



6.6 Typical Characteristics

2.805

2.804

2.803

2.802

2.801 Vout (V)

2.800

2.799

2.798 2.797

2.796

Over recommended operating temperature range $T_J = -40^{\circ}$ C to $+125^{\circ}$ C, $V_{EN} = V_{IN}$, $V_{IN} = V_{OUT(nom)} + 1$ V, $I_{OUT} = 1$ mA, $C_{OUT} = 1$ mA, C_{OUT} 10 μ F, C_{NR} = 0.01 μ F, V_{OUT(nom}) = 2.8 V (unless otherwise noted). Typical values are at T_J = 25°C.

2.805

2.800

2 7 9 5

2.790

2.785

2.780

Vout (V)

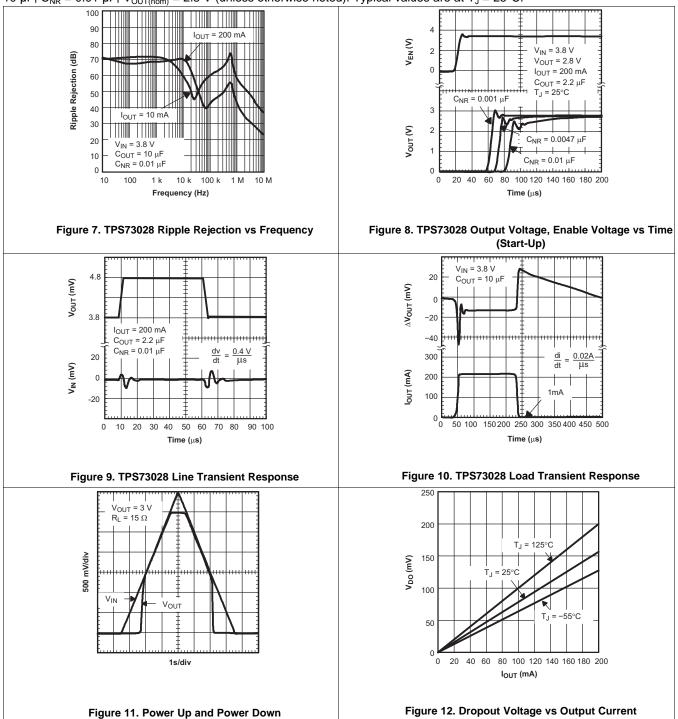
NSTRUMENTS

EXAS



Typical Characteristics (continued)

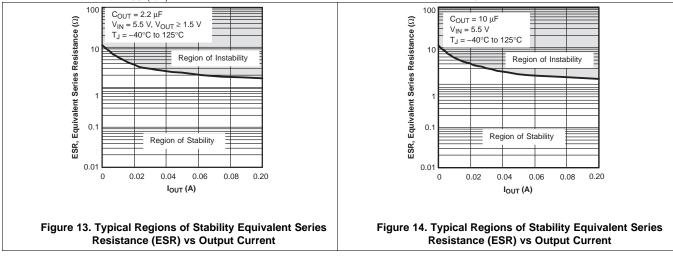
Over recommended operating temperature range $T_J = -40^{\circ}$ C to $+125^{\circ}$ C, $V_{EN} = V_{IN}$, $V_{IN} = V_{OUT(nom)} + 1$ V, $I_{OUT} = 1$ mA, $C_{OUT} = 10 \mu$ F, $C_{NR} = 0.01 \mu$ F, $V_{OUT(nom)} = 2.8$ V (unless otherwise noted). Typical values are at $T_J = 25^{\circ}$ C.





Typical Characteristics (continued)

Over recommended operating temperature range $T_J = -40^{\circ}$ C to $+125^{\circ}$ C, $V_{EN} = V_{IN}$, $V_{IN} = V_{OUT(nom)} + 1$ V, $I_{OUT} = 1$ mA, $C_{OUT} = 10 \mu$ F, $C_{NR} = 0.01 \mu$ F, $V_{OUT(nom)} = 2.8$ V (unless otherwise noted). Typical values are at $T_J = 25^{\circ}$ C.





7 Detailed Description

7.1 Overview

The TPS730 family of low-dropout (LDO) regulators has been optimized for use in noise-sensitive, batteryoperated equipment. The device features extremely low dropout voltages, high PSRR, ultra-low output noise, low quiescent current (170 μ A typically), and enable-input to reduce supply currents to less than 1 μ A when the regulator is turned off.

7.2 Functional Block Diagrams

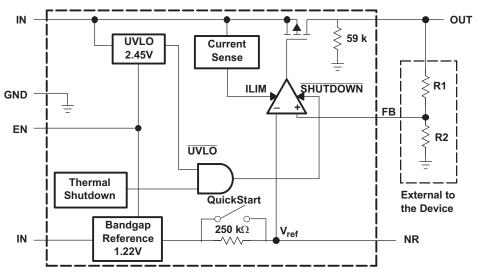


Figure 15. TPS730 Block Diagram (Adjustable-Voltage Version)

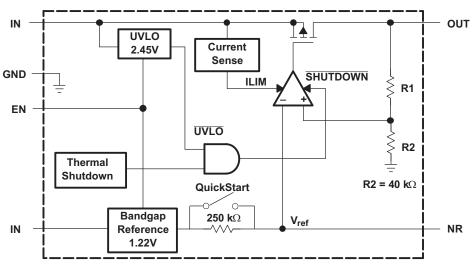


Figure 16. TPS730 Block Diagram (Fixed-Voltage Versions)



7.3 Feature Description

7.3.1 Undervoltage Lockout (UVLO)

The TPS730 uses an undervoltage lockout (UVLO) circuit that disables the output until the input voltage is greater than the rising UVLO voltage. This circuit ensures that the device does not exhibit any unpredictable behavior when the supply voltage is lower than the operational range of the internal circuitry, $V_{IN(min)}$.

7.3.2 Shutdown

The enable pin (EN) is active high. Enable the device by forcing the EN pin to exceed $V_{EN(high)}$ (1.7 V, minimum). Turn off the device by forcing the EN pin to drop below 0.7 V. If shutdown capability is not required, connect EN to IN.

7.3.3 Foldback Current Limit

The TPS730 features internal current limiting and thermal protection. During normal operation, the TPS730 limits output current to approximately 400 mA. When current limiting engages, the output voltage scales back linearly until the overcurrent condition ends. While current limiting is designed to prevent gross device failure, do not exceed the power dissipation ratings of the package or the absolute maximum voltage ratings of the device.

7.4 Device Functional Modes

7.4.1 Normal Operation

The device regulates to the nominal output voltage under the following conditions:

- The input voltage is at least as high as V_{IN(min)}.
- The input voltage is greater than the nominal output voltage added to the dropout voltage.
- The enable voltage is greater than V_{EN(min)}.
- The output current is less than the current limit.
- The device junction temperature is less than the maximum specified junction temperature.

7.4.2 Dropout Operation

If the input voltage is lower than the nominal output voltage plus the specified dropout voltage, but all other conditions are met for normal operation, the device operates in dropout mode. In this mode of operation, the output voltage is the same as the input voltage minus the dropout voltage. The transient performance of the device is significantly degraded because the pass device is in the linear region and no longer controls the current through the LDO. Line or load transients in dropout can result in large output voltage deviations.

7.4.3 Disabled

The device is disabled under the following conditions:

- The enable voltage is less than the enable falling threshold voltage or has not yet exceeded the enable rising threshold.
- The device junction temperature is greater than the thermal shutdown temperature.
- The input voltage is less than UVLO_{falling}.

Table 1 shows the conditions that lead to the different modes of operation.

OPERATING MODE	PARAMETER						
OPERATING MODE	V _{IN}	V _{EN}	Ι _{ουτ}	TJ			
Normal mode	$\begin{array}{l} V_{\text{IN}} > V_{\text{OUT(nom)}} + V_{\text{DO}} \text{ and} \\ V_{\text{IN}} > V_{\text{IN(min)}} \end{array}$	$V_{EN} > V_{EN(high)}$	I _{OUT} < I _{LIM}	T _J < 125°C			
Dropout mode	$V_{IN(min)} < V_{IN} < V_{OUT(nom)} + V_{DO}$	$V_{EN} > V_{EN(high)}$	—	T _J < 125°C			
Disabled mode (any true condition disables the device)	V _{IN} < UVLO _{falling}	V _{EN} < V _{EN(low)}	_	$T_{\rm J} > 165^{\circ}C^{(1)}$			

 Table 1. Device Functional Mode Comparison

(1) Approximate value for thermal shutdown.



8 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

The TPS730 family of low-dropout (LDO) regulators has been optimized for use in noise-sensitive batteryoperated equipment. The device features extremely low dropout voltages, high PSRR, ultra-low output noise, low quiescent current (170 μ A typically), and enable-input to reduce supply currents to less than 1 μ A when the regulator is turned off.

8.1.1 Adjustable Operation

The output voltage of the TPS73001 adjustable regulator is programmed using an external resistor divider as shown in Figure 17. The output voltage is calculated using Equation 1:

$$V_{OUT} = V_{REF} \times \left(1 + \frac{R_1}{R_2}\right)$$

Where:

V_{REF} = 1.225 V typical (the internal reference voltage)

(1)

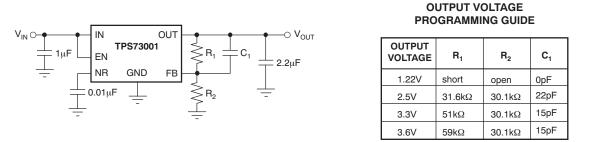
Resistors R₁ and R₂ should be chosen for approximately 50- μ A divider current. Lower value resistors can be used for improved noise performance, but the solution consumes more power. Higher resistors values can cause accuracy issues and other problems. The recommended design procedure is to choose R₂ = 30.1 k Ω to set the divider current at 50 μ A, C₁ = 15 pF for stability, and then calculate R₁ using Equation 2:

$$R_{1} = \left(\frac{V_{OUT}}{V_{REF}} - 1\right) \times R_{2}$$
(2)

To improve the stability of the adjustable version, TI suggests placing a small compensation capacitor between OUT and FB. For output voltages < 1.8 V, the value of this capacitor should be 100 pF. For output voltages > 1.8 V, use Equation 3 to calculate the approximate value of this capacitor.

$$C_{1} = \frac{(3 \times 10^{-7}) \times (R_{1} + R_{2})}{(R_{1} \times R_{2})}$$
(3)

Figure 17 shows the suggested value of this capacitor for several resistor ratios. If this capacitor is not used (such as in a unity-gain configuration) or if an output voltage < 1.8 V is chosen, then the minimum recommended output capacitor is 4.7 μ F instead of 2.2 μ F.







Application Information (continued)

8.1.2 Capacitor Recommendations

Low equivalent series resistance (ESR) capacitors should be used for the input, output, noise reduction, and bypass capacitors. Ceramic capacitors with X7R and X5R dielectrics are preferred. These dielectrics offer more stable characteristics. Ceramic X7R capacitors offer improved overtemperature performance, while ceramic X5R capacitors are more cost-effective and are available in higher values.

8.1.3 Input and Output Capacitor Requirements

A 0.1- μ F or larger ceramic input bypass capacitor, connected between IN and GND and located close to the TPS730, is required for stability and improves transient response, noise rejection, and ripple rejection. A higher-value input capacitor may be necessary if large, fast-rise-time load transients are anticipated or the device is located several inches from the power source.

Like most low-dropout regulators, the TPS730 requires an output capacitor connected between OUT and GND to stabilize the internal control loop. The minimum recommended capacitance is 2.2 μ F. Any 2.2- μ F or larger ceramic capacitor is suitable, provided the capacitance does not vary significantly over temperature. If load current is not expected to exceed 100 mA, a 1- μ F ceramic capacitor can be used. If a feed-forward capacitor is not used (such as in a unity-gain configuration) or if an output voltage less than 1.8 V is chosen, then the minimum recommended output capacitor is 4.7 μ F instead of 2.2 μ F. Table 2 lists the recommended output capacitors.

CONDITION	C _{OUT} (μF)
V_{OUT} < 1.8 V or C_{FF} = 0 nF	4.7
V _{OUT} > 1.8 V, I _{OUT} > 100 mA	2.2
V _{OUT} > 1.8 V, I _{OUT} < 100 mA	1

Table 2. Output Capacitor Sizing

8.1.4 Noise Reduction and Feed-Forward Capacitor Requirements

The internal voltage reference is a key source of noise in an LDO regulator. The TPS730 has an NR pin which is connected to the voltage reference through a 250-k Ω internal resistor. The 250-k Ω internal resistor, in conjunction with an external bypass capacitor connected to the NR pin, creates a low-pass filter to reduce the voltage reference noise and, therefore, the noise at the regulator output. In order for the regulator to operate properly, the current flow out of the NR pin must be at a minimum, because any leakage current creates an IR drop across the internal resistor thus creating an output error. Therefore, the bypass capacitor must have minimal leakage current. The bypass capacitor should be no more than 0.1 μ F to ensure that it is fully charged during the quick-start time provided by the internal switch shown in the *Functional Block Diagram* section.

As an example, the TPS73018 exhibits only 33 μV_{RMS} of output voltage noise using a 0.01- μ F ceramic bypass capacitor and a 2.2- μ F ceramic output capacitor. Note that the output starts up slower as the bypass capacitance increases due to the RC time constant at the NR pin that is created by the internal 250-k Ω resistor and external capacitor.

A feed-forward capacitor is recommended to improve the stability of the device. If $R_2 = 30.1 \text{ k}\Omega$, set C_1 to 15 pF for optimal performance. For voltages less than 1.8 V, the value of this capacitor should be 100 pF. For voltages greater than 1.8 V, the approximate value of this capacitor can be calculated as shown in Equation 3.

8.1.5 Reverse Current Operation

The TPS730 PMOS-pass transistor has a built-in back diode that conducts reverse current when the input voltage drops below the output voltage (for example, during power-down). Current is conducted from the output to the input and is not internally limited. If extended reverse voltage operation is anticipated, external limiting might be appropriate. If extended reverse voltage operation in anticipated, external limiting to 5% of the rated output current is recommended.



8.2 Typical Application

A typical application circuit is shown in Figure 18.

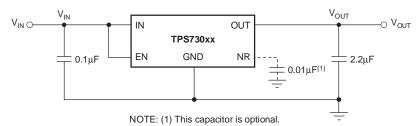


Figure 18. Typical Application Circuit

8.2.1 Design Requirements

Table 3 lists the design requirements.

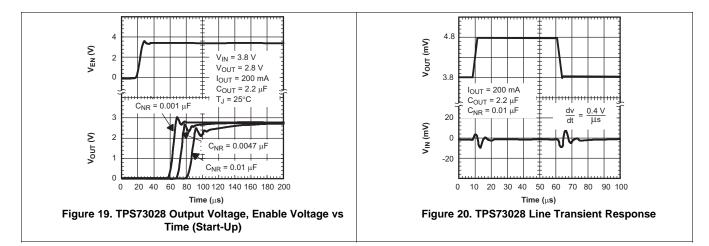
PARAMETER	DESIGN REQUIREMENT
Input voltage	4.2 V to 3 V (Lithium Ion battery)
Output voltage	1.8 V, ±1%
DC output current	10 mA
Peak output current	75 mA
Maximum ambient temperature	65°C

Table 3. Design Parameters

8.2.2 Detailed Design Procedure

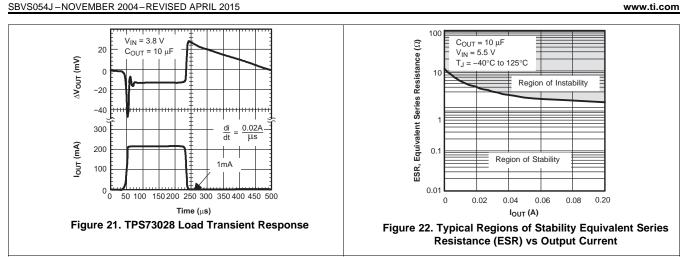
Pick the desired output voltage option. An input capacitor of 0.1 μ F is used as the battery is connected to the input through a via and a short 10-mil (0.01-in) trace. An output capacitor of 10 μ F is used to provide optimal response time for the load transient. Verify that the maximum junction temperature is not exceed by referring to Figure 24.

8.2.3 Application Curves





TPS730 SBVS054J – NOVEMBER 2004 – REVISED APRIL 2015



8.3 Do's and Don'ts

Do place at least one, low-ESR, 2.2- μ F capacitor as close as possible between the OUT pin of the regulator and the GND pin.

Do place at least one, low-ESR, 0.1- μ F capacitor as close as possible between the IN pin of the regulator and the GND pin.

Do provide adequate thermal paths away from the device.

Do not place the input or output capacitor more than 10 mm away from the regulator.

Do not exceed the absolute maximum ratings.

Do not float the Enable (EN) pin.

Do not resistively or inductively load the NR pin.

Do not let the output voltage get more than 0.3 V above the input voltage.



9 Power Supply Recommendations

These devices are designed to operate from an input voltage supply range from 2.7 V to 5.5 V. The input voltage range must provide adequate headroom in order for the device to have a regulated output. This input supply must be well-regulated and stable. A 0.1- μ F input capacitor is required for stability; if the input supply is noisy, additional input capacitors with low ESR can help improve the output noise performance.

10 Layout

10.1 Layout Guidelines

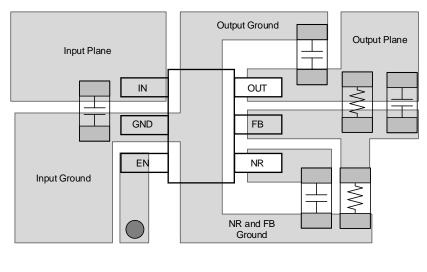
Layout is a critical part of good power-supply design. There are several signal paths that conduct fast-changing currents or voltages that can interact with stray inductance or parasitic capacitance to generate noise or degrade the power-supply performance. To help eliminate these problems, the IN pin should be bypassed to ground with a low ESR ceramic bypass capacitor with an X5R or X7R dielectric.

Equivalent series inductance (ESL) and equivalent series resistance (ESR) must be minimized to maximize performance and ensure stability. Every capacitor (C_{IN} , C_{OUT} , $C_{NR/SS}$, C_{FF}) must be placed as close as possible to the device and on the same side of the PCB as the regulator itself.

Do not place any of the capacitors on the opposite side of the PCB from where the regulator is installed. The use of vias and long traces is strongly discouraged because these circuits may impact system performance negatively, and even cause instability.

10.1.1 Board Layout Recommendations to Improve PSRR and Noise Performance

To improve AC measurements like PSRR, output noise, and transient response, TI recommends designing the board with separate ground planes for V_{IN} and V_{OUT} , with each ground plane connected only at the GND pin of the device. In addition, the ground connection for the bypass capacitor should connect directly to the GND pin of the device.



10.2 Layout Example

Denotes via

Figure 23. Layout Example (DBV Package)



10.3 Thermal Considerations

Thermal protection disables the output when the junction temperature rises to approximately 165°C, allowing the device to cool. When the junction temperature cools to approximately 140°C, the output circuitry is again enabled. Depending on power dissipation, thermal resistance, and ambient temperature, the thermal protection circuit may cycle on and off. This cycling limits regulator dissipation, protecting the device from damage as a result of overheating.

Any tendency to activate the thermal protection circuit indicates excessive power dissipation or an inadequate heatsink. For reliable operation, junction temperature must be limited to 125°C maximum. To estimate the margin of safety in a complete design (including heatsink), increase the ambient temperature until the thermal protection is triggered; use worst-case loads and signal conditions.

The TPS730 internal protection circuitry is designed to protect against overload conditions. This circuitry is not intended to replace proper heatsinking. Continuously running the TPS730 into thermal shutdown degrades device reliability.

10.4 Power Dissipation

Specified regulator operation is assured to a junction temperature of +125°C; the maximum junction temperature should be restricted to +125°C under normal operating conditions. This restriction limits the power dissipation the regulator can handle in any given application. To ensure the junction temperature is within acceptable limits, calculate the maximum allowable dissipation, $P_{D(max)}$, and the actual dissipation, P_D , which must be less than or equal to $P_{D(max)}$.

The maximum power dissipation limit is determined using Equation 4:

$$P_{D(max)} = \frac{T_J m a x - T_A}{R_{\Theta, IA}}$$

Where:

- T_Jmax is the maximum allowable junction temperature.
- R_{eJA} is the thermal resistance junction-to-ambient for the package (see the *Thermal Information* table).
- T_A is the ambient temperature.

The regulator dissipation is calculated using Equation 5:

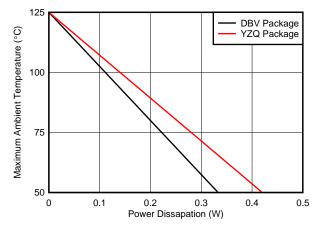
$$P_{D} = (V_{IN} - V_{OUT}) \times I_{OUT}$$

(5)

(4)

Power dissipation resulting from quiescent current is negligible. Excessive power dissipation triggers the thermal protection circuit.

Figure 24 shows the maximum ambient temperature versus the power dissipation of the TPS730. This figure assumes the device is soldered on a JEDEC standard, high-K layout with no airflow over the board. Actual board thermal impedances vary widely. If the application requires high power dissipation, having a thorough understanding of the board temperature and thermal impedances is helpful to ensure the TPS730 does not operate above a junction temperature of 125°C.







TPS730 SBVS054J-NOVEMBER 2004-REVISED APRIL 2015

www.ti.com

Power Dissipation (continued)

Estimating the junction temperature can be done by using the thermal metrics Ψ_{JT} and Ψ_{JB} , shown in the *Thermal Information* table. These metrics are a more accurate representation of the heat transfer characteristics of the die and the package than $R_{\theta JA}$. The junction temperature can be estimated with Equation 6.

$$\Psi_{JT}$$
: $T_J = T_T + \Psi_{JT} \bullet P_D$

$$\Psi_{JB}: T_J = T_B + \Psi_{JB} \bullet P_D$$

where

- P_D is the power dissipation shown by Equation 5,
- T_T is the temperature at the center-top of the IC package,
- T_B is the PCB temperature measured 1 mm away from the IC package on the PCB surface.

(6)

NOTE

Both $T_{\rm T}$ and $T_{\rm B}$ can be measured on actual application boards using a thermo-gun (an infrared thermometer).

For more information about measuring T_T and T_B , see the application note Using New Thermal Metrics (SBVA025), available for download at www.ti.com.

11 Device and Documentation Support

11.1 Device Support

11.1.1 Development Support

11.1.1.1 Spice Models

Computer simulation of circuit performance using SPICE is often useful when analyzing the performance of analog circuits and systems. A SPICE model for the TPS730 is available through the product folders under *Tools & Software*.

11.1.2 Device Nomenclature

Table 4. Ordering Information⁽¹⁾⁽²⁾

PRODUCT	V _{OUT}
	 XX(X) is the nominal output voltage (for example, 28 = 2.8 V; 285 = 2.85 V; 01 = adjustable version). YYY is the package designator. Z is the package quantity. R is for reel (3000 pieces), T is for tape (250 pieces).

(1) For the most current package and ordering information see the Package Option Addendum at the end of this document, or visit the device product folder on www.ti.com.

(2) Output voltages from 1.2 V to 4.8 V in 50-mV increments are available. Contact the factory for details and availability.

11.2 Documentation Support

11.2.1 Related Documentation

- Using New Thermal Metrics, SBVA025
- Pros and Cons of Using a Feedforward Capacitor with a Low-Dropout Regulator, SBVA042

11.3 Trademarks

NanoStar is a trademark of Texas Instruments. Bluetooth is a registered trademark of Bluetooth Sig, Inc. All other trademarks are the property of their respective owners.

11.4 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

11.5 Glossary

SLYZ022 — TI Glossary.

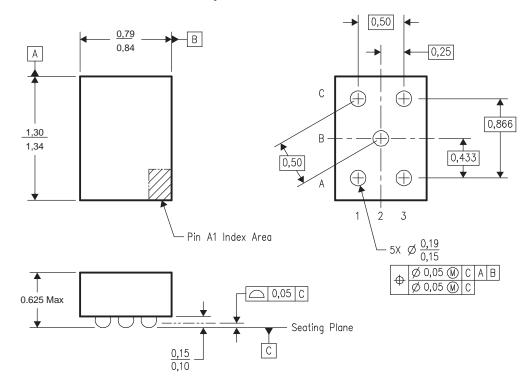
This glossary lists and explains terms, acronyms, and definitions.

12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



12.1 TPS730YZQ Nanostar[™] Wafer Chip Scale Information



NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. NanoStar[™] package configuration.

D. This package is tin-lead (SnPb); consult the factory for availability of lead-free material.

NanoStar is a trademark of Texas Instruments.

Figure 25. Nanostar[™] Wafer Chip Scale Package



PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TPS73001DBVR	ACTIVE	SOT-23	DBV	6	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PGVI	Samples
TPS73001DBVRG4	ACTIVE	SOT-23	DBV	6	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PGVI	Samples
TPS73001DBVT	ACTIVE	SOT-23	DBV	6	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PGVI	Samples
TPS73001DBVTG4	ACTIVE	SOT-23	DBV	6	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PGVI	Samples
TPS73018DBVR	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PHHI	Samples
TPS73018DBVRG4	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PHHI	Samples
TPS73018DBVT	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PHHI	Samples
TPS73018DBVTG4	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PHHI	Samples
TPS73025DBVR	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PGWI	Samples
TPS73025DBVRG4	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PGWI	Samples
TPS73025DBVT	OBSOLETE	SOT-23	DBV	5		TBD	Call TI	Call TI	-40 to 125	PGWI	
TPS730285DBVR	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PHII	Samples
TPS730285DBVT	OBSOLETE	SOT-23	DBV	5		TBD	Call TI	Call TI	-40 to 125	PHII	
TPS73028DBVR	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PGXI	Samples
TPS73028DBVRG4	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PGXI	Samples
TPS73028DBVT	OBSOLETE	SOT-23	DBV	5		TBD	Call TI	Call TI	-40 to 125	PGXI	
TPS73028YZQT	OBSOLETE	DSBGA	YZQ	5		TBD	Call TI	Call TI	-40 to 125	E2	
TPS73030DBVR	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PGYI	Samples
TPS73033DBVR	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PHUI	Samples
TPS73033DBVT	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PHUI	Samples
TPS73033DBVTG4	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PHUI	Samples



Orderable Device	Status (1)	Package Type	e Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
							(6)				
TPS73047DBVR	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PETI	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

Texas

STRUMENTS

TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

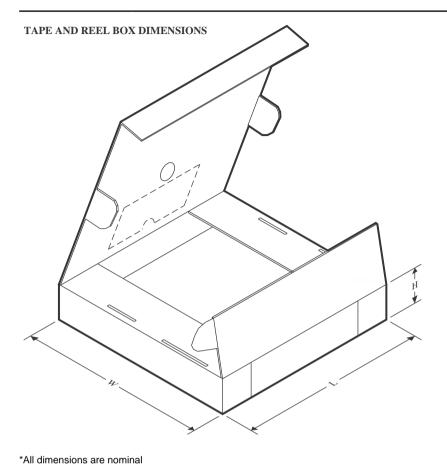


Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS73001DBVR	SOT-23	DBV	6	3000	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TPS73001DBVR	SOT-23	DBV	6	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPS73001DBVRG4	SOT-23	DBV	6	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPS73001DBVRG4	SOT-23	DBV	6	3000	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TPS73001DBVT	SOT-23	DBV	6	250	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TPS73018DBVR	SOT-23	DBV	5	3000	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TPS73018DBVT	SOT-23	DBV	5	250	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TPS73025DBVR	SOT-23	DBV	5	3000	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TPS730285DBVR	SOT-23	DBV	5	3000	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TPS73028DBVR	SOT-23	DBV	5	3000	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TPS73030DBVR	SOT-23	DBV	5	3000	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TPS73033DBVR	SOT-23	DBV	5	3000	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TPS73033DBVT	SOT-23	DBV	5	250	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TPS73047DBVR	SOT-23	DBV	5	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3



PACKAGE MATERIALS INFORMATION

18-Jan-2025



All dimensions are nominal							
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS73001DBVR	SOT-23	DBV	6	3000	180.0	180.0	18.0
TPS73001DBVR	SOT-23	DBV	6	3000	200.0	183.0	25.0
TPS73001DBVRG4	SOT-23	DBV	6	3000	200.0	183.0	25.0
TPS73001DBVRG4	SOT-23	DBV	6	3000	180.0	180.0	18.0
TPS73001DBVT	SOT-23	DBV	6	250	180.0	180.0	18.0
TPS73018DBVR	SOT-23	DBV	5	3000	180.0	180.0	18.0
TPS73018DBVT	SOT-23	DBV	5	250	180.0	180.0	18.0
TPS73025DBVR	SOT-23	DBV	5	3000	180.0	180.0	18.0
TPS730285DBVR	SOT-23	DBV	5	3000	180.0	180.0	18.0
TPS73028DBVR	SOT-23	DBV	5	3000	180.0	180.0	18.0
TPS73030DBVR	SOT-23	DBV	5	3000	180.0	180.0	18.0
TPS73033DBVR	SOT-23	DBV	5	3000	180.0	180.0	18.0
TPS73033DBVT	SOT-23	DBV	5	250	180.0	180.0	18.0
TPS73047DBVR	SOT-23	DBV	5	3000	200.0	183.0	25.0

DBV0006A



PACKAGE OUTLINE

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.2. This drawing is subject to change without notice.3. Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.25 per side.

- 4. Leads 1,2,3 may be wider than leads 4,5,6 for package orientation.
- 5. Refernce JEDEC MO-178.



DBV0006A

EXAMPLE BOARD LAYOUT

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



DBV0006A

EXAMPLE STENCIL DESIGN

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

9. Board assembly site may have different recommendations for stencil design.



DBV0005A



PACKAGE OUTLINE

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
 This drawing is subject to change without notice.
 Reference JEDEC MO-178.

- 4. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25 mm per side.
- 5. Support pin may differ or may not be present.



DBV0005A

EXAMPLE BOARD LAYOUT

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



DBV0005A

EXAMPLE STENCIL DESIGN

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



NOTES: (continued)

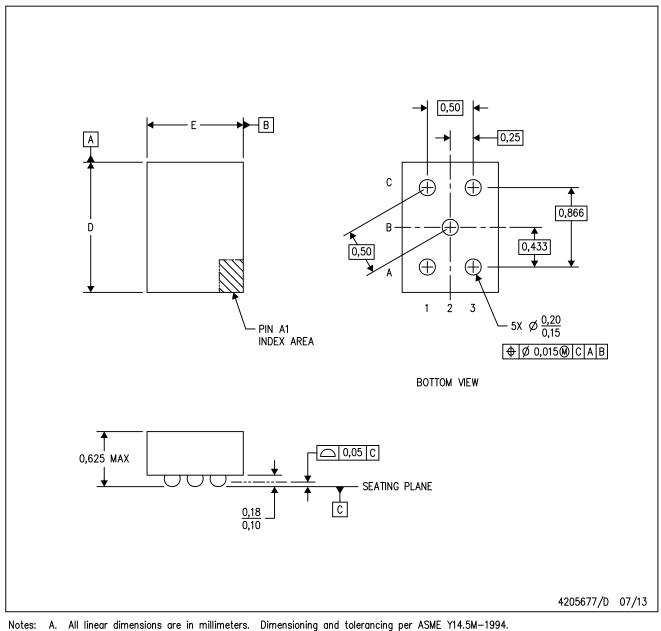
8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

9. Board assembly site may have different recommendations for stencil design.



YZQ (R-XBGA-N5)

DIE-SIZE BALL GRID ARRAY



- Β.
- This drawing is subject to change without notice. NanoFree™ package configuration. C.

NanoFree is a trademark of Texas Instruments.



IMPORTANT NOTICE AND DISCLAIMER

TI PROVIDES TECHNICAL AND RELIABILITY DATA (INCLUDING DATA SHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS AND IMPLIED, INCLUDING WITHOUT LIMITATION ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE OR NON-INFRINGEMENT OF THIRD PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for skilled developers designing with TI products. You are solely responsible for (1) selecting the appropriate TI products for your application, (2) designing, validating and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, regulatory or other requirements.

These resources are subject to change without notice. TI grants you permission to use these resources only for development of an application that uses the TI products described in the resource. Other reproduction and display of these resources is prohibited. No license is granted to any other TI intellectual property right or to any third party intellectual property right. TI disclaims responsibility for, and you will fully indemnify TI and its representatives against, any claims, damages, costs, losses, and liabilities arising out of your use of these resources.

TI's products are provided subject to TI's Terms of Sale or other applicable terms available either on ti.com or provided in conjunction with such TI products. TI's provision of these resources does not expand or otherwise alter TI's applicable warranties or warranty disclaimers for TI products.

TI objects to and rejects any additional or different terms you may have proposed.

Mailing Address: Texas Instruments, Post Office Box 655303, Dallas, Texas 75265 Copyright © 2025, Texas Instruments Incorporated