

# SN74LV8T595-EP Enhanced Product, 8-Bit Shift Registers With 3-State Output And **Logic Level Shifter**

#### 1 Features

- Wide operating range of 1.65V to 5.5V
- 5.5V tolerant input pins
- Single-supply voltage translator (refer to LVxT Enhanced Input Voltage):
  - Up translation:
    - 1.2V to 1.8V
    - 1.5V to 2.5V
    - 1.8V to 3.3V
    - 3.3V to 5.0V
  - Down translation:
    - 5.0V, 3.3V, 2.5V to 1.8V
    - 5.0V, 3.3V to 2.5V
    - 5.0V to 3.3V
- Up to 150Mbps with 5V or 3.3V  $V_{CC}$
- Supports standard function pinout
- Latch-up performance exceeds 250mA per JESD 17
- Supports defense and aerospace applications:
  - Controlled baseline
  - One assembly and test site
  - One fabrication site
  - Extended product life cycle
  - Product traceability

# 2 Applications

- Enable or disable a digital signal
- Controlling an indicator LED
- Translation between communication modules and system controllers

# 3 Description

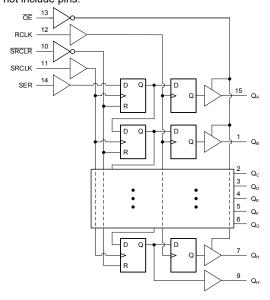
The SN74LV8T595-EP device contains an 8-bit, serial-in, parallel-out shift register that feeds an 8bit D-type storage register. The storage register has parallel 3-state outputs. Separate clocks are provided for both the shift and storage register. The shift register has a direct overriding clear ( SRCLR) input, serial (SER) input, and a serial output (QH) for cascading. When the output-enable (  $\overline{OE}$ ) input is high, the outputs are in a high-impedance state. Internal register data is not impacted by the operation of the OE input. The output level is referenced to the supply voltage (V<sub>CC</sub>) and supports 1.8V, 2.5V, 3.3V, and 5V CMOS levels.

The input is designed with a lower threshold circuit to support up translation for lower voltage CMOS inputs (for example, 1.2V input to 1.8V output or 1.8V input to 3.3V output). In addition, the 5V tolerant input pins enable down translation (for example, 3.3V to 2.5V output).

#### **Package Information**

PART NUMBER	PACKAGE <sup>(1)</sup>	PACKAGE SIZE <sup>(2)</sup>	BODY SIZE (NOM)(3)	
SN74LV8T595-EP	PW (TSSOP, 16)	5mm × 6.4mm	5mm × 4.4mm	

- For more information, see the Mechanical, Packaging, and Orderable Information section.
- The package size (length × width) is a nominal value and (2) includes pins, where applicable.
- The body size (length × width) is a nominal value and does not include pins.



Simplified Logic Diagram (Positive Logic)



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# **4 Pin Configuration and Functions**

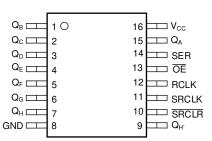


Figure 4-1. PW Package, 16-Pin TSSOP (Top View)

**Table 4-1. Pin Functions** 

P	IN	TYPE <sup>(1)</sup>	DESCRIPTION
NAME	NO.	ITPE'	DESCRIPTION
Q <sub>B</sub>	1	0	Q <sub>B</sub> output
Q <sub>C</sub>	2	0	Q <sub>C</sub> output
$Q_D$	3	0	Q <sub>D</sub> output
Q <sub>E</sub>	4	0	Q <sub>E</sub> output
Q <sub>F</sub>	5	0	Q <sub>F</sub> output
$Q_G$	6	0	Q <sub>G</sub> output
Q <sub>H</sub>	7	0	Q <sub>H</sub> output
GND	8	G	Ground
Q <sub>H</sub> '	9	0	Serial output, can be used for cascading
SRCLR	10	I	Shift register clear, active low
SRCLK	11	I	Shift register clock, rising edge triggered
RCLK	12	I	Output register clock, rising edge triggered
ŌĒ	13	I	Output Enable, active low
SER	14	I	Serial input
Q <sub>A</sub>	15	0	Q <sub>A</sub> output
V <sub>CC</sub>	16	Р	Positive supply

<sup>(1)</sup> I = input, O = output, I/O = input or output, G = ground, P = power.



# **5 Specifications**

# 5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1)

			MIN	MAX	UNIT
V <sub>CC</sub>	Supply voltage range		-0.5	7	V
VI	Input voltage range <sup>(2)</sup>		-0.5	7	V
Vo	Voltage range applied to any outp	ut in the high-impedance or power-off state <sup>(2)</sup>	-0.5	7	V
Vo	Output voltage range <sup>(2)</sup>		-0.5	V <sub>CC</sub> + 0.5	V
I <sub>IK</sub>	Input clamp current	V <sub>I</sub> < -0.5V		-20	mA
I <sub>OK</sub>	Output clamp current	$V_{O}$ < -0.5V or $V_{O}$ > $V_{CC+}$ 0.5V		±20	mA
Io	Continuous output current	$V_O = 0$ to $V_{CC}$		±25	mA
	Continuous output current through	N V <sub>CC</sub> or GND		±50	mA
T <sub>stg</sub>	Storage temperature		-65	150	°C

<sup>(1)</sup> Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute maximum ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If briefly operating outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not sustain damage, but it may not be fully functional. Operating the device in this manner may affect device reliability, functionality, performance, and shorten the device lifetime.

# 5.2 ESD Ratings

			VALUE	UNIT
V	Electrostatic	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±2000	\/
V <sub>(ESD)</sub>	discharge	Charged-device model (CDM), per ANSI/ESDA/JEDEC JS-002 <sup>(2)</sup>	±1000	v

<sup>(1)</sup> JEDEC document JEP155 states that 500V HBM allows safe manufacturing with a standard ESD control process.

# **5.3 Recommended Operating Conditions**

over operating free-air temperature range (unless otherwise noted)

Specification	Description	Condition	MIN	MAX	UNIT
V <sub>CC</sub>	Supply voltage		1.8	5.5	V
VI	Input voltage		0	5.5	V
Vo	Output voltage		0	V <sub>cc</sub>	V
	High-level input voltage	V <sub>CC</sub> = 1.65V to 2V	1.1		
V		V <sub>CC</sub> = 2.25V to 2.75V	1.28		V
$V_{IH}$		V <sub>CC</sub> = 3V to 3.6V	1.45		V
		V <sub>CC</sub> = 4.5V to 5.5V	2		
		V <sub>CC</sub> = 1.65V to 2V		0.5	
V	Low-Level input voltage	V <sub>CC</sub> = 2.25V to 2.75V		0.65	V
$V_{IL}$	Low-Level Input Voltage	V <sub>CC</sub> = 3V to 3.6V		0.75	V
		V <sub>CC</sub> = 4.5V to 5.5V		0.85	
		V <sub>CC</sub> = 1.6V to 2V		±3	
Io	Output current	V <sub>CC</sub> = 2.25V to 2.75V		±7	mA
		V <sub>CC</sub> = 3.3V to 5.0V		±15	
Io	Output Current	V <sub>CC</sub> = 4.5V to 5.5V		±25	mA
Δt/Δν	Input transition rise or fall rate	V <sub>CC</sub> = 1.6V to 5.0V		20	ns/V
T <sub>A</sub>	Operating free-air temperature		-55	125	°C

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<sup>(2)</sup> The input and output voltage ratings may be exceeded if the input and output current ratings are observed.

<sup>(2)</sup> JEDEC document JEP157 states that 250V CDM allows safe manufacturing with a standard ESD control process.



# **5.4 Thermal Information**

DVCKVCE	PACKAGE PINS	THERMAL METRIC (1)						
PACKAGE		$R_{\theta JA}$	R <sub>0JC(top)</sub>	$R_{\theta JB}$	$\Psi_{JT}$	$\Psi_{JB}$	R <sub>0JC(bot)</sub>	UNIT
PW (TSSOP)	16	117.4	52.5	75.2	4.7	74.5	-	°C/W

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application note.

# **5.5 Electrical Characteristics**

over operating free-air temperature range; typical ratings measured at  $T_A$  = 25°C (unless otherwise noted).

PARAMETE	R TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT	
	I <sub>OH</sub> = -50μA	1.2V to 5.5V	V <sub>CC</sub> -0.2				
	I <sub>OH</sub> = -1mA	1.2V	0.8				
	I <sub>OH</sub> = -2mA	1.65V to 2V	1.21	1.7 <sup>(1)</sup>			
V <sub>OH</sub>	I <sub>OH</sub> = -3mA	2.25V to 2.75V	1.93	2.4 <sup>(1)</sup>		V	
	I <sub>OH</sub> = -5.5mA	3V to 3.6V	2.49	3.08 <sup>(1)</sup>			
	I <sub>OH</sub> = -8mA	4.5V to 5.5V	3.95	4.65 <sup>(1)</sup>			
	I <sub>OH</sub> = –24mA	4.5V to 5.5V	3.15				
	I <sub>OL</sub> = 50μA	1.2V to 5.5V			0.1		
	I <sub>OL</sub> = 1mA	1.2 V			0.2		
	I <sub>OL</sub> = 2mA	1.65V to 2V		0.1(1)	0.25		
V <sub>OL</sub>	I <sub>OL</sub> = 3mA	2.25V to 2.75V		0.1 <sup>(1)</sup>	0.2	V	
	I <sub>OL</sub> = 5.5mA	3V to 3.6V		0.2(1)	0.25		
	I <sub>OL</sub> = 8mA	4.5V to 5.5V		0.3(1)	0.35		
	I <sub>OL</sub> = 24mA	4.5V to 5.5V			0.75		
I <sub>I</sub>	V <sub>I</sub> = 0V or V <sub>CC</sub>	0V to 5.5V		±0.1	±1	μA	
Icc	$V_I = V_{CC}$ or GND, $I_O = 0$	1.8V to 5.5V		2	20	μA	
A.	One input at 0.3V or 3.4V, other inputs at 0 or $V_{CC}$ , $I_{O}$ = 0	5.5V		1.35	1.5	mA	
ΔI <sub>CC</sub>	One input at 0.3V or 1.1V, other inputs at 0 or $V_{CC}$ , $I_{O} = 0$	1.8V			68	μA	
C <sub>I</sub>	V <sub>I</sub> = V <sub>CC</sub> or GND	5V		3	5	pF	
Co	V <sub>O</sub> = V <sub>CC</sub> or GND	5V		5	8	pF	
l <sub>oz</sub>	$V_O = V_{CC}$ or GND and $V_{CC} = 5.5V$	5.5V			±2.5	μΑ	
C <sub>PD</sub> (2) (3)	C <sub>L</sub> = 50pF, F = 10MHz	1.8V to 5.5V		·	200	pF	

- Typical value at nearest nominal voltage (1.8V, 2.5V, 3.3V, and 5V)  $C_{PD}$  is used to determine the dynamic power consumption, per channel.  $P_D = V_{CC}^2 x F_I x (C_{PD} + C_L)$  where  $F_I =$  input frequency,  $C_L =$  output load capacitance,  $V_{CC} =$  supply voltage.

# **5.6 Switching Characteristics**

over operating free-air temperature range; typical values measured at  $T_A = 25$ °C (unless otherwise noted)

PARAMETER	FROM (INPUT)	TO (OUTPUT)	Load Capacitance	Vaa	-55°C to 125°C			UNIT
PARAMETER	PROW (INPOT)	10 (001701)	Load Capacitance	V <sub>cc</sub>	MIN	TYP	MAX	
T <sub>PZL</sub>	OE	Q	CL = 15pF	1.8V		25.3	29.5	ns
T <sub>PZH</sub>	OE	Q	CL = 15pF	1.8V		25	29.0	ns
T <sub>PLZ</sub>	OE	Q	CL = 15pF	1.8V		20.4	22.5	ns
T <sub>PHZ</sub>	OE	Q	CL = 15pF	1.8V		22.8	26.0	ns
T <sub>PLH</sub>	RCLK	QA-QH	CL = 15pF	1.8V		25	29.0	ns



over operating free-air temperature range; typical values measured at  $T_A$  = 25 $^{\circ}$ C (unless otherwise noted)

			es measured at T <sub>A</sub> =		-55°C to 125°C		,C	
PARAMETER	FROM (INPUT)	TO (OUTPUT)	Load Capacitance	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
T <sub>PHL</sub>	RCLK	QA-QH	CL = 15pF	1.8V		36.5	43.0	ns
T <sub>PLH</sub>	SRCLK	QH'	CL = 15pF	1.8V		24.5	28.5	ns
T <sub>PHL</sub>	SRCLK	QH'	CL = 15pF	1.8V		31	36.5	ns
T <sub>PHL</sub>	SRCLR	QH'	CL = 15pF	1.8V		36	42.0	ns
T <sub>PZL</sub>	OE	Q	C <sub>L</sub> = 50pF	1.8V	,	30.1	34.5	ns
T <sub>PZH</sub>	OE	Q	C <sub>L</sub> = 50pF	1.8V	,	28.6	33.0	ns
T <sub>PLZ</sub>	OE	Q	C <sub>L</sub> = 50pF	1.8V		26.9	30.0	ns
T <sub>PHZ</sub>	OE	Q	C <sub>L</sub> = 50pF	1.8V		29.2	32.5	ns
T <sub>PLH</sub>	RCLK	QA-QH	C <sub>L</sub> = 50pF	1.8V		28.5	33.0	ns
T <sub>PHL</sub>	RCLK	QA-QH	C <sub>L</sub> = 50pF	1.8V	,	41.4	48.0	ns
T <sub>PLH</sub>	SRCLK	QH'	C <sub>L</sub> = 50pF	1.8V		27.9	32.5	ns
T <sub>PHL</sub>	SRCLK	QH'	C <sub>L</sub> = 50pF	1.8V		35.6	41.5	ns
T <sub>PHL</sub>	SRCLR	QH'	C <sub>L</sub> = 50pF	1.8V		40.5	47.0	ns
T <sub>PZL</sub>	OE	Q	CL = 15pF	2.5V		14.4	18.0	ns
T <sub>PZH</sub>	OE	Q	CL = 15pF	2.5V		14	17.5	ns
T <sub>PLZ</sub>	OE	Q	CL = 15pF	2.5V		11.6	13.5	ns
T <sub>PHZ</sub>	OE	Q	CL = 15pF	2.5V		12.9	15.5	ns
T <sub>PLH</sub>	RCLK	QA-QH	CL = 15pF	2.5V		13.6	17.0	ns
T <sub>PHL</sub>	RCLK	QA-QH	CL = 15pF	2.5V	-	19.1	24.5	ns
T <sub>PLH</sub>	SRCLK	QH'	CL = 15pF	2.5V		13.1	15.5	ns
T <sub>PHL</sub>	SRCLK	QH'	CL = 15pF	2.5V		16.3	21.0	ns
T <sub>PHL</sub>	SRCLR	QH'	CL = 15pF	2.5V		19.4	24.5	ns
T <sub>PZL</sub>	OE	Q	C <sub>L</sub> = 50pF	2.5V		17.6	21.5	ns
T <sub>PZH</sub>	OE	Q	C <sub>L</sub> = 50pF	2.5V	-	16.6	20.5	ns
T <sub>PLZ</sub>	OE	Q	C <sub>L</sub> = 50pF	2.5V		16.1	18.0	ns
T <sub>PHZ</sub>	OE	Q	C <sub>L</sub> = 50pF	2.5V		17.5	20.0	ns
T <sub>PLH</sub>	RCLK	QA-QH	C <sub>L</sub> = 50pF	2.5V		16.1	19.5	ns
T <sub>PHL</sub>	RCLK	QA-QH	C <sub>L</sub> = 50pF	2.5V		22.2	28.0	ns
T <sub>PLH</sub>	SRCLK	QH'	C <sub>L</sub> = 50pF	2.5V	-	15.3	19.0	ns
T <sub>PHL</sub>	SRCLK	QH'	C <sub>L</sub> = 50pF	2.5V		19.4	24.0	ns
T <sub>PHL</sub>	SRCLR	QH'	C <sub>L</sub> = 50pF	2.5V		22.5	28.0	ns
T <sub>PZL</sub>	OE	Q	CL = 15pF	3.3V		10.2	13.0	ns
T <sub>PZH</sub>	OE	Q	CL = 15pF	3.3V		10.1	12.5	ns
T <sub>PLZ</sub>	OE	Q	CL = 15pF	3.3V		7.97	9.4	ns
T <sub>PHZ</sub>	OE	Q	CL = 15pF	3.3V		9.12	11.0	ns
T <sub>PLH</sub>	RCLK	QA-QH	CL = 15pF	3.3V		9.33	12.0	ns
T <sub>PHL</sub>	RCLK	QA-QH	CL = 15pF	3.3V		13.2	17.5	ns
T <sub>PLH</sub>	SRCLK	QH'	CL = 15pF	3.3V		9.01	11.5	ns
T <sub>PHL</sub>	SRCLK	QH'	CL = 15pF	3.3V		11.1	14.7	ns
T <sub>PHL</sub>	SRCLR	QH'	CL = 15pF	3.3V		13.5	17.0	ns
T <sub>PZL</sub>	OE	Q	C <sub>L</sub> = 50pF	3.3V		13	15.8	ns
T <sub>PZH</sub>	OE	Q	C <sub>L</sub> = 50pF	3.3V		12.4	15.0	ns
T <sub>PLZ</sub>	OE	Q	C <sub>L</sub> = 50pF	3.3V		11.5	13.5	ns
T <sub>PHZ</sub>	OE	Q	C <sub>L</sub> = 50pF	3.3V		12.5	14.5	ns



over operating free-air temperature range; typical values measured at  $T_A = 25$ °C (unless otherwise noted)

DADAMETED	EDOM (INDUT)	TO (OUTDUT)	Load Canasitanas	, v	-55°	C to 125°	C O	LINIT
PARAMETER	FROM (INPUT)	TO (OUTPUT)	Load Capacitance	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
T <sub>PLH</sub>	RCLK	QA-QH	C <sub>L</sub> = 50pF	3.3V		11.5	14.5	ns
T <sub>PHL</sub>	RCLK	QA-QH	C <sub>L</sub> = 50pF	3.3V		15.8	20.0	ns
T <sub>PLH</sub>	SRCLK	QH'	C <sub>L</sub> = 50pF	3.3V		10.8	13.5	ns
T <sub>PHL</sub>	SRCLK	QH'	C <sub>L</sub> = 50pF	3.3V		13.6	17.5	ns
T <sub>PHL</sub>	SRCLR	QH'	C <sub>L</sub> = 50pF	3.3V		15.8	20.0	ns
T <sub>PZL</sub>	OE	Q	CL = 15pF	5V		6.87	8.7	ns
T <sub>PZH</sub>	OE	Q	CL = 15pF	5V		7.05	8.5	ns
T <sub>PLZ</sub>	OE	Q	CL = 15pF	5V		5.95	6.8	ns
T <sub>PHZ</sub>	OE	Q	CL = 15pF	5V		6.75	8.0	ns
T <sub>PLH</sub>	RCLK	QA-QH	CL = 15pF	5V		7.17	8.7	ns
T <sub>PHL</sub>	RCLK	QA-QH	CL = 15pF	5V		9.6	12.5	ns
T <sub>PLH</sub>	SRCLK	QH'	CL = 15pF	5V		6.7	8.5	ns
T <sub>PHL</sub>	SRCLK	QH'	CL = 15pF	5V		7.98	10.5	ns
T <sub>PHL</sub>	SRCLR	QH'	CL = 15pF	5V		9	11.5	ns
T <sub>PZL</sub>	OE	Q	C <sub>L</sub> = 50pF	5V		9.23	11.0	ns
T <sub>PZH</sub>	OE	Q	C <sub>L</sub> = 50pF	5V		8.85	10.5	ns
T <sub>PLZ</sub>	OE	Q	C <sub>L</sub> = 50pF	5V		8.26	9.3	ns
T <sub>PHZ</sub>	OE	Q	C <sub>L</sub> = 50pF	5V		8.42	10.0	ns
T <sub>PLH</sub>	RCLK	QA-QH	C <sub>L</sub> = 50pF	5V		9	10.8	ns
T <sub>PHL</sub>	RCLK	QA-QH	C <sub>L</sub> = 50pF	5V		11.6	14.5	ns
T <sub>PLH</sub>	SRCLK	QH'	C <sub>L</sub> = 50pF	5V		8.3	10.0	ns
T <sub>PHL</sub>	SRCLK	QH'	C <sub>L</sub> = 50pF	5V		9.89	12.5	ns
T <sub>PHL</sub>	SRCLR	QH'	C <sub>L</sub> = 50pF	5V		10.8	13.5	ns

# **5.7 Timing Characteristics**

over recommended operating free-air temperature range (unless otherwise noted)

DADAMETED	DECODIDEION	COMPITION	.,	T <sub>A</sub> = 25	°C	-55°C to 125°C		UNIT
PARAMETER	DESCRIPTION	CONDITION	V <sub>cc</sub>	MIN	MAX	MIN	MAX	UNII
f <sub>CLOCK</sub>	Clock frequency		1.8V		13.5		11.4	MHz
t <sub>W</sub>	Pulse duration	RCLK or SRCLK high or low	1.8V	16		18		ns
t <sub>W</sub>	Pulse duration	SCRCLR low	1.8V	12		13		ns
t <sub>SU</sub>	Setup time	SER before SRCLK↑	1.8V	10		12		ns
t <sub>SU</sub>	Setup time	SRCLK↑ before RCLK↑	1.8V	29		51		ns
t <sub>SU</sub>	Setup time	SRCLR low before RCLK↑	1.8V	35		54		ns
t <sub>SU</sub>	Setup time	SRCLR high (inactive) before SRCLK↑	1.8V	8		10		ns
t <sub>H</sub>	Hold time	SER after SRCLK↑	1.8V	1		1		ns
f <sub>CLOCK</sub>	Clock frequency		2.5V		29		22	MHz
t <sub>W</sub>	Pulse duration	RCLK or SRCLK high or low	2.5V	11		13		ns
t <sub>W</sub>	Pulse duration	SCRCLR low	2.5V	9		10		ns
t <sub>SU</sub>	Setup time	SER before SRCLK↑	2.5V	8		9		ns
t <sub>SU</sub>	Setup time	SRCLK↑ before RCLK↑	2.5V	17		30		ns
t <sub>SU</sub>	Setup time	SRCLR low before RCLK↑	2.5V	20		33		ns
t <sub>SU</sub>	Setup time	SRCLR high (inactive) before SRCLK↑	2.5V	6		7		ns



over recommended operating free-air temperature range (unless otherwise noted)

DADAMETED	DECODIDATION	CONDITION	.,	T <sub>A</sub> = 25	5°C	-55°C to		
PARAMETER	DESCRIPTION	CONDITION	V <sub>CC</sub>	MIN	MAX	MIN	MAX	UNIT
t <sub>H</sub>	Hold time	SER after SRCLK↑	2.5V	1		1		ns
f <sub>CLOCK</sub>	Clock frequency		3.3V		46		32	MHz
t <sub>W</sub>	Pulse duration	RCLK or SRCLK high or low	3.3V	11		12		ns
t <sub>W</sub>	Pulse duration	SCRCLR low	3.3V	9		10		ns
t <sub>SU</sub>	Setup time	SER before SRCLK↑	3.3V	8		9		ns
t <sub>SU</sub>	Setup time	SRCLK↑ before RCLK↑	3.3V	13		21		ns
t <sub>SU</sub>	Setup time	SRCLR low before RCLK↑	3.3V	17		24		ns
t <sub>SU</sub>	Setup time	SRCLR high (inactive) before SRCLK↑	3.3V	6		7		ns
t <sub>H</sub>	Hold time	SER after SRCLK↑	3.3V	1	1 1			ns
f <sub>CLOCK</sub>	Clock frequency		5V		64		1	MHz
t <sub>W</sub>	Pulse duration	RCLK or SRCLK high or low	5V	11		12		ns
t <sub>W</sub>	Pulse duration	SCRCLR low	5V	9		10		ns
t <sub>SU</sub>	Setup time	SER before SRCLK↑	5V	7		8		ns
t <sub>SU</sub>	Setup time	SRCLK↑ before RCLK↑	5V	11		17		ns
t <sub>SU</sub>	Setup time	SRCLR low before RCLK↑	5V	12		18		ns
t <sub>SU</sub>	Setup time	SRCLR high (inactive) before SRCLK↑	5V	6	6 7			ns
t <sub>H</sub>	Hold time	SER after SRCLK↑	5V	1		1		ns

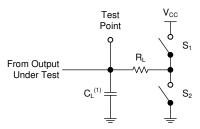


# **6 Parameter Measurement Information**

Phase relationships between waveforms were chosen arbitrarily for the examples listed in the following table. All input pulses are supplied by generators having the following characteristics: PRR  $\leq$  1MHz,  $Z_O = 50\Omega$ ,  $t_t < 2.5$ ns.

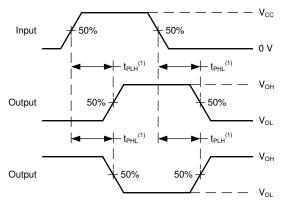
The outputs are measured individually with one input transition per measurement.

TEST	S1	S2	$R_L$	CL	ΔV	V <sub>cc</sub>
t <sub>PLH</sub> , t <sub>PHL</sub>	OPEN	OPEN	_	15pF, 50pF	_	ALL
t <sub>PLZ</sub> , t <sub>PZL</sub>	CLOSED	OPEN	1kΩ	15pF, 50pF	0.15V	≤ 2.5V
t <sub>PHZ</sub> , t <sub>PZH</sub>	OPEN	CLOSED	1kΩ	15pF, 50pF	0.15V	≤ 2.5V
t <sub>PLZ</sub> , t <sub>PZL</sub>	CLOSED	OPEN	1kΩ	15pF, 50pF	0.3V	> 2.5V
t <sub>PHZ</sub> , t <sub>PZH</sub>	OPEN	CLOSED	1kΩ	15pF, 50pF	0.3V	> 2.5V



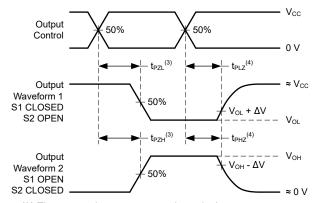
(1) C<sub>L</sub> includes probe and test-fixture capacitance.

Figure 6-1. Load Circuit for 3-State Outputs



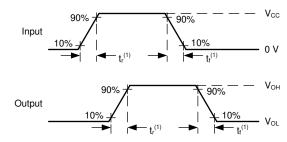
(1) The greater between  $t_{\text{PLH}}$  and  $t_{\text{PHL}}$  is the same as  $t_{\text{pd}}.$ 

Figure 6-2. Voltage Waveforms Propagation Delays



- (3) The greater between  $t_{\mbox{\scriptsize PZL}}$  and  $t_{\mbox{\scriptsize PZH}}$  is the same as  $t_{\mbox{\scriptsize en}}$ .
- (4) The greater between  $t_{\mbox{\scriptsize PLZ}}$  and  $t_{\mbox{\scriptsize PHZ}}$  is the same as  $t_{\mbox{\scriptsize dis}}.$

Figure 6-3. Voltage Waveforms Propagation Delays



(1) The greater between  $t_r$  and  $t_f$  is the same as  $t_t$ .

Figure 6-4. Voltage Waveforms, Input and Output Transition Times

# 7 Detailed Description

#### 7.1 Overview

The SN74LV8T595-EP device contains an 8-bit, serial-in, parallel-out shift register that feeds an 8-bit D-type storage register. The storage register has parallel 3-state outputs. Separate clocks are provided for both the shift and storage register. The shift register has a direct overriding clear ( $\overline{SRCLR}$ ) input, serial (SER) input, and a serial output ( $Q_{H'}$ ) for cascading. When the output-enable ( $\overline{OE}$ ) input is high, the storage register outputs are in a high-impedance state. Internal register data and serial output ( $Q_{H'}$ ) are not impacted by the operation of the  $\overline{OE}$  input.

# 7.2 Functional Block Diagram

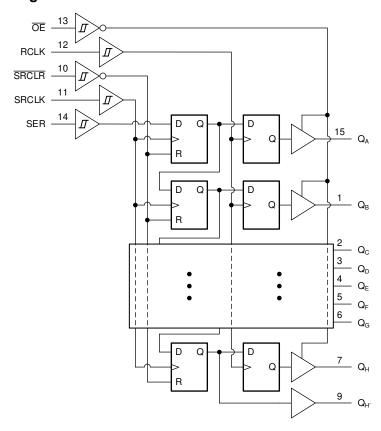


Figure 7-1. Logic Diagram (Positive Logic) for the SN74LV8T595-EP

# 7.3 Feature Description

#### 7.3.1 Balanced CMOS 3-State Outputs

This device includes balanced CMOS 3-state outputs. Driving high, driving low, and high impedance are the three states that these outputs can be in. The term *balanced* indicates that the device can sink and source similar currents. The drive capability of this device may create fast edges into light loads, so routing and load conditions should be considered to prevent ringing. Additionally, the outputs of this device can drive larger currents than the device can sustain without being damaged. It is important for the output power of the device to be limited to avoid damage due to overcurrent. The electrical and thermal limits defined in the *Absolute Maximum Ratings* must be followed at all times.

When placed into the high-impedance state, the output will neither source nor sink current, with the exception of minor leakage current as defined in the *Electrical Characteristics* table. In the high-impedance state, the output voltage is not controlled by the device and is dependent on external factors. If no other drivers are connected to the node, then this is known as a floating node and the voltage is unknown. A pull-up or pull-down resistor can be connected to the output to provide a known voltage at the output while it is in the high-impedance state.

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The value of the resistor will depend on multiple factors, including parasitic capacitance and power consumption limitations. Typically, a  $10k\Omega$  resistor can be used to meet these requirements.

Unused 3-state CMOS outputs should be left disconnected.



#### 7.3.2 LVxT Enhanced Input Voltage

The SN74LV8T595-EP belongs to TI's LVxT family of logic devices with integrated voltage level translation. This family of devices was designed with reduced input voltage thresholds to support up-translation, and inputs tolerant of signals with up to 5.5V levels to support down-translation. For proper functionality, input signals must remain at or above the specified  $V_{IH(MIN)}$  level for a HIGH input state, and at or below the specified  $V_{IL(MAX)}$  for a LOW input state. Figure 7-2 shows the typical  $V_{IH}$  and  $V_{IL}$  levels for the LVxT family of devices, as well as the voltage levels for standard CMOS devices for comparison.

The inputs are high impedance and are typically modeled as a resistor in parallel with the input capacitance given in the *Electrical Characteristics*. The worst case resistance is calculated with the maximum input voltage, given in the *Absolute Maximum Ratings*, and the maximum input leakage current, given in the *Electrical Characteristics*, using Ohm's law  $(R = V \div I)$ .

Input signals must transition between valid logic states quickly, as defined by the input transition rate in the *Recommended Operating Conditions* table. Failing to meet this specification will result in excessive power consumption and could cause oscillations. More details can be found in the *Implications of Slow or Floating CMOS Inputs* application report.

Do not leave inputs floating at any time during operation. Unused inputs must be terminated at a valid high or low voltage level. If a system will not be actively driving an input at all times, then a pull-up or pull-down resistor can be added to provide a valid input voltage during these times. The resistor value will depend on multiple factors; however, a  $10k\Omega$  resistor is recommended and will typically meet all requirements.

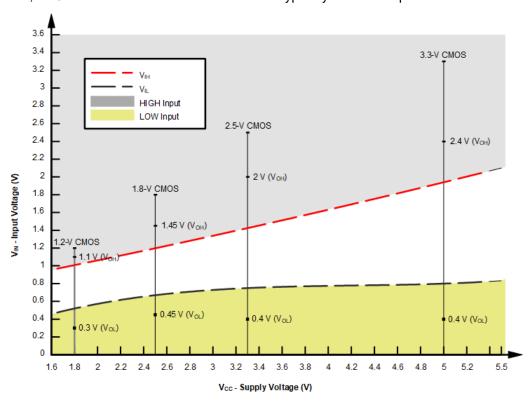


Figure 7-2. LVxT Input Voltage Levels



#### 7.3.2.1 Up Translation

Input signals can be up translated using the SN74LV8T595-EP. The voltage applied at V<sub>CC</sub> will determine the output voltage and the input thresholds as described in the Recommended Operating Conditions and Electrical Characteristics tables. When connected to a high-impedance input, the output voltage will be approximately V<sub>CC</sub> in the HIGH state, and 0V in the LOW state.

The inputs have reduced thresholds that allow for input HIGH state levels, which are much lower than standard values. For example, standard CMOS inputs for a device operating at a 5V supply will have a V<sub>IH(MIN)</sub> of 3.5V. For the SN74LV8T595-EP, VIH(MIN) with a 5V supply is only 2V, which would allow for up-translation from a typical 2.5V to 5V signals.

Ensure that the input signals in the HIGH state are above V<sub>IH(MIN)</sub> and input signals in the LOW state are lower than  $V_{IL(MAX)}$  as shown in Figure 7-3.

Up Translation Combinations are as follows:

- 1.8V V<sub>CC</sub> Inputs from 1.2V
- 2.5V V<sub>CC</sub> Inputs from 1.8V
- $3.3 \mbox{V}_{CC}$  Inputs from 1.8V and 2.5V  $5.0 \mbox{V}_{CC}$  Inputs from 2.5V and 3.3V

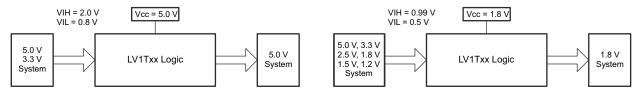


Figure 7-3. LVxT Up and Down Translation Example

#### 7.3.2.2 Down Translation

Signals can be translated down using the SN74LV8T595-EP. The voltage applied at the V<sub>CC</sub> will determine the output voltage and the input thresholds as described in the Recommended Operating Conditions and Electrical Characteristics tables.

When connected to a high-impedance input, the output voltage will be approximately V<sub>CC</sub> in the HIGH state, and 0V in the LOW state. Ensure that the input signals in the HIGH state are between  $V_{IH(MIN)}$  and 5.5V, and input signals in the LOW state are lower than V<sub>IL(MAX)</sub> as shown in Figure 7-2.

For example, standard CMOS inputs for devices operating at 5.0V, 3.3V or 2.5V can be down-translated to match 1.8V CMOS signals when operating from 1.8V V<sub>CC</sub>. See Figure 7-3.

Down Translation Combinations are as follows:

- 1.8V V<sub>CC</sub> Inputs from 2.5V, 3.3V, and 5.0V
- 2.5V V<sub>CC</sub> Inputs from 3.3V and 5.0V
- $3.3V V_{CC}$  Inputs from 5.0V

#### 7.3.3 Clamp Diode Structure

As Figure 7-4 shows, the outputs to this device have both positive and negative clamping diodes, and the inputs to this device have negative clamping diodes only.

#### CAUTION

Voltages beyond the values specified in the Absolute Maximum Ratings table can cause damage to the device. The input and output voltage ratings may be exceeded if the input and output clampcurrent ratings are observed.

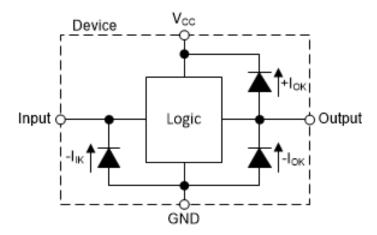


Figure 7-4. Electrical Placement of Clamping Diodes for Each Input and Output

# 7.4 Device Functional Modes

The table below lists the functional modes of the SN74LV8T595-EP.

**Table 7-1. Function Table** 

		INPUTS			FUNCTION
SER	SRCLK	SRCLR	RCLK	ŌĒ	FUNCTION
Х	Х	Х	Х	Н	Outputs Q <sub>A</sub> – Q <sub>H</sub> are disabled
Х	Х	Х	Х	L	Outputs Q <sub>A</sub> – Q <sub>H</sub> are enabled.
Х	Х	L	Х	Х	Shift register is cleared.
L	1	Н	Х	х	First stage of the shift register goes low. Other stages store the data of previous stage, respectively.
Н	1	Н	х	х	First stage of the shift register goes high. Other stages store the data of previous stage, respectively.
Х	Х	Н	1	Х	Shift-register data is stored in the storage register.
Х	<b>↑</b>	Н	1	Х	Data in shift register is stored in the storage register, the data is then shifted through.



# 8 Application and Implementation

#### Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

# 8.1 Application Information

In this application, the SN74LV8T595-EP is used to control seven-segment displays. Utilizing the serial output and combining a few of the input signals, this implementation reduces the number of I/O pins required to control the displays from sixteen to four. Unlike other I/O expanders, the SN74LV8T595-EP does not need a communication interface for control. It can be easily operated with simple GPIO pins.

The OE pin is used to easily disable the outputs when the displays need to be turned off or connected to a PWM signal to control brightness. However, this pin can be tied low and the outputs of the SN74LV8T595-EP can be controlled accordingly to turn off all the outputs reducing the I/O needed to three. There is no practical limitation to how many SN74LV8T595-EP devices can be cascaded. To add more, the serial output will need to be connected to the following serial input and the clocks will need to be connected accordingly. With separate control for the shift registers and output registers, the desired digit can be displayed while the data for the next digit is loaded into the shift register.

At power-up, the initial state of the shift registers and output registers are unknown. To give them a defined state, the shift register needs to be cleared and then clocked into the output register. An RC circuit can be connected to the  $\overline{SRCLR}$  pin as shown in the *Typical Application Block Diagram* to initialize the shift register to all zeros. With the  $\overline{OE}$  pin pulled up with a resistor, this process can be performed while the outputs are in a high impedance state eliminating any erroneous data causing issues with the displays.



# 8.2 Typical Application

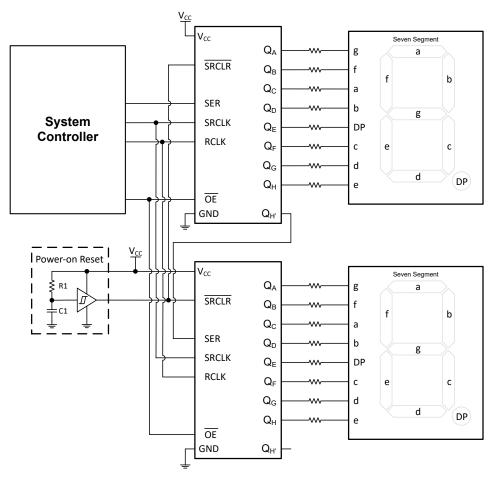


Figure 8-1. Typical Application Block Diagram



# 8.2.1 Design Requirements

#### 8.2.1.1 Power Considerations

Ensure the desired supply voltage is within the range specified in the *Recommended Operating Conditions*. The supply voltage sets the device's electrical characteristics of the device as described in the *Electrical Characteristics* section.

The positive voltage supply must be capable of sourcing current equal to the total current to be sourced by all outputs of the SN74LV8T595-EP plus the maximum static supply current,  $I_{CC}$ , listed in the *Electrical Characteristics*, and any transient current required for switching. The logic device can only source as much current that is provided by the positive supply source. Ensure the maximum total current through  $V_{CC}$  listed in the *Absolute Maximum Ratings* is not exceeded.

The ground must be capable of sinking current equal to the total current to be sunk by all outputs of the SN74LV8T595-EP plus the maximum supply current, I<sub>CC</sub>, listed in the *Electrical Characteristics*, and any transient current required for switching. The logic device can only sink as much current that can be sunk into its ground connection. Ensure the maximum total current through GND listed in the *Absolute Maximum Ratings* is not exceeded.

The SN74LV8T595-EP can drive a load with a total capacitance less than or equal to 50pF while still meeting all of the data sheet specifications. Larger capacitive loads can be applied; however, it is not recommended to exceed 50pF.

The SN74LV8T595-EP can drive a load with total resistance described by  $R_L \ge V_O$  /  $I_O$ , with the output voltage and current defined in the *Electrical Characteristics* table with  $V_{OH}$  and  $V_{OL}$ . When outputting in the HIGH state, the output voltage in the equation is defined as the difference between the measured output voltage and the supply voltage at the  $V_{CC}$  pin.

Total power consumption can be calculated using the information provided in *CMOS Power Consumption and Cpd Calculation*.

Thermal increase can be calculated using the information provided in *Thermal Characteristics of Standard Linear* and Logic (SLL) Packages and Devices.

### CAUTION

The maximum junction temperature,  $T_{J(max)}$  listed in the *Absolute Maximum Ratings*, is an additional limitation to prevent damage to the device. Do not violate any values listed in the *Absolute Maximum Ratings*. These limits are provided to prevent damage to the device.



#### 8.2.1.2 Input Considerations

Input signals must cross to be considered a logic LOW, and to be considered a logic HIGH. Do not exceed the maximum input voltage range found in the *Absolute Maximum Ratings*.

Unused inputs must be terminated to either  $V_{CC}$  or ground. The unused inputs can be directly terminated if the input is completely unused, or they can be connected with a pull-up or pull-down resistor if the input will be used sometimes, but not always. A pull-up resistor is used for a default state of HIGH, and a pull-down resistor is used for a default state of LOW. The drive current of the controller, leakage current into the SN74LV8T595-EP (as specified in the *Electrical Characteristics*), and the desired input transition rate limits the resistor size. A  $10k\Omega$  resistor value is often used due to these factors.

Refer to the *Feature Description* section for additional information regarding the inputs for this device.

#### 8.2.1.3 Output Considerations

The positive supply voltage is used to produce the output HIGH voltage. Drawing current from the output will decrease the output voltage as specified by the  $V_{OH}$  specification in the *Electrical Characteristics*. The ground voltage is used to produce the output LOW voltage. Sinking current into the output will increase the output voltage as specified by the  $V_{OH}$  specification in the *Electrical Characteristics*.

Push-pull outputs that could be in opposite states, even for a very short time period, should never be connected directly together. This can cause excessive current and damage to the device.

Two channels within the same device with the same input signals can be connected in parallel for additional output drive strength.

Unused outputs can be left floating. Do not connect outputs directly to V<sub>CC</sub> or ground.

Refer to the *Feature Description* section for additional information regarding the outputs for this device.



# 8.2.2 Detailed Design Procedure

- Add a decoupling capacitor from V<sub>CC</sub> to GND. The capacitor needs to be placed physically close to the device and electrically close to both the V<sub>CC</sub> and GND pins. An example layout is shown in the *Layout* section.
- 2. Ensure the capacitive load at the output is ≤ 50pF. This is not a hard limit; by design, however, it will optimize performance. This can be accomplished by providing short, appropriately sized traces from the SN74LV8T595-EP to one or more of the receiving devices.
- 3. Ensure the resistive load at the output is larger than  $(V_{CC} / I_{O(max)})\Omega$ . Doing this will prevent the maximum output current from the *Absolute Maximum Ratings* from being violated. Most CMOS inputs have a resistive load measured in M $\Omega$ ; much larger than the minimum calculated previously.
- 4. Thermal issues are rarely a concern for logic gates; the power consumption and thermal increase, however, can be calculated using the steps provided in the application report, *CMOS Power Consumption and Cpd Calculation*.

# 8.2.3 Application Curves

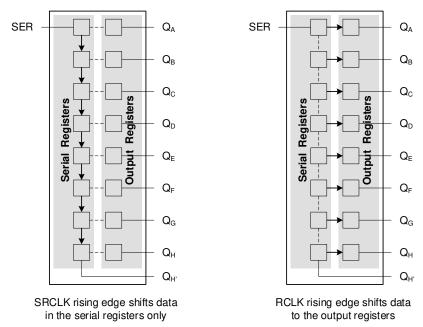


Figure 8-2. Simplified Functional Diagram Showing Clock Operation

# 8.3 Power Supply Recommendations

The power supply can be any voltage between the minimum and maximum supply voltage rating located in the *Recommended Operating Conditions*. Each  $V_{CC}$  terminal should have a good bypass capacitor to prevent power disturbance. A  $0.1\mu F$  capacitor is recommended for this device. It is acceptable to parallel multiple bypass capacitors to reject different frequencies of noise. The  $0.1\mu F$  and  $1\mu F$  capacitors are commonly used in parallel. The bypass capacitor should be installed as close to the power terminal as possible for best results.

# 8.4 Layout

#### 8.4.1 Layout Guidelines

- Bypass capacitor placement
  - Place near the positive supply terminal of the device
  - Provide an electrically short ground return path
  - Use wide traces to minimize impedance
  - Keep the device, capacitors, and traces on the same side of the board whenever possible
- Signal trace geometry



- 8mil to 12mil trace width
- Lengths less than 12cm to minimize transmission line effects
- Avoid 90° corners for signal traces
- Use an unbroken ground plane below signal traces
- Flood fill areas around signal traces with ground
- For traces longer than 12cm
  - Use impedance controlled traces
  - · Source-terminate using a series damping resistor near the output
  - · Avoid branches; buffer signals that must branch separately

# 8.4.2 Layout Example

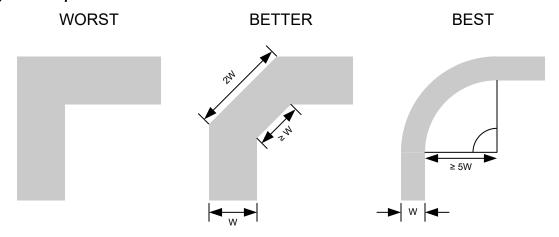


Figure 8-3. Example Trace Corners for Improved Signal Integrity

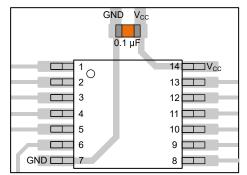


Figure 8-4. Example Bypass Capacitor Placement for TSSOP and Similar Packages

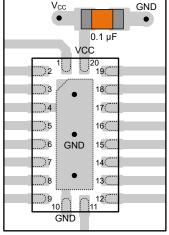


Figure 8-5. Example Bypass Capacitor Placement for WQFN and Similar Packages

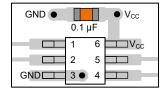


Figure 8-6. Example Bypass Capacitor Placement for SOT, SC70 and Similar Packages



Figure 8-7. Example Damping Resistor Placement for Improved Signal Integrity



# 9 Device and Documentation Support

TI offers an extensive line of development tools. Tools and software to evaluate the performance of the device, generate code, and develop solutions are listed below.

# 9.1 Documentation Support

#### 9.1.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, CMOS Power Consumption and Cpd Calculation application report
- Texas Instruments, *Designing With Logic* application report
- Texas Instruments, Thermal Characteristics of Standard Linear and Logic (SLL) Packages and Devices application report

# 9.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

# 9.3 Support Resources

TI E2E<sup>™</sup> support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

#### 9.4 Trademarks

TI E2E<sup>™</sup> is a trademark of Texas Instruments.

All trademarks are the property of their respective owners.

## 9.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### 9.6 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

# 10 Revision History

DATE	REVISION	NOTES				
January 2025	*	Initial Release				

# 11 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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#### PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
							(6)				
SN74LV8T595MPWREP	ACTIVE	TSSOP	PW	16	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM		LT595EP	Samples

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

#### OTHER QUALIFIED VERSIONS OF SN74LV8T595-EP:

# **PACKAGE OPTION ADDENDUM**

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● Catalog : SN74LV8T595

Automotive : SN74LV8T595-Q1

NOTE: Qualified Version Definitions:

- Catalog TI's standard catalog product
- Automotive Q100 devices qualified for high-reliability automotive applications targeting zero defects

# **PACKAGE MATERIALS INFORMATION**

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# TAPE AND REEL INFORMATION



# TAPE DIMENSIONS + K0 - P1 - B0 W Cavity - A0 -

	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

# QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device		Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
SN74LV8T595MPWREP	TSSOP	PW	16	3000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1

# PACKAGE MATERIALS INFORMATION

www.ti.com 16-Feb-2025



# \*All dimensions are nominal

Device Package Ty		Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)	
SN74LV8T595MPWREP	TSSOP	PW	16	3000	353.0	353.0	32.0	



SMALL OUTLINE PACKAGE



### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-153.



SMALL OUTLINE PACKAGE



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE PACKAGE



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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