







TEXAS INSTRUMENTS

**LSF0002** SDLS975 – APRIL 2024

# LSF0002 Ultra-Small Auto-Bidirectional Translator for Open Drain or Push Pull Applications

## 1 Features

- Provides bidirectional voltage translation with no direction pin
- Supports up to 100MHz up translation and greater than 100MHz down translation at ≤ 30pF capacitive load and up To 40MHz up or down translation at 50pF capacitive load
- Allows bidirectional voltage-level translation between
  - 0.95V ↔ 1.8/2.5/3.3/5V
  - 1.2V ↔ 1.8/2.5/3.3/5V
  - $\quad 1.8V \leftrightarrow 2.5/3.3/5V$
  - 2.5V  $\leftrightarrow$  3.3/5V
  - 3.3V  $\leftrightarrow$  5V
- Low standby current
- 5V tolerance I/O port to support TTL
- Low R<sub>ON</sub> provides less signal distortion
- High-impedance I/O pins for EN = Low
- Flow-through pinout for easy PCB trace routing
- Latch-up performance >100mA per JESD 17
- –40°C to 125°C operating temperature range

## 2 Applications

- GPIO, MDIO, PMBus, SMBus, SDIO, UART, I<sup>2</sup>C, and other interfaces in telecom infrastructure
- Enterprise systems
- Communications equipment
- Personal electronics
- Industrial applications

## **3 Description**

The LSF0002 supports bidirectional voltage translation without the need for DIR pin, which minimizes system effort (for PMBus,  $I^2$ C, SMBus, and so forth). The LSF family of devices supports up to 100MHz up translation and greater than 100MHz down translation at  $\leq$  30pF capacitive load and up to 40MHz up or down translation at 50pF capacitive load, which allows the LSF family to support more consumer or telecom interfaces (MDIO or SDIO).

LSF family supports 5V tolerance on I/O port, which makes it compatible with TTL levels in industrial and telecom applications. The LSF family can set up different voltage translation levels, which makes it very flexible.

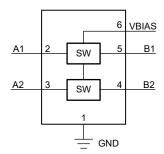
Unlike the LSF0x0x family, the LSF0002 does not require VREF\_A and VREF\_B power supplies and the 200k $\Omega$  bias resistor. The LSF0002 utilizes the V<sub>BIAS</sub> pin that enables translation by being biased to the same voltage as the lower power supply at the I/Os that is being translated to and from.

#### **Package Information**

PART NUMBER	PACKAGE <sup>(1)</sup>	PACKAGE SIZE <sup>(2)</sup>
LSF0002	DTQ (X2SON, 6)	1mm × 0.8mm

(1) For more information, see Section 11.

(2) The package size (length × width) is a nominal value and includes pins, where applicable.



**Functional Block Diagram** 



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## **4** Pin Configuration and Functions

Pinout drawings are not to scale



### Figure 4-1. LSF0002 DTQ Package, 6-Pin X2SON (Top View)

#### Table 4-1. Pin Functions

F	PIN	TYPE <sup>(1)</sup>	DESCRIPTION	
NAME	NO.		DESCRIPTION	
A1	2	I/O		
A2	3	I/O	Auto Didirectional Data part	
B1	5	I/O	Auto-Bidirectional Data port	
B2	4	I/O		
VBIAS	6		Enable input/ Supply Voltage	
GND	1	_	Ground	

(1) I= input, O = output



## **5** Specifications

#### 5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

			MIN	MAX	UNIT
VI	Input voltage <sup>(2)</sup>		-0.5	7	V
V <sub>I/O</sub>	Input/output voltage <sup>(2)</sup>		-0.5	7	V
	Continuous channel current			128	mA
I <sub>IK</sub>	Input clamp current	V <sub>1</sub> < 0		-50	mA
TJ	Junction temperature			150	°C
T <sub>stg</sub>	Storage temperature		-65	150	°C

(1) Operation outside the Absolute Maximum Rating may cause permanent device damage. Absolute Maximum Rating do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Condition. If used outside the Recommended Operating Condition but within the Absolute Maximum Rating, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.

(2) The input and input/output voltage ratings may be exceeded if the input and input/output clamp-current ratings are observed.

#### 5.2 ESD Ratings

			VALUE	UNIT
V <sub>(ESD)</sub>	Electrostatic discharge	Human body model (HBM), per AEC Q100-002 <sup>(3)</sup>	±2000	V
V <sub>(ESD)</sub>	Electrostatic discharge	Charged device model (CDM), per AEC Q100-001	±1000	V
N	V <sub>(ESD)</sub> Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±2000	V
V(ESD)		Charged device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup>	±1000	V

(1) JEDEC document JEP155 states that 500V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250V CDM allows safe manufacturing with a standard ESD control process.

(3) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

#### **5.3 Recommended Operating Conditions**

over operating free-air temperature range (unless otherwise noted)

	PARAMETER		MIN	MAX	UNIT
V <sub>BIAS</sub>	Reference Voltage		0	5.5	V
V <sub>I/O</sub>	Input/Output voltage	A1, A2, B1, B2	0	5.5	V
IPASS	Pass switch current			64	mA
T <sub>A</sub>	Operating free-air temperature		-40	125	°C



### 5.4 Thermal Information

		LSF0002	
	THERMAL METRIC <sup>(1)</sup>	DTQ (X2SON)	UNIT
		6 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	294.4	°C/W
R <sub>0JC(top)</sub>	Junction-to-case (top) thermal resistance	188.9	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	216.8	°C/W
$\Psi_{JT}$	Junction-to-top characterization parameter	26.5	°C/W
$\Psi_{JB}$	Junction-to-board characterization parameter	216.0	°C/W
R <sub>0JC(bot)</sub>	Junction-to-case (bottom) thermal resistance	n/a	°C/W

(1) For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

### **5.5 Electrical Characteristics**

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CON	DITIONS	MIN TYP <sup>(1)</sup>	MAX	UNIT	
V <sub>IK</sub>	I <sub>I</sub> = -18mA	EN = 0V			-1.2	V	
I <sub>IH</sub>	V <sub>1</sub> = 5V	EN = 0V			5.0	μΑ	
I <sub>CCBA</sub>	$V_{ref_B} = V_{EN} = 5$	5.5V, V <sub>ref_A</sub> = 4.5V,	$I_{O} = 0, V_{I} = V_{CC} \text{ or } GND$	1		μΑ	
C <sub>I(ref_A/B/EN)</sub>	V <sub>I</sub> = 3V or 0			11		pF	
C <sub>io(off)</sub>	$V_{O} = 3V \text{ or } 0,$	V <sub>EN</sub> = 0V		4.0	6.0	pF	
C <sub>io(on)</sub>	V <sub>O</sub> = 3V or 0,	V <sub>EN</sub> = 3V		10.5	12.5	pF	
$V_{IK} \qquad I_{I} = -18mA \qquad EN = 0V \qquad   I_{IH} \qquad V_{I} = 5V \qquad EN = 0V \qquad   I_{CCBA} \qquad V_{ref_B} = V_{EN} = 5.5V, V_{ref_A} = 4.5V, I_{O} = 0 \\ \hline I_{(cref_A/B/EN)} \qquad V_{I} = 3V \text{ or } 0 \qquad   I_{O} = 3V \text{ or } 0, \qquad V_{EN} = 3V \\ \hline V_{O} = 3V \text{ or } 0, \qquad V_{EN} = 3V \\ \hline V_{O} = 3V \text{ or } 0, \qquad V_{EN} = 3V \\ \hline V_{I} = 0, \qquad I_{O} = 64mA \qquad   V_{ref_S} \\ \hline V_{ref_S} \\ \hline V_{I} = 0, \qquad I_{O} = 32mA \qquad   V_{ref_S} \\ \hline V_{I} = 1.8V, \qquad I_{O} = 15mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   I_{O} = 10mA \qquad   V_{ref_S} \\ \hline V_{I} = 0V, \qquad   V_{I} = 00mA \qquad   V_{I} = 0 \\ \hline V_{I} = 0V, \qquad   V_{I} = 0mA \qquad   V_{I} = 0 \\ \hline V_{I} = 0V, \qquad   V_{I} = 0mA \qquad   V_{I} = 0 \\ \hline V_{I} = 0V, \qquad   V_{I} = 0mA \qquad   V_{I} = 0 \\ \hline V_{I} = 0V, \qquad   V_{I} = 0mA \qquad   V_{I} = 0 \\ \hline V_{I} = 0V, \qquad   V_{I} = 0mA \qquad   V_{I} = 0 \\ \hline V_{I} = 0V, \qquad   V_{I} = 0mA \qquad   V_{I} = 0 \\ \hline V_{I} = 0V, \qquad   V_{I} = 0mA \qquad   V_{I} = 0 \\ \hline V_{I} = 0V, \qquad   V_{I} = 0mA \qquad   V_{I} = 0 \\ \hline V_{I} = 0V, \qquad   V_{I} = 0 \\ \hline V_{I} = $			$V_{ref_A}$ = 3.3V; $V_{ref_B}$ = $V_{EN}$ = 5V	8.0			
	V <sub>1</sub> = 0,	V <sub>1</sub> = 0,	I <sub>O</sub> = 64mA	$V_{ref\_A}$ = 1.8V; $V_{ref\_B}$ = $V_{EN}$ = 5V	9.0		Ω
	$V_{ref\_A}$ = 1.0V; $V_{ref\_B}$ = $V_{EN}$ = 5V	10					
		L = 20mA	$V_{ref_A}$ = 1.8V; $V_{ref_B}$ = $V_{EN}$ = 5V	10		Ω	
r <sub>ON</sub> <sup>(2)</sup>	v <sub>1</sub> = 0,	1 <sub>0</sub> – 3211A	$V_{ref_A}$ = 2.5V; $V_{ref_B}$ = $V_{EN}$ = 5V	15		12	
	V <sub>I</sub> = 1.8V,	I <sub>O</sub> = 15mA	$V_{ref_A}$ = 3.3V; $V_{ref_B}$ = $V_{EN}$ = 5V	9.0		Ω	
	V <sub>I</sub> = 1.0V,	I <sub>O</sub> = 10mA	$V_{ref_A}$ = 1.8V; $V_{ref_B}$ = $V_{EN}$ = 3.3V	18		Ω	
	V <sub>1</sub> = 0V,	I <sub>O</sub> = 10mA	$V_{ref_A}$ = 1.0V; $V_{ref_B}$ = $V_{EN}$ = 3.3V	20		Ω	
Cio(off) Cio(on)	V <sub>1</sub> = 0V,	I <sub>O</sub> = 10mA	$V_{ref_A}$ = 1.0V; $V_{ref_B}$ = $V_{EN}$ = 1.8V	30		Ω	

(1) All typical values are at  $T_A = 25^{\circ}C$ .

(2) Measured by the voltage drop between the A and B pins at the indicated current through the switch. Minimum ON-state resistance is determined by the lowest voltage of the two (A or B) pins.



## 5.6 Switching Characteristics (Translating Down): $B_N = 3.3V$

over recommended operating free-air temperature range,  $B_N = 3.3V$ ,  $B_N = VIH = A_N + 1$ , VIL = 0, and  $VM = 0.5A_N$  (unless otherwise noted) (see *Parameter Measurement Information*)

	PARAMETER	TEST CONDITION	S	MIN	TYP	MAX	UNIT
			C <sub>L</sub> = 15pF		0.3		
T <sub>PLH</sub>	Low-to-high propagation delay		C <sub>L</sub> = 30pF		0.7		ns
		From (input) A or B to (output) B or	C <sub>L</sub> = 50pF		1.1		
T <sub>PHL</sub> High to low propagation c		A	C <sub>L</sub> = 15pF		0.4		
	High to low propagation delay		C <sub>L</sub> = 30pF		0.8		ns
			C <sub>L</sub> = 50pF		1.2		

### 5.7 Switching Characteristics (Translating Down): $B_N = 2.5V$

over recommended operating free-air temperature range,  $B_N = 2.5V$ ,  $B_N = VIH = A_N + 1$ , VIL = 0, and  $VM = 0.5A_N$  (unless otherwise noted) (see *Parameter Measurement Information*)

	PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
			C <sub>L</sub> = 15pF		0.35		
T <sub>PLH</sub>	Low-to-high propagation delay		C <sub>L</sub> = 30pF		0.8		ns
		From (input) A or B to (output) B or A	C <sub>L</sub> = 50pF		1.2		
	T <sub>PHL</sub> High to low propagation delay		C <sub>L</sub> = 15pF		0.5		
T <sub>PHL</sub>			C <sub>L</sub> = 30pF		1		ns
			C <sub>L</sub> = 50pF		1.3		

#### 5.8 Switching Characteristics (Translating Up): $B_N = 3.3V$

over recommended operating free-air temperature range,  $B_N = 3.3V$ ,  $B_N = VT = A_N + 1$ ,  $Vref_A = VIH$ , VIL = 0,  $VM = 0.5A_N$  and RL = 300 (unless otherwise noted) (see *Parameter Measurement Information*)

	PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT	
			C <sub>L</sub> = 15pF		0.4			
T <sub>PLH</sub>	Low-to-high propagation delay		C <sub>L</sub> = 30pF		0.8		ns	
		From (input) A or B to (output) B or A	C <sub>L</sub> = 50pF		1			
	T <sub>PHL</sub> High to low propagation delay		C <sub>L</sub> = 15pF		0.4			
T <sub>PHL</sub>			C <sub>L</sub> = 30pF		0.9		ns	
			C <sub>L</sub> = 50pF		1			

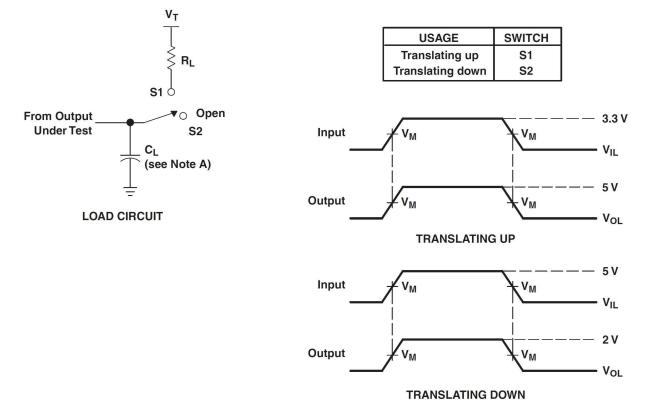
### 5.9 Switching Characteristics (Translating Up):B<sub>N</sub> = 2.5V

over recommended operating free-air temperature range,  $B_N = 2.5V$ ,  $B_N = VT = A_N + 1$ ,  $A_N = VIH$ , VIL = 0,  $VM = 0.5A_N$  and RL = 300 (unless otherwise noted) (see *Parameter Measurement Information*)

	PARAMETER	TEST CONDITION	MIN	TYP	MAX	UNIT	
			C <sub>L</sub> = 15pF		0.45		
T <sub>PLH</sub> Low-to-high	Low-to-high propagation delay		C <sub>L</sub> = 30pF		0.9		ns
		From (input) A or B to (output) B or A	C <sub>L</sub> = 50pF		1.1		
	T <sub>PHL</sub> High to low propagation delay		C <sub>L</sub> = 15pF		0.6		
T <sub>PHL</sub>			C <sub>L</sub> = 30pF		1.1		ns
			C <sub>L</sub> = 50pF		1.3		



## **6** Parameter Measurement Information



- A.  $C_L$  includes probe and jig capacitance.
- B. Generators that have the following characteristics generate all input pulses: PRR  $\leq$  10MHs, Z<sub>0</sub> = 50 $\Omega$ , t<sub>r</sub>  $\leq$  2ns, t<sub>f</sub>  $\leq$  2ns.
- C. The outputs are measured one at a time, with one transition per measurement.

#### Figure 6-1. Load Circuit for Outputs

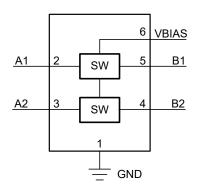


## 7 Detailed Description

### 7.1 Overview

The LSF0002 can be used in level-translation applications for interfacing devices or systems operating with one another that operate at different interface voltages. The LSF family is an excellent choice for use in applications where an open-drain driver is connected to the data I/Os. With appropriate pull-up resistors and layout, LSF can achieve 100MHz. The LSF family can also be used in applications where a push-pull driver is connected to the data I/Os. For an overview of device setup and operation, see *The Logic Minute* training series on *Understanding the LSF Family of Bidirectional, Multi-Voltage Level Translators*.

#### 7.2 Functional Block Diagram



### 7.3 Feature Description

#### 7.3.1 Auto Bidirectional Voltage Translation

The LSF family is an auto bidirectional voltage level translator that is operational from 0.95 to 5.5V. This allows for bidirectional voltage translation between 0.95V and 5.5V without the need for a direction pin in open-drain or push-pull applications. The LSF family supports level translation applications with transmission speeds greater than 100Mbps for open-drain systems using a 30pF capacitance and  $250\Omega$  pullup resistor. Both the output driver of the controller and the peripheral device output can be push-pull or open-drain (pull-up resistors may be required). In both up and down translation, the B-side is often referred to as the high side and refers to devices connected to the B ports. The A-side can be referred to as the low side.



#### 7.3.2 V<sub>BIAS</sub>/ Enable

To enable the I/O pins, the  $V_{BIAS}$  input should be referenced towards the lower power supply (in the following example,  $V_{EXT,A}$ ) during voltage translation. To be in the high impedance state during power-up, power-down, or during operation, the  $V_{BIAS}$  pin must be pulled low and at GND or disabled by an open-drain driver without a pullup resistor. Use the  $V_{BIAS}$  pin to properly bias the I/O channels. A filter capacitor on  $V_{BIAS}$  is also recommended for a stable supply at the device.

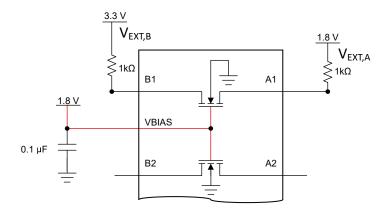


Figure 7-1. V<sub>BIAS</sub> Tied to Lower Power Supply

The supply voltage of open drain I/O devices can be completely different from the supplies used for the LSF and has no impact on the operation. For additional details on how to use the enable pin, see the *Using the Enable Pin with the LSF Family video*.

Table 7-1. Enable Pin Function Table
--------------------------------------

	Data Port State				
Tied directly to V <sub>BIAS</sub>	An = Bn				
L	Hi-Z				

#### 7.4 Device Functional Modes

For each channel (n), when either the An or Bn port is LOW, the switch provides a low impedance path between the An and Bn ports; the corresponding Bn or An port will be pulled LOW. The low R<sub>ON</sub> of the switch allows connections to be made with minimal propagation delay and signal distortion.

Table 7-1 provides a summary of device operation. For additional details on the functional operation of the LSF family of devices, see the *Down Translation with the LSF Family* and *Up Translation with the LSF Family* videos.

Signal Direction <sup>(1)</sup>	Input State	Switch State	Functionality
B to A (Down Translation)	B = LOW	ON (Low Impedance)	A-side voltage is pulled low through the switch to the B-side voltage
B to A (Down Translation)	B = HIGH	OFF (High Impedance)	A-side voltage is clamped at $V_{EXT,A}$ <sup>(2)</sup>
A to B (Up Translation)	A = LOW	ON (Low Impedance)	B-side voltage is pulled low through the switch to the A-side voltage
	A = HIGH	OFF (High Impedance)	B-side voltage is clamped at $V_{\text{EXT,A}}$ and then pulled up to the $V_{\text{EXT,B}}$ supply voltage

#### Table 7-2. Device Functionality

(1) The downstream channel should not be actively driven through a low impedance driver, or else bus contention may occur.

(2) The A-side can have a pullup to V<sub>EXT,A</sub> for additional current drive capability or may also be pulled above V<sub>EXT,A</sub> with a pullup resistor. Specifications in the *Recommended Operating Conditions* section should always be followed.



## 8 Application and Implementation

#### Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

#### 8.1 Application Information

The LSF devices can perform voltage translation for open-drain or push-pull interfaces. Table 8-1 provides common interfaces and the corresponding device recommendation from the LSF family, which supports the corresponding bit count.

Table 8-1. Vo	Itage Translato	for Common	Interfaces
---------------	-----------------	------------	------------

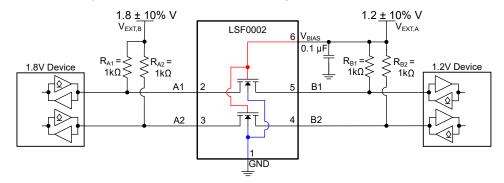
Part Name	Channel Number	Interface
LSF0002	2	GPIO, MDIO, SMBus, PMBus, and I <sup>2</sup> C

Some important reminders regarding the LSF0002 are as follows:

- LSF devices are switch-based, not buffer-based (for more information, see the TXB family for buffer-based devices).
- Specific data rates cannot be calculated by using 1/Tpd.
- V<sub>EXTA</sub>/V<sub>EXTB</sub> are referenced to the external power supplies at the I/Os.
- V<sub>BIAS</sub> should be tied towards the lowest voltage being translated to at the I/Os.

#### 8.2 Typical Applications

#### 8.2.1 Open-Drain Interface (I<sup>2</sup>C, PMBus, SMBus, and GPIO)







#### 8.2.1.1 Design Requirements

#### 8.2.1.1.1 Enable and Disable Guidelines

In the previous figure,  $V_{BIAS}$  is referenced to the lower power supply ( $V_{EXT,A}$ ) of 1.8V power supply and  $V_{EXT,B}$  is set to 3.3V. The A1 and A2 channels have a maximum output voltage equal to  $V_{EXT,A}$  and the B1 and B2 channels have a maximum output voltage equal to  $V_{EXT,B}$ .

The LSF0002 has an  $V_{BIAS}$  input that is used to disable the device by driving this node LOW to GND, placing all I/Os in the high-impedance state. Since the LSF family of devices are switch-type voltage translators, the power consumption is very low. TI recommends always enabling the LSF family for bidirectional applications (I<sup>2</sup>C, SMBus, PMBus, or MDIO).

	PARAMETER	MIM	N TYP	MAX	UNIT
V <sub>EXT,A</sub> <sup>(1)</sup>	reference voltage (A)	0.9	9	5.5	V
V <sub>EXT,B</sub>	reference voltage (B)	V <sub>EXT,A</sub> + 0.8	3	5.5	V
V <sub>BIAS</sub>	input voltage on V <sub>BIAS</sub> pin	V <sub>EXT,</sub>	4	V <sub>EXT,A</sub> + 0.8	V
V <sub>PU</sub>	pull-up supply voltage		0	V <sub>EXT,B</sub>	V

#### Table 8-2. Application Operating Condition

(1) V<sub>EXT.A</sub> is required to be the lowest voltage level across all inputs and outputs.

Note	
The 200k $\Omega$ , bias resistor is not required for the LSF0002	<u>)</u> .

#### 8.2.1.2 Detailed Design Procedure

#### 8.2.1.2.1 Bidirectional Translation

For the bidirectional translation configuration (higher voltage to lower voltage or lower voltage to higher voltage), the  $V_{BIAS}$  input must be connected to  $V_{EXT,A}$ . This allows  $V_{BIAS}$  to regulate the bias of the I/O channels for proper translation. A filter capacitor on  $V_{BIAS}$  is recommended for a stable supply at the device. The controller output driver can be push-pull or open-drain (pull-up resistors may be required) and the peripheral device output can be push-pull or open-drain (pull-up resistors are required to pull the Bn outputs to  $V_{PU}$ ).

#### Note

If either output is push-pull, data must be unidirectional or the outputs must be tri-state and be controlled by some direction-control mechanism to prevent HIGH-to-LOW bus contention in either direction. If both outputs are open-drain, no direction control is needed.

#### 8.2.1.2.2 Pull-Up Resistor Sizing

The pull-up resistor value needs to limit the current through the pass transistor when it is in the ON state to about 15mA. Doing this causes a voltage drop of 260mV to 350mV to have a valid LOW signal on the downstream channel. If the current through the pass transistor is higher than 15mA, then the voltage drop is also higher in the ON state. To set the current through each pass transistor at 15mA, calculate the pull-up resistor value using the following equation:

$$Rpu = \frac{(Vpu - 0.35 V)}{0.015 A}$$
(1)

Table 8-3 provides resistor values, reference voltages, and currents at 8mA, 5mA, and 3mA. The resistor value shown in the +10% column (or a larger value) should be used so that the voltage drop across the transistor is 350mV or less. The external driver must be able to sink the total current from the resistors on both sides of the LSF family device at 0.175V, although the 15mA applies only to current flowing through the LSF family device. The device driving the low state at 0.175V must sink current from one or more of the pull-up resistors and maintain  $V_{OL}$ . A decrease in resistance will increase current, and thus result in increased  $V_{OL}$ .



V <sub>PU</sub> <sup>(1) (2)</sup>	8m	ıA	5n	nA	3n	۱A				
VPU (V)(=)	NOMINAL (Ω)	+10% <sup>(3)</sup> (Ω)	NOMINAL (Ω)	+10% <sup>(3)</sup> (Ω)	NOMINAL (Ω)	+10% <sup>(3)</sup> (Ω)				
5V	581	639	930	1023	1550	1705				
3.3V	369	406	590	649	983	1082				
2.5V	269	296	430	473	717	788				
1.8V	181	199	290	319	483	532				
1.5V	144	158	230	253	383	422				
1.2V	106	117	170	187	283	312				

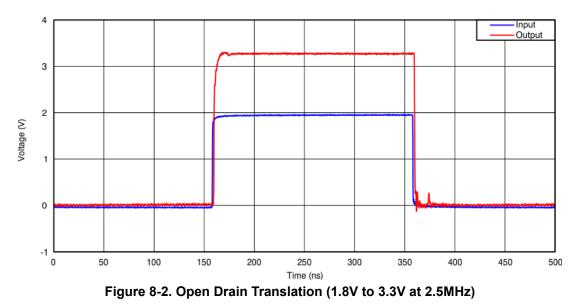
#### Table 8-3. Pull-Up Resistor Values

(1) Calculated for V<sub>OL</sub> = 0.35V

(2) Assumes output driver  $V_{OL} = 0.175V$  at stated current

(3) +10% to compensate for  $V_{DD}$  range and resistor tolerance

#### 8.2.1.3 Application Curve



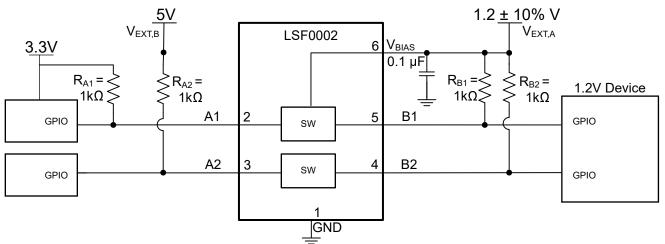
#### 8.2.2 Mixed-Mode Voltage Translation

The supply voltage ( $V_{EXT A,B}$ ) for each channel can be individually set with a pull-up resistor. Figure 8-3 shows an example of this mixed-mode multi-voltage translation. For additional details on multi-voltage translation, see the *Multi-voltage Translation with the LSF Family* video.

With  $V_{EXT,B}$  pulled up to 5V and  $V_{EXT,A}$  connected to 1.2V, all channels will be clamped to 1.2V at which point a pullup can be used to define the high level voltage for a given channel.

- **Push-Pull Down Translation (5V to 1.2V):** Channel 1 is an example of this setup. When A2 is 5V, B1 is clamped to 1.2V, and when A1 is LOW, B1 is driven LOW through the switch.
- **Push-Pull Up Translation (1.2V to 3.3V)**: Channel 2 is an example of this setup. When B2 is 1.2V, the switch is high impedance and the A2 channel is pulled up to 3.3V. When B2 is LOW, A2 is driven LOW through the switch.
- **Push-Pull Down Translation (3.3V to 1.2V):** Channels 2 is an example of this setup. When A2 is driven to 3.3V, B2 is clamped to 1.2V, and when A2 is LOW, B2 is driven LOW through the switch.







## 8.2.3 Voltage Translation for $V_{ref_B} < V_{ref_A} + 0.8V$

As described in the *Enable and Disable Guidelines* section, it is generally recommended that  $V_{EXT,B} > V_{EXT,A} + 0.8V$ ; however, the device can still operate in the condition where  $V_{EXT,B} < V_{EXT,A} + 0.8V$  as long as additional considerations are made for the design.

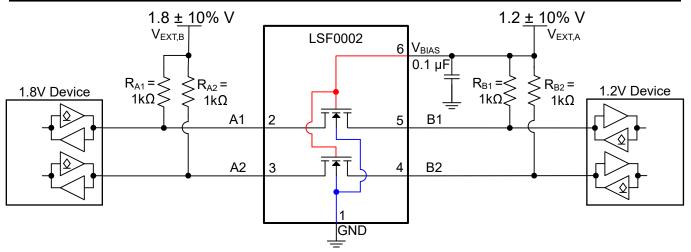
**Typical Operation** ( $V_{EXT,B} > V_{EXT,A} + 0.8V$ ): in this scenario, pullup resistors are not required on the A-side for proper down-translation. When down translating from B to A, the A-side I/O ports will clamp at  $V_{EXT,A}$  to provide proper voltage translation. For further explanation of device operation, see the *Down Translation with the LSF Family* video.

**Requirements for V<sub>EXT,B</sub> < V<sub>EXT,A</sub> + 0.8V Operation:** in this scenario, there is not a large enough voltage difference between V<sub>EXT,A</sub> and V<sub>EXT,B</sub> so that the A side I/O ports will be clamped at V<sub>EXT,A</sub>, but rather at a voltage approximately equal to V<sub>EXT,B</sub> – 0.8V. For example, if V<sub>EXT,B</sub> = 1.8V and V<sub>EXT,A</sub> = 1.2V, the A-side I/Os will clamp to a voltage around 1.0V. Therefore, to operate in such a condition, the following additional design considerations must be met:

- $V_{EXT,B}$  must be greater than  $V_{EXT,A}$  during operation ( $V_{EXT,B} > V_{EXT,A}$ )
- Pullup resistors should be populated on A-side I/O ports for the line to be fully pulled up to the desired voltage.

Figure 8-4 shows an example of this setup, where  $1.2V \leftrightarrow 1.8V$  translation is achieved with the LSF0002. This type of setup also applies for other voltage nodes such as  $1.8V \leftrightarrow 2.5V$ ,  $1.05V \leftrightarrow 1.5V$ , and others as long as the *Recommended Operating Conditions* table is followed.





#### Figure 8-4. 1.2V to 1.8V Level Translation with LSF0002

#### 8.3 Power Supply Recommendations

There are no power sequence requirements for the LSF family. Table 8-4 provides recommended operating voltages for all supply and input pins.

	PARAMETER	MIN	TYP	MAX	UNIT				
V <sub>EXT,A</sub> <sup>(1)</sup>	reference voltage (A)	0.9		5.5	V				
V <sub>EXT,B</sub>	reference voltage (B)	V <sub>EXT,A</sub> + 0.8		5.5	V				
V <sub>BIAS</sub>	input voltage on EN pin	V <sub>EXT,A</sub>		$V_{EXT,A}$ + 0.8	V				
V <sub>PU</sub>	pull-up supply voltage	0		V <sub>EXT,B</sub>	V				

#### Table 8-4. Recommended Operating Voltages

(1)  $V_{EXT,A}$  is required to be the lowest voltage level across all inputs and outputs.

#### 8.4 Layout

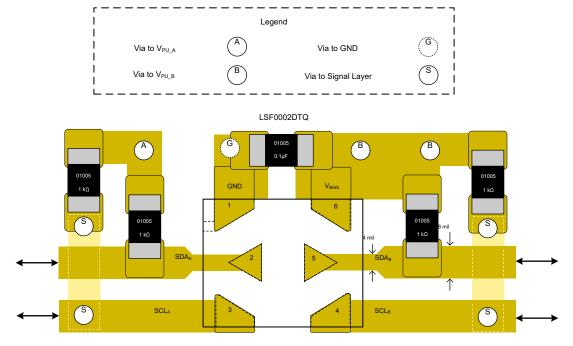
#### 8.4.1 Layout Guidelines

Because the LSF family is a switch-type level translator, the signal integrity is highly related with a pull-up resistor and PCB capacitance condition. Therefore, do as follows:

- Short the signal trace as short as possible to reduce capacitance and minimize stub from the pull-up resistor.
- Place the LSF device as close to the high voltage side as possible.
- Select the appropriate pull-up resistor that applies to translation levels and the driving capability of the transmitter.



#### 8.4.2 Layout Example



\*Pullup Resistors are not required to be adjacent to the device, and can be present on the I<sup>2</sup>C line elsewhere in the system

#### Figure 8-5. Short Trace Layout



Figure 8-6. Device Placement



## 9 Device and Documentation Support

#### 9.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, LSF Translator Family Evaluation Module user's guide
- Texas Instruments, Biasing Requirements for TXS, TXB, and LSF Auto-Bidirectional Translators application note
- Texas Instruments, Voltage Level Translation with the LSF Family application note
- The Logic Minute Video Training Series on Understanding the LSF Family of Devices:
  - Texas Instruments, Introduction Voltage Level Translation with the LSF Family
  - Texas Instruments, Understanding the Bias Circuit for the LSF Family
  - Texas Instruments, Using the Enable Pin with the LSF Family
  - Texas Instruments, Translation Basics with the LSF Family
  - Texas Instruments, Down Translation with the LSF Family
  - Texas Instruments, Up Translation with the LSF Family
  - Texas Instruments, Multi-Voltage Translation with the LSF Family
  - Texas Instruments, Single Supply Translation with the LSF Family

#### 9.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

#### **9.3 Support Resources**

TI E2E<sup>™</sup> support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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#### 9.4 Trademarks

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#### 9.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### 9.6 Glossary

TI Glossary This glossary lists and explains terms, acronyms, and definitions.

#### **10 Revision History**

DATE	REVISION	NOTES
April 2024	*	Initial Release

### 11 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



## PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LSF0002DTQR	ACTIVE	X2SON	DTQ	6	12000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	J	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

<sup>(3)</sup> MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

<sup>(5)</sup> Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

<sup>(6)</sup> Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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## TAPE AND REEL INFORMATION





#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LSF0002DTQR	X2SON	DTQ	6	12000	180.0	8.4	0.92	1.12	0.47	4.0	8.1	Q2



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## PACKAGE MATERIALS INFORMATION

25-Sep-2024



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LSF0002DTQR	X2SON	DTQ	6	12000	182.0	182.0	20.0

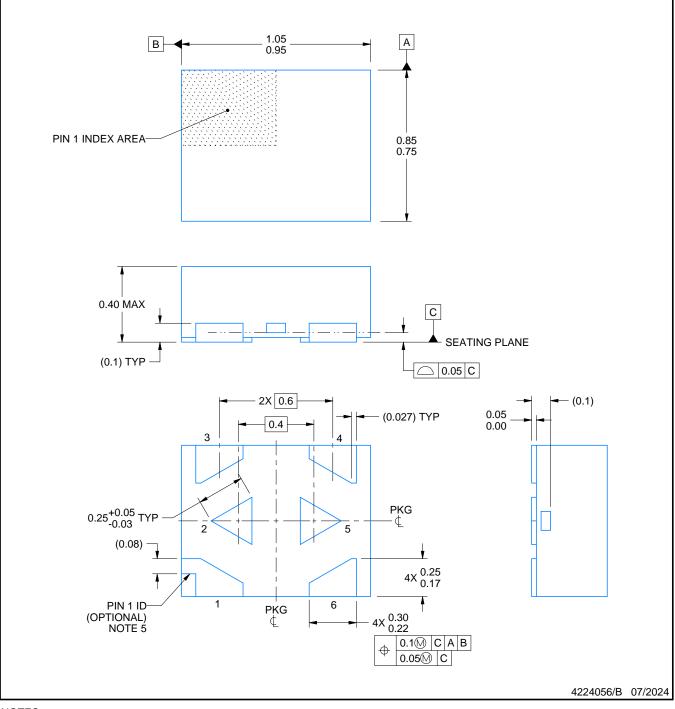
# **DTQ0006A**



# **PACKAGE OUTLINE**

## X2SON - 0.4 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- The package thermal pads must be soldered to the printed circuit board for optimal thermal and mechanical performance.
   The size and shape of this feature may vary.
- 5. Features may not exist. Recommend use of pin 1 marking on top of package for orientation purposes.

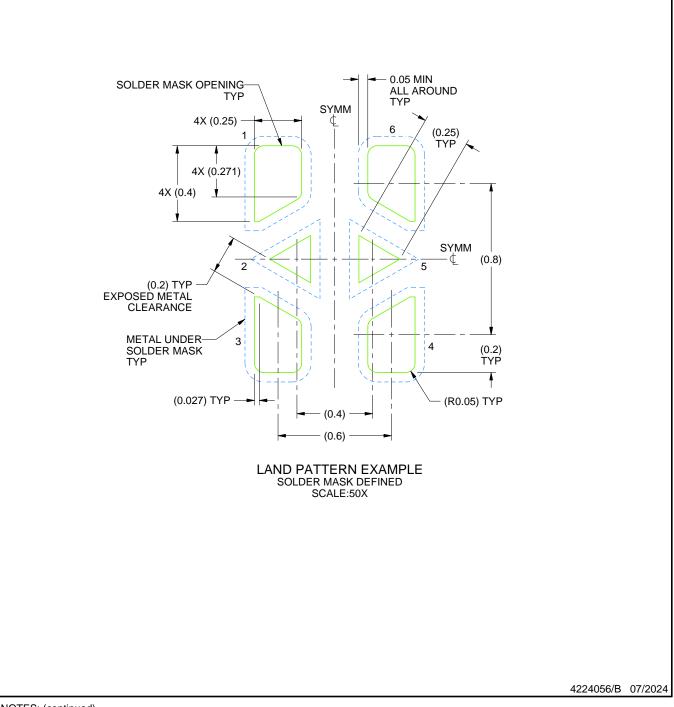


# **DTQ0006A**

# **EXAMPLE BOARD LAYOUT**

## X2SON - 0.4 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



NOTES: (continued)

6. This package is designed to be soldered to a thermal pads on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).

7. Vias are optional depending on application, refer to device data sheet. If some or all are implemented, recommended via locations are shown.

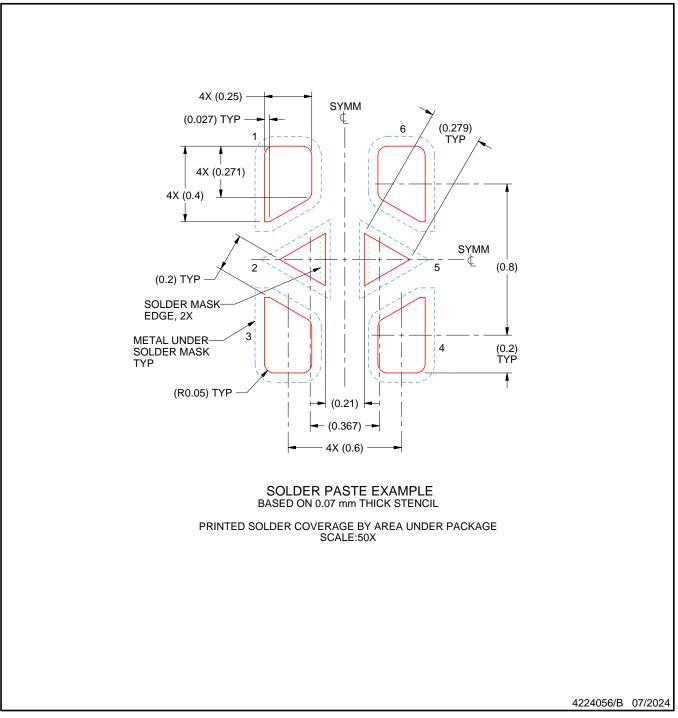


# DTQ0006A

# **EXAMPLE STENCIL DESIGN**

## X2SON - 0.4 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



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