

LMH6642Q/LMH6643Q Low Power, 130 MHz, 75 mA Rail-to-Rail Output Amplifiers

1 Features

- ($V_S = \pm 5\text{ V}$, $T_A = 25^\circ\text{C}$, $R_L = 2\text{ k}\Omega$, $A_V = +1$. Typical Values Unless Specified).
- -3dB BW ($A_V = +1$) 130 MHz
- Supply Voltage Range 2.7 V to 10 V
- Slew Rate, ($A_V = -1$) $130\text{V}/\mu\text{s}$ ⁽¹⁾
- Supply Current (no load) 2.7 mA/amp
- Output Short Circuit Current $+115\text{ mA}/-145\text{ mA}$
- Linear Output Current $\pm 75\text{ mA}$
- Input Common Mode Voltage 0.5V Beyond V^- , 1V from V^+
- Output Voltage Swing 40mV from Rails
- Input Voltage Noise (100kHz) $17\text{ nV}/\sqrt{\text{Hz}}$
- Input Current Noise (100kHz) $0.9\text{ pA}/\sqrt{\text{Hz}}$
- THD (5MHz, $R_L = 2\text{ k}\Omega$, $V_O = 2V_{PP}$, $A_V = +2$) -62 dBc
- Settling Time 68ns
- Fully Characterized for 3 V, 5 V, and $\pm 5\text{ V}$
- Overdrive Recovery 100ns
- Output Short Circuit Protected⁽²⁾
- No Output Phase Reversal with CMVR Exceeded
- LMH6643QMM and LMH6642QMF are AEC-Q100 Grade 3 Qualified and are Manufactured on an Automotive Grade Flow

⁽¹⁾ Slew rate is the average of the rising and falling slew rates

⁽²⁾ Output short circuit duration is infinite for $V_S < 6\text{ V}$ at room temperature and below. For $V_S > 6\text{ V}$, allowable short circuit duration is 1.5 ms.

2 Applications

- Active Filters
- CD/DVD ROM
- ADC Buffer Amp
- Portable Video
- Current Sense Buffer
- Automotive

3 Description

The LMH664X family true single supply voltage feedback amplifiers offer high speed (130 MHz), low distortion (-62 dBc), and exceptionally high output current (approximately 75 mA) at low cost and with reduced power consumption when compared against existing devices with similar performance.

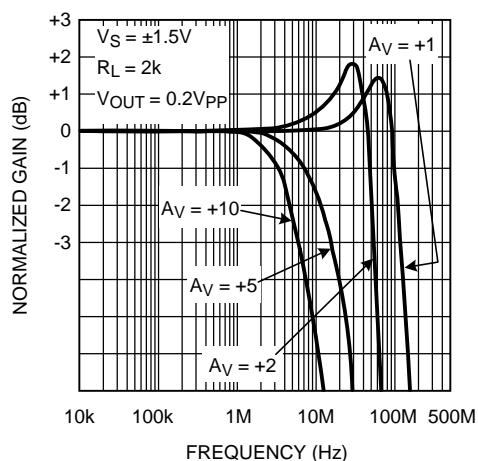
Input common mode voltage range extends to 0.5 V below V^- and 1 V from V^+ . Output voltage range extends to within 40mV of either supply rail, allowing wide dynamic range especially desirable in low voltage applications. The output stage is capable of approximately 75 mA in order to drive heavy loads. Fast output Slew Rate ($130\text{ V}/\mu\text{s}$) ensures large peak-to-peak output swings can be maintained even at higher speeds, resulting in exceptional full power bandwidth of 40 MHz with a 3-V supply. These characteristics, along with low cost, are ideal features for a multitude of industrial and commercial applications.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)
LMH6642-Q1	SOT-23 (5)	2.90 mm x 1.60 mm
LMH6643-Q1	VSSOP (8)	3.00 mm x 3.00 mm

⁽¹⁾ For all available packages, see the orderable addendum at the end of the datasheet.

Closed Loop Gain vs. Frequency for Various Gain



Large Signal Frequency Response

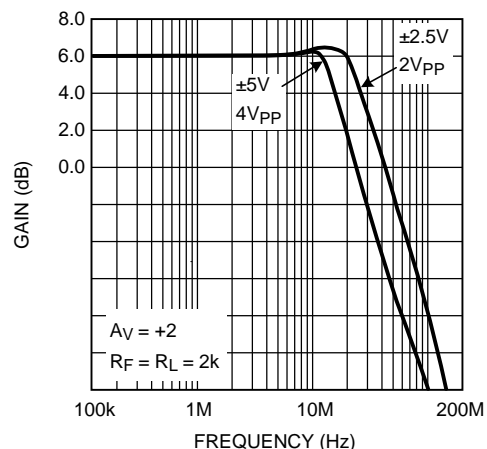


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4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision B (March 2013) to Revision C	Page
• Changed data sheet structure and organization. Added, updated, or renamed the following sections: Device Information Table, Pin Configuration and Functions, Application and Implementation; Device and Documentation Support; Mechanical, Packaging, and Ordering Information	1
• Changed "Junction Temperature Range" to "Operating Temperature Range" in Recommended Operating Conditions.....	4
• Deleted $T_J = 25^\circ\text{C}$ in Electrical Characteristics tables.....	5
• Deleted $T_J = 25^\circ\text{C}$ in Typical Performance Characteristics section.	11

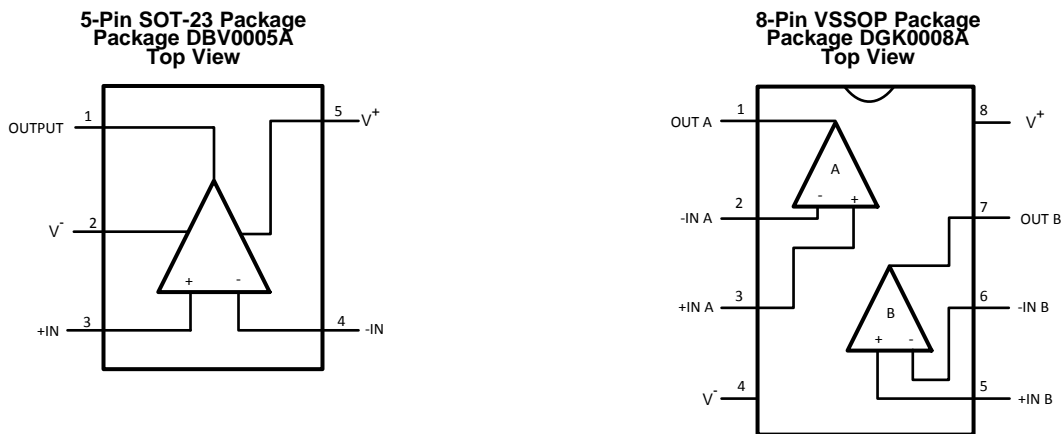
Changes from Revision A (March 2013) to Revision B	Page
• Changed layout of National Data Sheet to TI format	24

5 Description (continued)

Careful attention has been paid to ensure device stability under all operating voltages and modes. The result is a very well behaved frequency response characteristic (0.1dB gain flatness up the 12MHz under 150Ω load and $A_V = +2$) with minimal peaking (typically 2dB maximum) for any gain setting and under both heavy and light loads. This along with fast settling time (68ns) and low distortion allows the device to operate well in ADC buffer, and high frequency filter applications as well as other applications.

This device family offers professional quality video performance with low DG (0.01%) and DP (0.01°) characteristics. Differential Gain and Differential Phase characteristics are also well maintained under heavy loads (150Ω) and throughout the output voltage range. The LMH664X family is offered in single (LMH6642) and dual (LMH6643).

6 Pin Configuration and Functions



Pin Functions

NAME	PIN		I/O	DESCRIPTION
	LMH6642Q	LMH6643Q		
-IN	4		I	Inverting Input
+IN	3		I	Non-inverting Input
-IN A		2	I	ChA Inverting Input
+IN A		3	I	ChA Non-inverting Input
-IN B		6	I	ChB Inverting Input
+IN B		5	I	ChB Non-inverting Input
OUT A		1	O	ChA Output
OUT B		7	O	ChB Output
OUTPUT	1		O	Output
V ⁻	2	4	I	Negative Supply
V ⁺	5	8	I	Positive Supply

7 Specifications

7.1 Absolute Maximum Ratings⁽¹⁾⁽²⁾⁽¹⁾

over operating free-air temperature range (unless otherwise noted)

	MIN	MAX	UNIT
V _{IN} Differential		±2.5	V
Output Short Circuit Duration		See ⁽³⁾ and ⁽⁴⁾	
Supply Voltage (V ⁺ - V ⁻)		13.5	V
Voltage at Input/Output pins		V ⁺ +0.8 V ⁻ -0.8	V
Input Current		±10	mA
Junction Temperature ⁽⁵⁾		+150	°C
Soldering Information			
Infrared or Convection Reflow (20 sec)		235	°C
Wave Soldering Lead Temp.(10 sec)		260	°C

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not ensured. For ensured specifications and the test conditions, see the Electrical Characteristics.
- (2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.
- (3) Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150°C.
- (4) Output short circuit duration is infinite for V_S < 6 V at room temperature and below. For V_S > 6 V, allowable short circuit duration is 1.5ms.
- (5) The maximum power dissipation is a function of T_{J(MAX)}, R_{θJA}, and T_A. The maximum allowable power dissipation at any ambient temperature is P_D = (T_{J(MAX)} - T_A)/R_{θJA}. All numbers apply for packages soldered directly onto a PC board.

7.2 Handling Ratings

		MIN	MAX	UNIT
T _{stg}	Storage temperature range	-65	+150	°C
V _(ESD)	Electrostatic discharge	Human body model (HBM), per AEC Q100-002 ⁽¹⁾	2000	V
		Machine Model (MM) ⁽²⁾	200	
		Charged Device Model (CDM), per AEC Q100-011	1000	

- (1) AEC Q100-002 indicates HBM stressing is done in accordance with the ANSI/ESDA/JEDEC JS-001 specification, 1.5kΩ in series with 100pF.
- (2) Machine Model, 0Ω in series with 200pF.

7.3 Recommended Operating Conditions⁽¹⁾

over operating free-air temperature range (unless otherwise noted)

	MIN	MAX	UNIT
Supply Voltage (V ⁺ - V ⁻)	2.7	10	V
Operating Temperature Range ⁽²⁾	-40	+85	°C

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not ensured. For ensured specifications and the test conditions, see the Electrical Characteristics.
- (2) The maximum power dissipation is a function of T_{J(MAX)}, R_{θJA}, and T_A. The maximum allowable power dissipation at any ambient temperature is P_D = (T_{J(MAX)} - T_A)/R_{θJA}. All numbers apply for packages soldered directly onto a PC board.

7.4 Thermal Information

THERMAL METRIC ⁽¹⁾	DBV05A	DGK08A	UNIT
	5 PINS	8 PINS	
R _{θJA} Junction-to-ambient thermal resistance ⁽²⁾	265°C/W	235°C/W	°C/W

- (1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](http://www.ti.com/lit/zip/Spra953).
- (2) The maximum power dissipation is a function of T_{J(MAX)}, R_{θJA}, and T_A. The maximum allowable power dissipation at any ambient temperature is P_D = (T_{J(MAX)} - T_A)/R_{θJA}. All numbers apply for packages soldered directly onto a PC board.

7.5 3V Electrical Characteristics

Unless otherwise specified, all limits ensured for $V^+ = 3V$, $V^- = 0V$, $V_{CM} = V_O = V^+/2$, V_{ID} (input differential voltage) as noted (where applicable) and $R_L = 2k\Omega$ to $V^+/2$. **Boldface** limits apply at the temperature extremes.

PARAMETER		TEST CONDITIONS	MIN ⁽¹⁾	TYP ⁽²⁾	MAX ⁽¹⁾	UNIT
BW	–3dB BW	$A_V = +1$, $V_{OUT} = 200mV_{PP}$	80	115		MHz
		$A_V = +2$, -1 , $V_{OUT} = 200mV_{PP}$		46		
BW _{0.1dB}	0.1dB Gain Flatness	$A_V = +2$, $R_L = 150\Omega$ to $V^+/2$, $R_L = 402\Omega$, $V_{OUT} = 200mV_{PP}$		19		MHz
PBW	Full Power Bandwidth	$A_V = +1$, $-1dB$, $V_{OUT} = 1V_{PP}$		40		MHz
e_n	Input-Referred Voltage Noise	$f = 100kHz$		17		nV/\sqrt{Hz}
		$f = 1kHz$		48		
i_n	Input-Referred Current Noise	$f = 100kHz$		0.90		pA/\sqrt{Hz}
		$f = 1kHz$		3.3		
THD	Total Harmonic Distortion	$f = 5MHz$, $V_O = 2V_{PP}$, $A_V = -1$, $R_L = 100\Omega$ to $V^+/2$		–48		dBc
DG	Differential Gain	$V_{CM} = 1V$, NTSC, $A_V = +2$ $R_L = 150\Omega$ to $V^+/2$		0.17%		
		$R_L = 1k\Omega$ to $V^+/2$		0.03%		
DP	Differential Phase	$V_{CM} = 1V$, NTSC, $A_V = +2$ $R_L = 150\Omega$ to $V^+/2$		0.05		deg
		$R_L = 1k\Omega$ to $V^+/2$		0.03		
CT Rej.	Cross-Talk Rejection	$f = 5MHz$, Receiver: $R_f = R_g = 510\Omega$, $A_V = +2$		47		dB
T_S	Settling Time	$V_O = 2V_{PP}$, $\pm 0.1\%$, 8pF Load, $V_S = 5V$		68		ns
SR	Slew Rate ⁽³⁾	$A_V = -1$, $V_I = 2V_{PP}$	90	120		V/ μs
V_{OS}	Input Offset Voltage	For LMH6642		± 1	± 5 ± 7	mV
		For LMH6643		± 1	± 3.4 ± 7	
TC V_{OS}	Input Offset Average Drift	⁽⁴⁾		± 5		$\mu V/^\circ C$
I_B	Input Bias Current	⁽⁵⁾		–1.50	–2.60 –3.25	μA
I_{OS}	Input Offset Current			20	800 1000	nA
R_{IN}	Common Mode Input Resistance			3		M Ω
C_{IN}	Common Mode Input Capacitance			2		pF
CMVR	Input Common-Mode Voltage Range	CMRR $\geq 50dB$		–0.5	–0.2 –0.1	V
				1.8 1.6	2.0	
CMRR	Common Mode Rejection Ratio	V_{CM} Stepped from 0V to 1.5V	72	95		dB
A_{VOL}	Large Signal Voltage Gain	$V_O = 0.5V$ to $2.5V$ $R_L = 2k\Omega$ to $V^+/2$	80 75	96		dB
		$V_O = 0.5V$ to $2.5V$ $R_L = 150\Omega$ to $V^+/2$	74 70	82		
V_O	Output Swing High	$R_L = 2k\Omega$ to $V^+/2$, $V_{ID} = 200mV$	2.90	2.98		V
		$R_L = 150\Omega$ to $V^+/2$, $V_{ID} = 200mV$	2.80	2.93		
	Output Swing Low	$R_L = 2k\Omega$ to $V^+/2$, $V_{ID} = -200mV$		25	75	mV
		$R_L = 150\Omega$ to $V^+/2$, $V_{ID} = -200mV$		75	150	

(1) All limits are ensured by testing or statistical analysis.

(2) Typical values represent the most likely parametric norm.

(3) Slew rate is the average of the rising and falling slew rates.

(4) Offset voltage average drift determined by dividing the change in V_{OS} at temperature extremes by the total temperature change.

(5) Positive current corresponds to current flowing into the device.

3V Electrical Characteristics (continued)

Unless otherwise specified, all limits ensured for $V^+ = 3V$, $V^- = 0V$, $V_{CM} = V_O = V^+/2$, V_{ID} (input differential voltage) as noted (where applicable) and $R_L = 2k\Omega$ to $V^+/2$. **Boldface** limits apply at the temperature extremes.

PARAMETER		TEST CONDITIONS	MIN ⁽¹⁾	TYP ⁽²⁾	MAX ⁽¹⁾	UNIT
I_{SC}	Output Short Circuit Current	Sourcing to $V^+/2$ $V_{ID} = 200mV$ ⁽⁶⁾	50 35	95		mA
		Sinking to $V^+/2$ $V_{ID} = -200mV$ ⁽⁶⁾	55 40	110		
I_{OUT}	Output Current	$V_{OUT} = 0.5V$ from either supply		± 65		mA
+PSRR	Positive Power Supply Rejection Ratio	$V^+ = 3.0V$ to $3.5V$, $V_{CM} = 1.5V$	75	85		dB
I_S	Supply Current (per channel)	No Load		2.70	4.00 4.50	mA

(6) Short circuit test is a momentary test. Output short circuit duration is infinite for $V_S < 6V$ at room temperature and below. For $V_S > 6V$, allowable short circuit duration is 1.5ms.

7.6 5V Electrical Characteristics

Unless otherwise specified, all limits ensured for $V^+ = 5V$, $V^- = 0V$, $V_{CM} = V_O = V^+/2$, V_{ID} (input differential voltage) as noted (where applicable) and $R_L = 2k\Omega$ to $V^+/2$. **Boldface** limits apply at the temperature extremes.

PARAMETER		TEST CONDITIONS	MIN ⁽¹⁾	TYP ⁽²⁾	MAX ⁽¹⁾	UNIT
BW	–3dB BW	$A_V = +1$, $V_{OUT} = 200mV_{PP}$	90	120		MHz
		$A_V = +2$, -1 , $V_{OUT} = 200mV_{PP}$		46		
BW _{0.1dB}	0.1dB Gain Flatness	$A_V = +2$, $R_L = 150\Omega$ to $V^+/2$, $R_f = 402\Omega$, $V_{OUT} = 200mV_{PP}$		15		MHz
PBW	Full Power Bandwidth	$A_V = +1$, $-1dB$, $V_{OUT} = 2V_{PP}$		22		MHz
e_n	Input-Referred Voltage Noise	$f = 100kHz$		17		nV/\sqrt{Hz}
		$f = 1kHz$		48		
i_n	Input-Referred Current Noise	$f = 100kHz$		0.90		pA/\sqrt{Hz}
		$f = 1kHz$		3.3		
THD	Total Harmonic Distortion	$f = 5MHz$, $V_O = 2V_{PP}$, $A_V = +2$		–60		dBc
DG	Differential Gain	NTSC, $A_V = +2$ $R_L = 150\Omega$ to $V^+/2$		0.16%		
		$R_L = 1k\Omega$ to $V^+/2$		0.05%		
DP	Differential Phase	NTSC, $A_V = +2$ $R_L = 150\Omega$ to $V^+/2$		0.05		deg
		$R_L = 1k\Omega$ to $V^+/2$		0.01		
CT Rej.	Cross-Talk Rejection	$f = 5MHz$, Receiver: $R_f = R_g = 510\Omega$, $A_V = +2$		47		dB
T_S	Settling Time	$V_O = 2V_{PP}$, $\pm 0.1\%$, 8pF Load		68		ns
SR	Slew Rate ⁽³⁾	$A_V = -1$, $V_I = 2V_{PP}$	95	125		V/ μs
V_{OS}	Input Offset Voltage	For LMH6642		± 1	± 5 ± 7	mV
		For LMH6643		± 1	± 3.4 ± 7	
TC V_{OS}	Input Offset Average Drift	⁽⁴⁾		± 5		$\mu V/^\circ C$
I_B	Input Bias Current	⁽⁵⁾		–1.70	–2.60 –3.25	μA
I_{OS}	Input Offset Current			20	800 1000	nA
R_{IN}	Common Mode Input Resistance			3		M Ω
C_{IN}	Common Mode Input Capacitance			2		pF
CMVR	Input Common-Mode Voltage Range	CMRR $\geq 50dB$		–0.5	–0.2 –0.1	V
				3.8 3.6	4.0	
CMRR	Common Mode Rejection Ratio	V_{CM} Stepped from 0V to 3.5V	72	95		dB
A_{VOL}	Large Signal Voltage Gain	$V_O = 0.5V$ to 4.50V $R_L = 2k\Omega$ to $V^+/2$	86 82	98		dB
		$V_O = 0.5V$ to 4.25V $R_L = 150\Omega$ to $V^+/2$	76 72	82		
V_O	Output Swing High	$R_L = 2k\Omega$ to $V^+/2$, $V_{ID} = 200mV$	4.90	4.98		V
		$R_L = 150\Omega$ to $V^+/2$, $V_{ID} = 200mV$	4.65	4.90		
	Output Swing Low	$R_L = 2k\Omega$ to $V^+/2$, $V_{ID} = -200mV$		25	100	mV
		$R_L = 150\Omega$ to $V^+/2$, $V_{ID} = -200mV$		100	150	

(1) All limits are ensured by testing or statistical analysis.

(2) Typical values represent the most likely parametric norm.

(3) Slew rate is the average of the rising and falling slew rates.

(4) Offset voltage average drift determined by dividing the change in V_{OS} at temperature extremes by the total temperature change.

(5) Positive current corresponds to current flowing into the device.

5V Electrical Characteristics (continued)

Unless otherwise specified, all limits ensured for $V^+ = 5V$, $V^- = 0V$, $V_{CM} = V_O = V^+/2$, V_{ID} (input differential voltage) as noted (where applicable) and $R_L = 2k\Omega$ to $V^+/2$. **Boldface** limits apply at the temperature extremes.

PARAMETER		TEST CONDITIONS	MIN ⁽¹⁾	TYP ⁽²⁾	MAX ⁽¹⁾	UNIT
I _{SC}	Output Short Circuit Current	Sourcing to $V^+/2$ $V_{ID} = 200mV$ ⁽⁶⁾	55 40	115		mA
		Sinking to $V^+/2$ $V_{ID} = -200mV$ ⁽⁶⁾	70 55	140		
I _{OUT}	Output Current	$V_O = 0.5V$ from either supply		±70		mA
+PSRR	Positive Power Supply Rejection Ratio	$V^+ = 4.0V$ to $6V$	79	90		dB
I _S	Supply Current (per channel)	No Load		2.70	4.25 5.00	mA

(6) Short circuit test is a momentary test. Output short circuit duration is infinite for $V_S < 6V$ at room temperature and below. For $V_S > 6V$, allowable short circuit duration is 1.5ms.

7.7 ±5V Electrical Characteristics

Unless otherwise specified, all limits ensured for $V^+ = 5V$, $V^- = -5V$, $V_{CM} = V_O = 0V$, V_{ID} (input differential voltage) as noted (where applicable) and $R_L = 2k\Omega$ to ground. **Boldface** limits apply at the temperature extremes.

PARAMETER		TEST CONDITIONS	MIN ⁽¹⁾	TYP ⁽²⁾	MAX ⁽¹⁾	UNIT
BW	–3dB BW	$A_V = +1$, $V_{OUT} = 200mV_{PP}$	95	130		MHz
		$A_V = +2$, -1 , $V_{OUT} = 200mV_{PP}$		46		
BW _{0.1dB}	0.1dB Gain Flatness	$A_V = +2$, $R_L = 150\Omega$ to $V^+/2$, $R_f = 806\Omega$, $V_{OUT} = 200mV_{PP}$		12		MHz
PBW	Full Power Bandwidth	$A_V = +1$, $-1dB$, $V_{OUT} = 2V_{PP}$		24		MHz
e_n	Input-Referred Voltage Noise	$f = 100kHz$		17		nV/\sqrt{Hz}
		$f = 1kHz$		48		
i_n	Input-Referred Current Noise	$f = 100kHz$		0.90		pA/\sqrt{Hz}
		$f = 1kHz$		3.3		
THD	Total Harmonic Distortion	$f = 5MHz$, $V_O = 2V_{PP}$, $A_V = +2$		–62		dBc
DG	Differential Gain	NTSC, $A_V = +2$ $R_L = 150\Omega$ to $V^+/2$		0.15%		
		$R_L = 1k\Omega$ to $V^+/2$		0.01%		
DP	Differential Phase	NTSC, $A_V = +2$ $R_L = 150\Omega$ to $V^+/2$		0.04		deg
		$R_L = 1k\Omega$ to $V^+/2$		0.01		
CT Rej.	Cross-Talk Rejection	$f = 5MHz$, Receiver: $R_f = R_g = 510\Omega$, $A_V = +2$		47		dB
T_S	Settling Time	$V_O = 2V_{PP}$, $\pm 0.1\%$, 8pF Load, $V_S = 5V$		68		ns
SR	Slew Rate ⁽³⁾	$A_V = -1$, $V_I = 2V_{PP}$	100	135		V/ μ s
V_{OS}	Input Offset Voltage	For LMH6642		± 1	± 5 ± 7	mV
		For LMH6643		± 1	± 3.4 ± 7	
TC V_{OS}	Input Offset Average Drift	⁽⁴⁾		± 5		$\mu V/^\circ C$
I_B	Input Bias Current	⁽⁵⁾		–1.60	–2.60 –3.25	μA
I_{OS}	Input Offset Current			20	800 1000	nA
R_{IN}	Common Mode Input Resistance			3		M Ω
C_{IN}	Common Mode Input Capacitance			2		pF
CMVR	Input Common-Mode Voltage Range	CMRR $\geq 50dB$		–5.5	–5.2 –5.1	V
			3.8 3.6	4.0		
CMRR	Common Mode Rejection Ratio	V_{CM} Stepped from $-5V$ to $3.5V$	74	95		dB
A_{VOL}	Large Signal Voltage Gain	$V_O = -4.5V$ to $4.5V$, $R_L = 2k\Omega$	88 84	96		dB
		$V_O = -4.0V$ to $4.0V$, $R_L = 150\Omega$	78 74	82		
V_O	Output Swing High	$R_L = 2k\Omega$, $V_{ID} = 200mV$	4.90	4.96		V
		$R_L = 150\Omega$, $V_{ID} = 200mV$	4.65	4.80		
	Output Swing Low	$R_L = 2k\Omega$, $V_{ID} = -200mV$		–4.96	–4.90	V
		$R_L = 150\Omega$, $V_{ID} = -200mV$		–4.80	–4.65	

(1) All limits are ensured by testing or statistical analysis.

(2) Typical values represent the most likely parametric norm.

(3) Slew rate is the average of the rising and falling slew rates.

(4) Offset voltage average drift determined by dividing the change in V_{OS} at temperature extremes by the total temperature change.

(5) Positive current corresponds to current flowing into the device.

±5V Electrical Characteristics (continued)

Unless otherwise specified, all limits ensured for $V^+ = 5V$, $V^- = -5V$, $V_{CM} = V_O = 0V$, V_{ID} (input differential voltage) as noted (where applicable) and $R_L = 2k\Omega$ to ground. **Boldface** limits apply at the temperature extremes.

PARAMETER		TEST CONDITIONS	MIN ⁽¹⁾	TYP ⁽²⁾	MAX ⁽¹⁾	UNIT
I _{SC}	Output Short Circuit Current	Sourcing to Ground $V_{ID} = 200mV$ ⁽⁶⁾	60 35	115		mA
		Sinking to Ground $V_{ID} = -200mV$ ⁽⁶⁾	85 65	145		
I _{OUT}	Output Current	$V_O = 0.5V$ from either supply	± 75			mA
PSRR	Power Supply Rejection Ratio	$(V^+, V^-) = (4.5V, -4.5V)$ to $(5.5V, -5.5V)$	78	90		dB
I _S	Supply Current (per channel)	No Load		2.70	4.50 5.50	mA

(6) Short circuit test is a momentary test. Output short circuit duration is infinite for $V_S < 6V$ at room temperature and below. For $V_S > 6V$, allowable short circuit duration is 1.5ms.

7.8 Typical Performance Characteristics

$V^+ = +5V$, $V^- = -5V$, $R_F = R_L = 2k\Omega$. Unless otherwise specified.

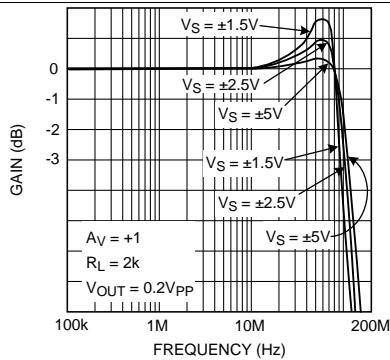


Figure 1. Closed Loop Frequency Response for Various Supplies

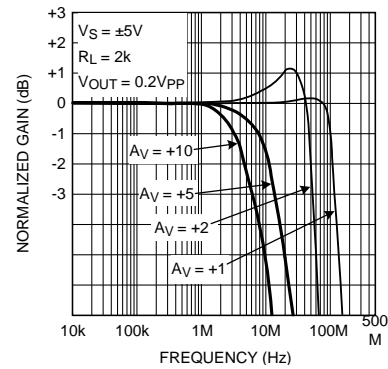


Figure 2. Closed Loop Gain vs. Frequency for Various Gain

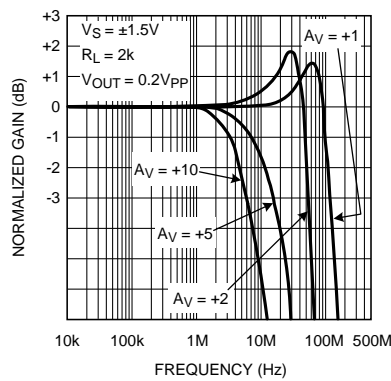


Figure 3. Closed Loop Gain vs. Frequency for Various Gain

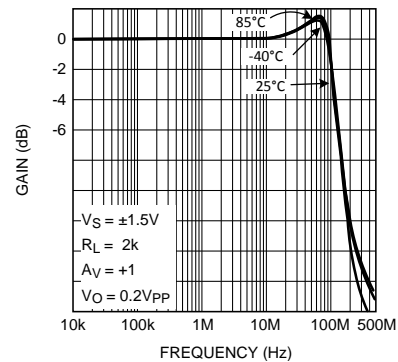


Figure 4. Closed Loop Frequency Response for Various Temperature

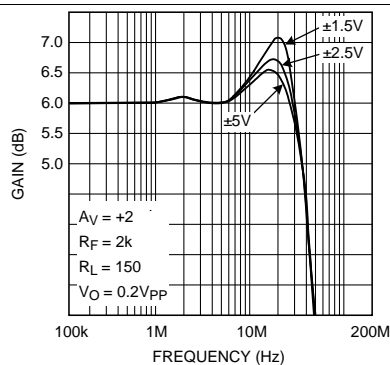


Figure 5. Closed Loop Gain vs. Frequency for Various Supplies

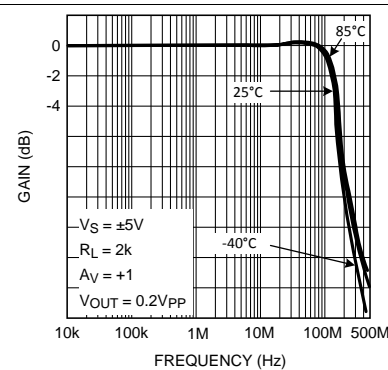


Figure 6. Closed Loop Frequency Response for Various Temperature

Typical Performance Characteristics (continued)

$V^+ = +5V$, $V^- = -5V$, $R_F = R_L = 2k\Omega$. Unless otherwise specified.

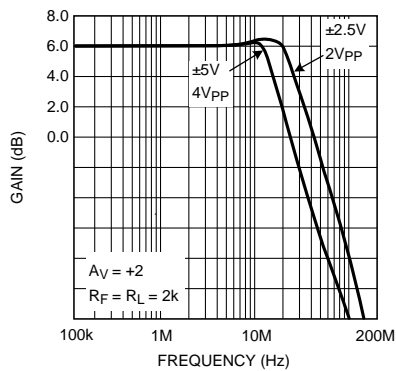


Figure 7. Large Signal Frequency Response

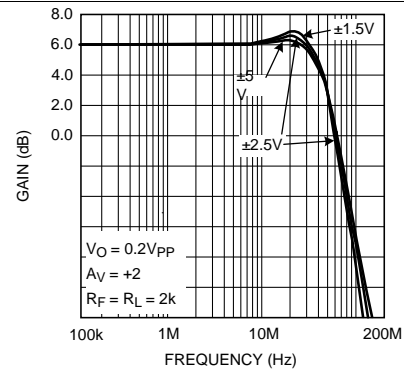


Figure 8. Closed Loop Small Signal Frequency Response for Various Supplies

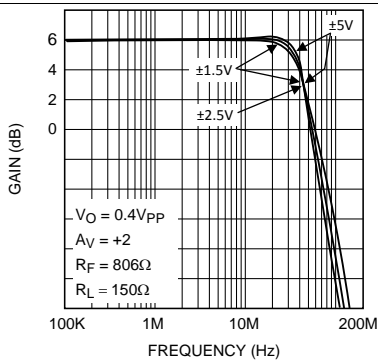


Figure 9. Closed Loop Frequency Response for Various Supplies

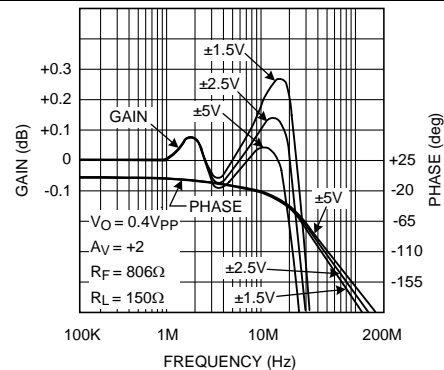


Figure 10. ± 0.1 dB Gain Flatness for Various Supplies

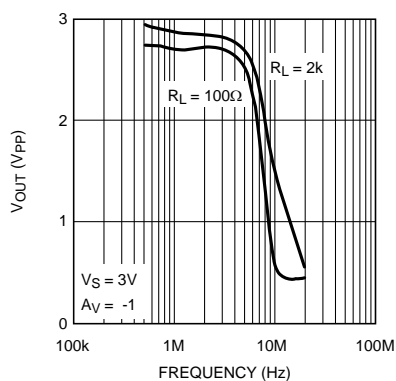


Figure 11. V_{OUT} (Vpp) for THD < 0.5%

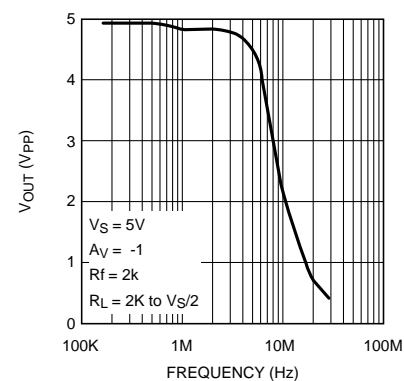


Figure 12. V_{OUT} (Vpp) for THD < 0.5%

Typical Performance Characteristics (continued)

$V^+ = +5V$, $V^- = -5V$, $R_F = R_L = 2k\Omega$. Unless otherwise specified.

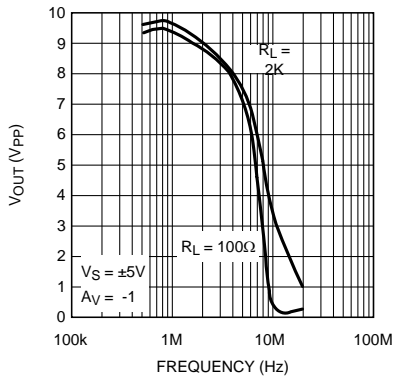


Figure 13. V_{OUT} (Vpp) for THD < 0.5%

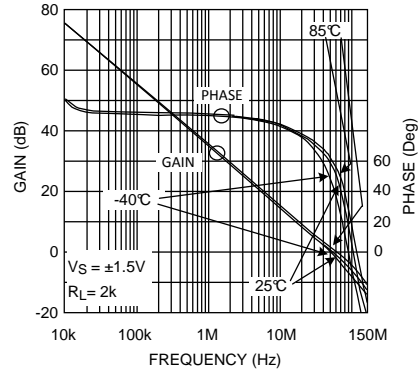


Figure 14. Open Loop Gain/Phase for Various Temperature

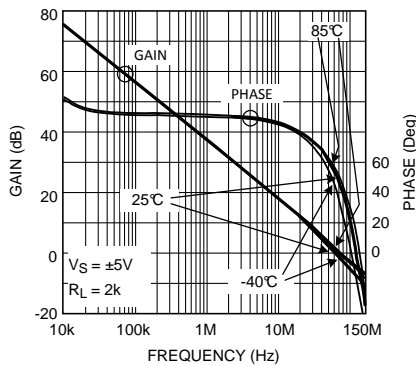


Figure 15. Open Loop Gain/Phase for Various Temperature

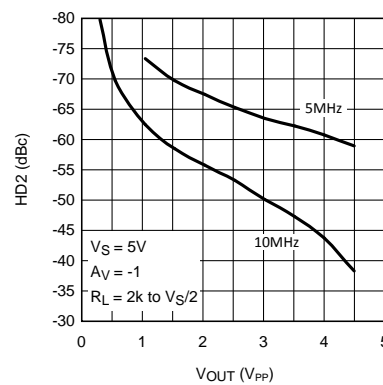


Figure 16. HD2 (dBc) vs. Output Swing

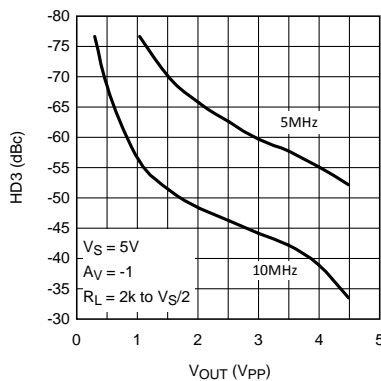


Figure 17. HD3 (dBc) vs. Output Swing

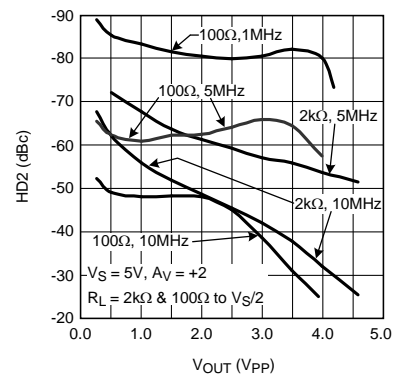


Figure 18. HD2 vs. Output Swing

Typical Performance Characteristics (continued)

$V^+ = +5V$, $V^- = -5V$, $R_F = R_L = 2k\Omega$. Unless otherwise specified.

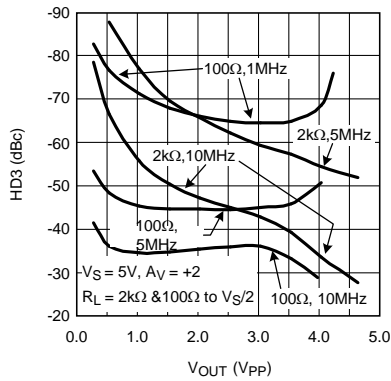


Figure 19. HD3 vs. Output Swing

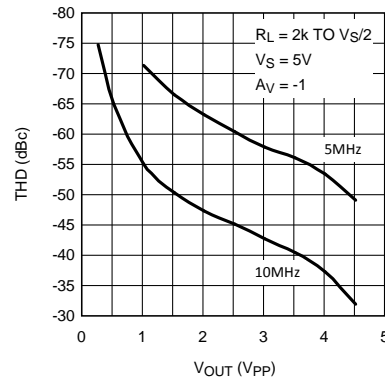


Figure 20. THD (dBc) vs. Output Swing

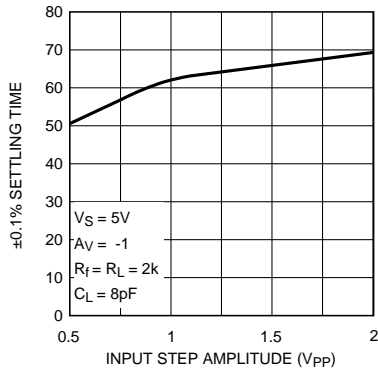


Figure 21. Settling Time vs. Input Step Amplitude (Output Slew and Settle Time)

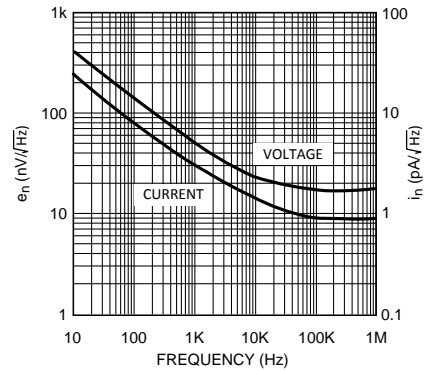


Figure 22. Input Noise vs. Frequency

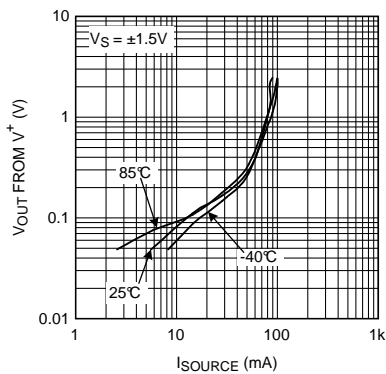


Figure 23. V_{OUT} from V^+ vs. I_{SOURCE}

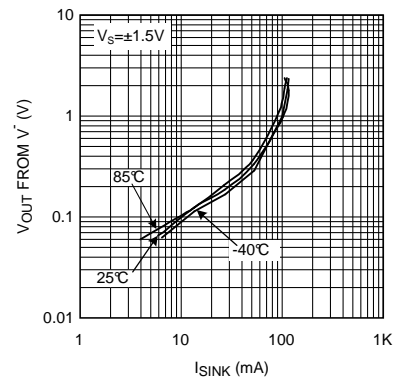


Figure 24. V_{OUT} from V^- vs. I_{SINK}

Typical Performance Characteristics (continued)

$V^+ = +5V$, $V^- = -5V$, $R_F = R_L = 2k\Omega$. Unless otherwise specified.

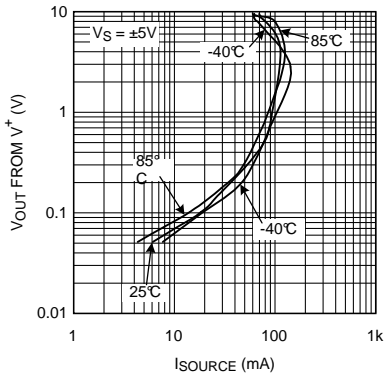


Figure 25. V_{OUT} from V^+ vs. I_{SOURCE}

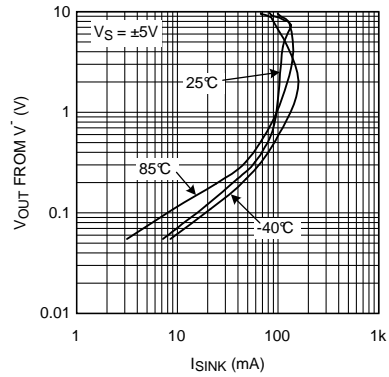


Figure 26. V_{OUT} from V^- vs. I_{SINK}

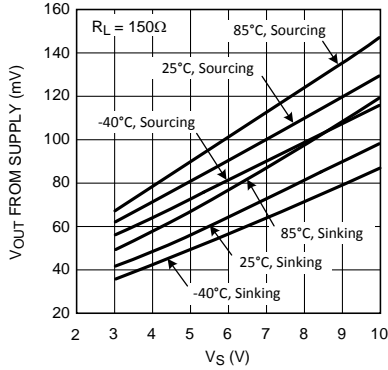


Figure 27. Swing vs. V_S

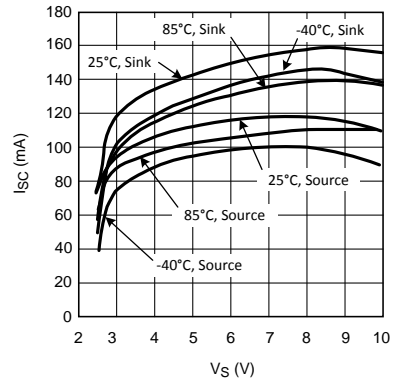


Figure 28. Short Circuit Current (to $V_S/2$) vs. V_S

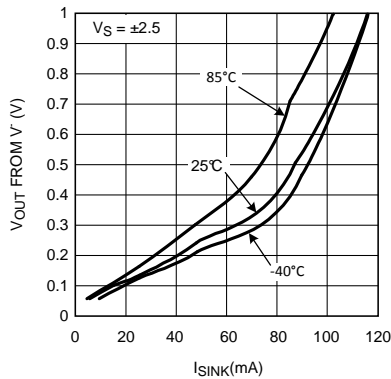


Figure 29. Output Sinking Saturation Voltage vs. I_{OUT}

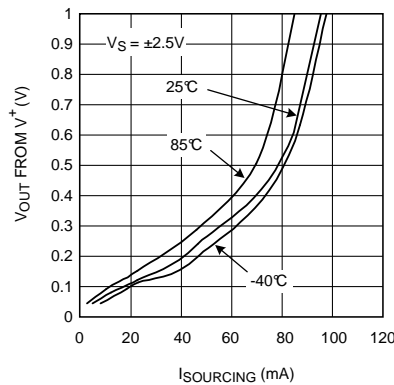


Figure 30. Output Sourcing Saturation Voltage vs. I_{OUT}

Typical Performance Characteristics (continued)

$V^+ = +5V$, $V^- = -5V$, $R_F = R_L = 2k\Omega$. Unless otherwise specified.

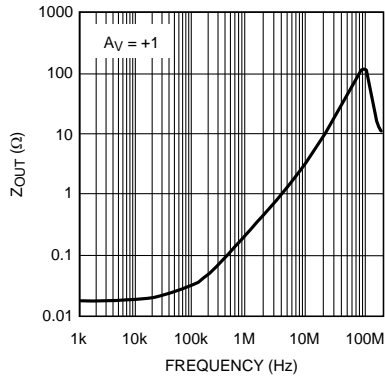


Figure 31. Closed Loop Output Impedance vs. Frequency, $A_V = +1$

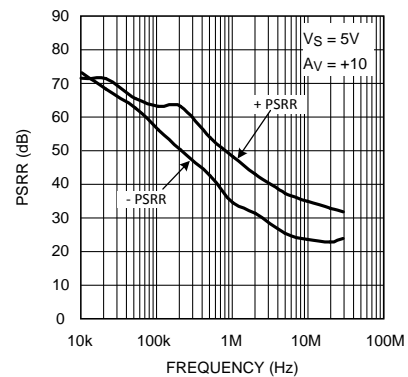


Figure 32. PSRR vs. Frequency

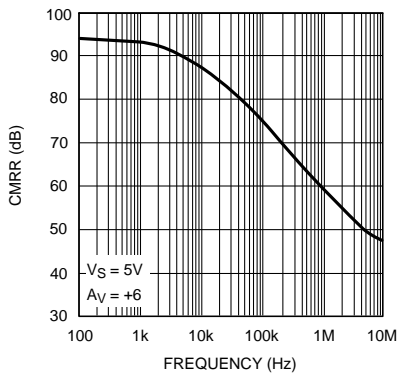


Figure 33. CMRR vs. Frequency

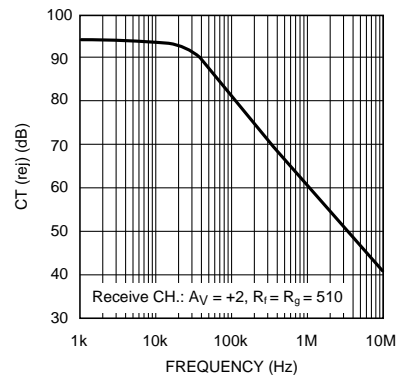


Figure 34. Crosstalk Rejection vs. Frequency (Output to Output)

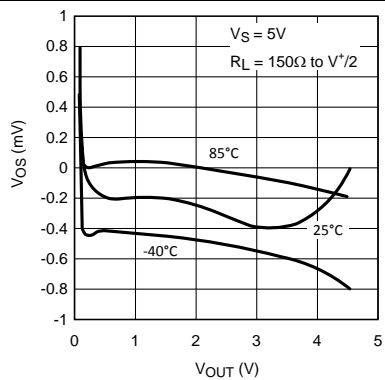


Figure 35. V_{OS} vs. V_{OUT} (Typical Unit)

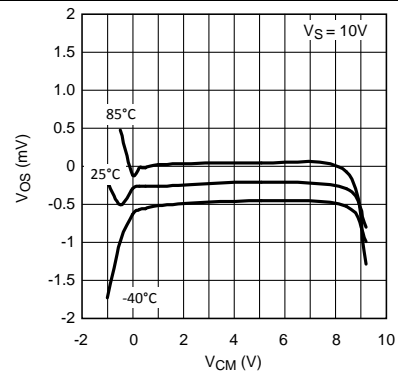


Figure 36. V_{OS} vs. V_{CM} (Typical Unit)

Typical Performance Characteristics (continued)

$V^+ = +5V$, $V^- = -5V$, $R_F = R_L = 2k\Omega$. Unless otherwise specified.

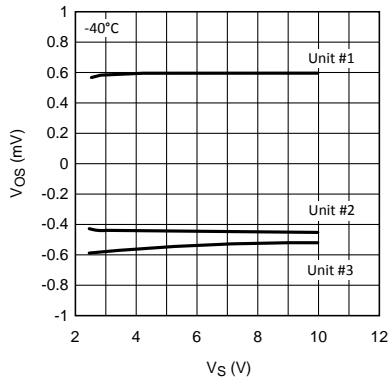


Figure 37. V_{OS} vs. V_S (for 3 Representative Units)

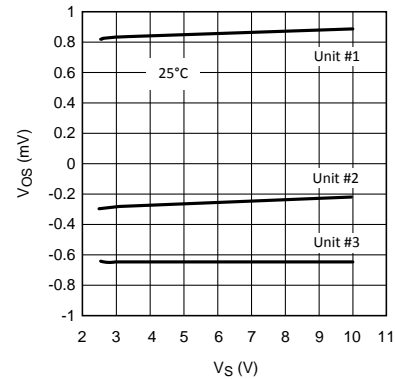


Figure 38. V_{OS} vs. V_S (for 3 Representative Units)

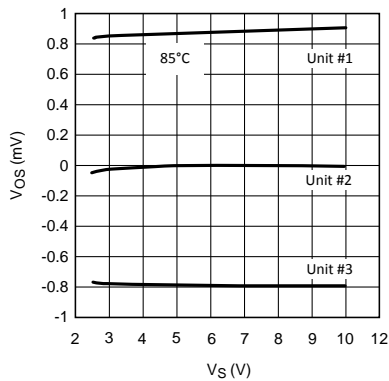


Figure 39. V_{OS} vs. V_S (for 3 Representative Units)

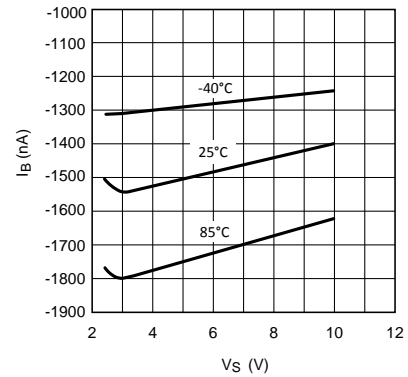


Figure 40. I_B vs. V_S

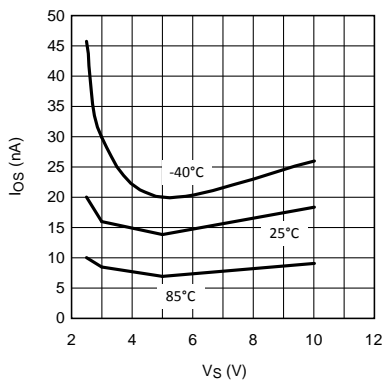


Figure 41. I_{OS} vs. V_S

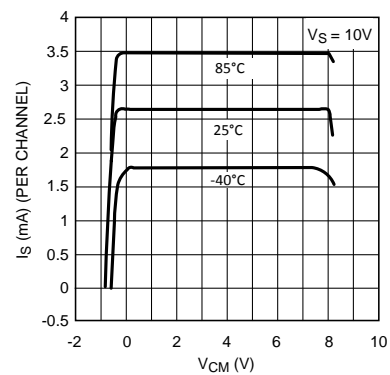


Figure 42. I_S vs. V_{CM}

Typical Performance Characteristics (continued)

$V^+ = +5V$, $V^- = -5V$, $R_F = R_L = 2k\Omega$. Unless otherwise specified.

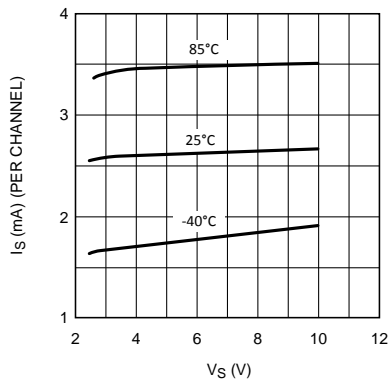


Figure 43. I_S vs. V_S

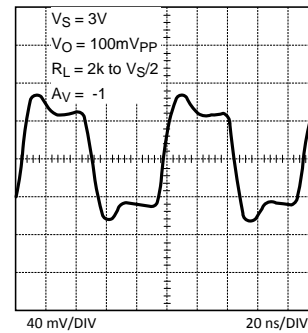


Figure 44. Small Signal Step Response

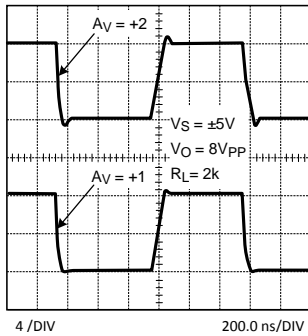


Figure 45. Large Signal Step Response

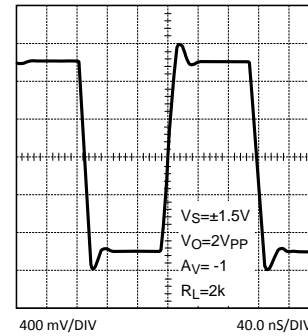


Figure 46. Large Signal Step Response

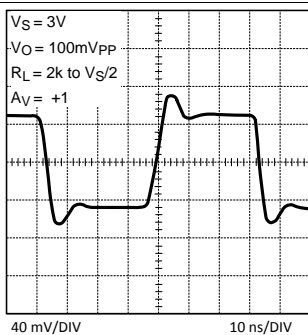


Figure 47. Small Signal Step Response

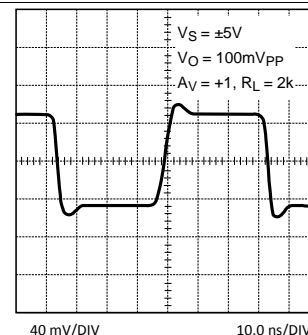


Figure 48. Small Signal Step Response

Typical Performance Characteristics (continued)

$V^+ = +5V$, $V^- = -5V$, $R_F = R_L = 2k\Omega$. Unless otherwise specified.

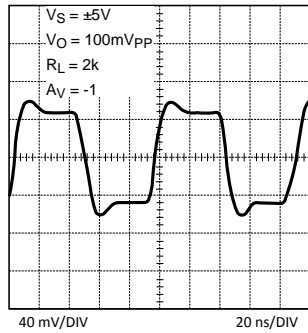


Figure 49. Small Signal Step Response

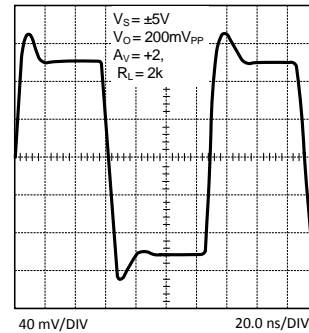


Figure 50. Small Signal Step Response

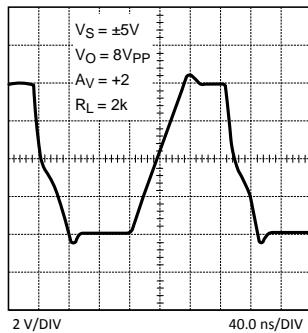


Figure 51. Large Signal Step Response

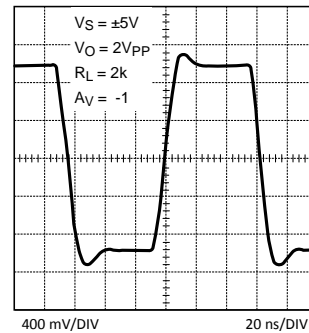


Figure 52. Large Signal Step Response

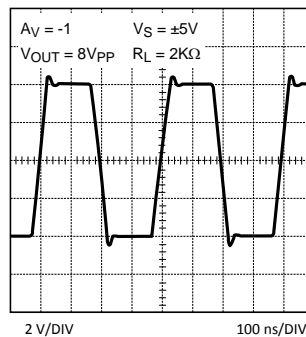


Figure 53. Large Signal Step Response

8 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Circuit Description

The LMH664X family is based on Texas Instruments' proprietary VIP10 dielectrically isolated bipolar process.

This device family architecture features the following:

- Complimentary bipolar devices with exceptionally high f_t (~8 GHz) even under low supply voltage (2.7 V) and low bias current.
- A class A-B “turn-around” stage with improved noise, offset, and reduced power dissipation compared to similar speed devices (patent pending).
- Common Emitter push-push output stage capable of 75mA output current (at 0.5V from the supply rails) while consuming only 2.7 mA of total supply current per channel. This architecture allows output to reach within milli-volts of either supply rail.
- Consistent performance over the entire operating supply voltage range with little variation for the most important specifications (for example, BW, SR, I_{OUT} , and so forth).
- Significant power saving (~40%) compared to competitive devices on the market with similar performance.

8.1.1 Application Hints

This Op Amp family is a drop-in replacement for the AD805X family of high speed Op Amps in most applications. In addition, the LMH664X will typically save about 40% on power dissipation, due to lower supply current, when compared to competition. All AD805X family's ensured parameters are included in the list of LMH664X ensured specifications in order to ensure equal or better level of performance. However, as in most high performance parts, due to subtleties of applications, it is strongly recommended that the performance of the part to be evaluated is tested under actual operating conditions to ensure full compliance to all specifications.

With 3V supplies and a common mode input voltage range that extends 0.5V below V^- , the LMH664X find applications in low voltage/low power applications. Even with 3V supplies, the -3dB BW (@ $A_V = +1$) is typically 115MHz with a tested limit of 80MHz. Production testing ensures that process variations with not compromise speed. High frequency response is exceptionally stable confining the typical -3dB BW over the industrial temperature range to $\pm 2.5\%$.

As can be seen from the [Typical Performance Characteristics](#), the LMH664X output current capability (~75mA) is enhanced compared to AD805X. This enhancement, increases the output load range, adding to the LMH664X's versatility.

Because of the LMH664X's high output current capability attention should be given to device junction temperature in order not to exceed the Absolute Maximum Rating.

Circuit Description (continued)

This device family was designed to avoid output phase reversal. With input overdrive, the output is kept near supply rail (or as closed to it as mandated by the closed loop gain setting and the input voltage). See [Figure 54](#):

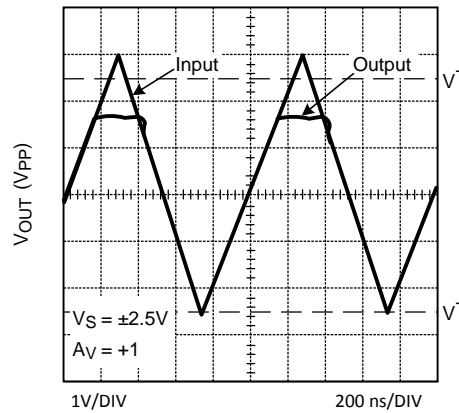


Figure 54. Input and Output Shown with CMVR Exceeded

However, if the input voltage range of $-0.5V$ to $1V$ from V^+ is exceeded by more than a diode drop, the internal ESD protection diodes will start to conduct. The current in the diodes should be kept at or below $10mA$.

Output overdrive recovery time is less than $100ns$ as can be seen from [Figure 55](#) plot:

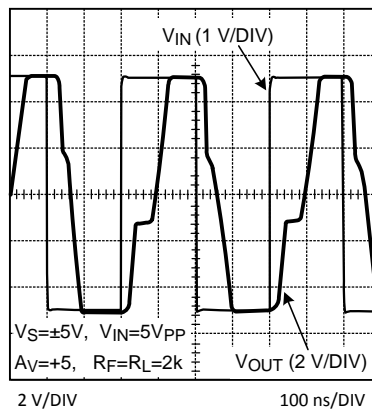


Figure 55. Overload Recovery Waveform

Circuit Description (continued)

8.1.2 Input and Output Topology

All input / output pins are protected against excessive voltages by ESD diodes connected to V^+ and V^- rails (see [Figure 56](#)). These diodes start conducting when the input / output pin voltage approaches $1V_{be}$ beyond V^+ or V^- to protect against over voltage. These diodes are normally reverse biased. Further protection of the inputs is provided by the two resistors (R in [Figure 56](#)), in conjunction with the string of anti-parallel diodes connected between both bases of the input stage. The combination of these resistors and diodes reduces excessive differential input voltages approaching $2V_{be}$. The most common situation when this occurs is when the device is used as a comparator (or with little or no feedback) and the device inputs no longer follow each other. In such a case, the diodes may conduct. As a consequence, input current increases and the differential input voltage is clamped. It is important to make sure that the subsequent current flow through the device input pins does not violate the Absolute Maximum Ratings of the device. To limit the current through this protection circuit, extra series resistors can be placed. Together with the built-in series resistors of several hundred ohms, these external resistors can limit the input current to a safe number (i.e. $< 10\text{mA}$). Be aware that these input series resistors may impact the switching speed of the device and could slow down the device.

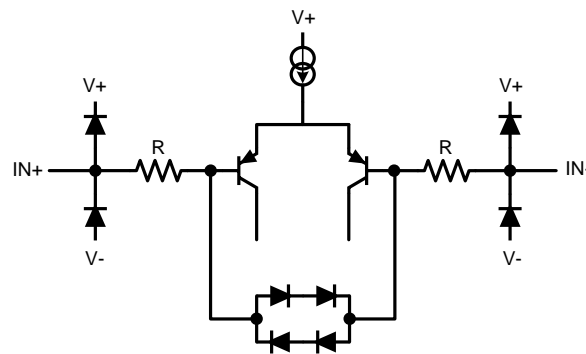


Figure 56. Input Equivalent Circuit

8.2 Single Supply, Low Power Photodiode Amplifier

The circuit shown in [Figure 57](#) is used to amplify the current from a photo-diode into a voltage output. In this circuit, the emphasis is on achieving high bandwidth and the transimpedance gain setting is kept relatively low. Because of its high slew rate limit and high speed, the LMH664X family lends itself well to such an application.

This circuit achieves approximately 1V/mA of transimpedance gain and capable of handling up to 1mA_{pp} from the photodiode. Q1, in a common base configuration, isolates the high capacitance of the photodiode (C_d) from the Op Amp input in order to maximize speed. Input is AC coupled through C1 to ease biasing and allow single supply operation. With 5V single supply, the device input/output is shifted to near half supply using a voltage divider from V_{CC}. Note that Q1 collector does not have any voltage swing and the Miller effect is minimized. D1, tied to Q1 base, is for temperature compensation of Q1's bias point. Q1 collector current was set to be large enough to handle the peak-to-peak photodiode excitation and not too large to shift the U1 output too far from mid-supply.

No matter how low an R_f is selected, there is a need for C_f in order to stabilize the circuit. The reason for this is that the Op Amp input capacitance and Q1 equivalent collector capacitance together (C_{IN}) will cause additional phase shift to the signal fed back to the inverting node. C_f will function as a zero in the feedback path counteracting the effect of the C_{IN} and acting to stabilize the circuit. By proper selection of C_f such that the Op Amp open loop gain is equal to the inverse of the feedback factor at that frequency, the response is optimized with a theoretical 45° phase margin.

$$C_F = \sim \text{SQRT} \left[\frac{C_{IN}}{(2\pi \cdot \text{GBWP} \cdot R_F)} \right]$$

where GBWP is the Gain Bandwidth Product of the Op Amp (1)

Optimized as such, the I-V converter will have a theoretical pole, f_p, at:

$$f_p = \text{SQRT} \left[\frac{\text{GBWP}}{(2\pi R_F \cdot C_{IN})} \right]$$
(2)

With Op Amp input capacitance of 3pF and an estimate for Q1 output capacitance of about 3pF as well, C_{IN} = 6pF. From [Typical Performance Characteristics](#), LMH6642/6643 family GBWP is approximately 57 MHz. Therefore, with R_f = 1k, from [Equation 1](#) and [Equation 2](#) above.

C_f = ~4.1 pF, and f_p = 39 MHz

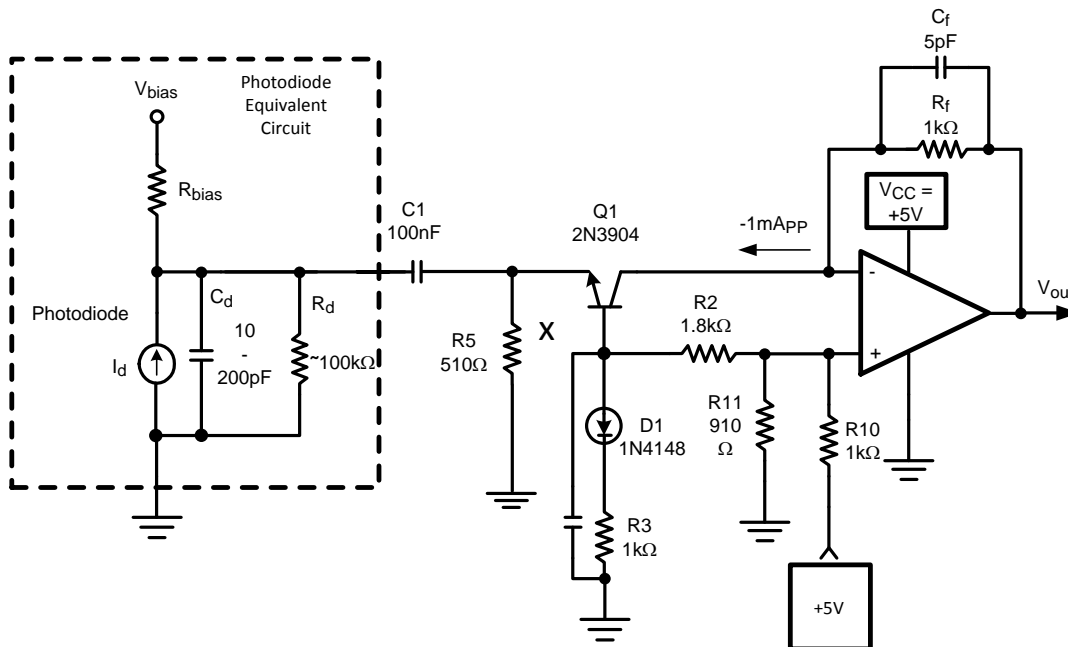


Figure 57. Single Supply Photodiode I-V Converter

Single Supply, Low Power Photodiode Amplifier (continued)

For this example, optimum C_f was empirically determined to be around 5pF. This time domain response is shown in [Figure 58](#) below showing about 9 ns rise/fall times, corresponding to about 39 MHz for f_p . The overall supply current from the +5 V supply is around 5 mA with no load.

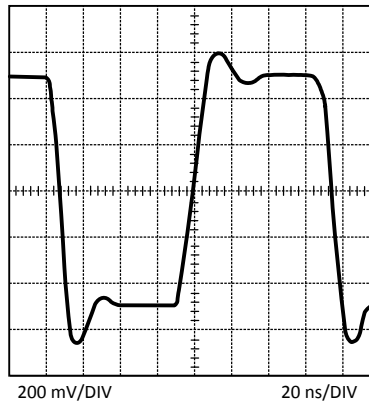


Figure 58. Converter Step Response (1V_{pp}, 20 ns/DIV)

8.3 Printed Circuit Board Layout and Component Values Section

Generally, a good high frequency layout will keep power supply and ground traces away from the inverting input and output pins. Parasitic capacitances on these nodes to ground will cause frequency response peaking and possible circuit oscillations (see Application Note OA-15 ([SNOA367](#)) for more information). Texas Instruments suggests the following evaluation boards as a guide for high frequency layout and as an aid in device testing and characterization:

DEVICE	PACKAGE	EVALUATION BOARD PN
LMH6642QMF	5-Pin SOT-23	LMH730216
LMH6643QMM	8-Pin VSSOP	LMH730123

Another important parameter in working with high speed/high performance amplifiers, is the component values selection. Choosing external resistors that are large in value will effect the closed loop behavior of the stage because of the interaction of these resistors with parasitic capacitances. These capacitors could be inherent to the device or a by-product of the board layout and component placement. Either way, keeping the resistor values lower, will diminish this interaction to a large extent. On the other hand, choosing very low value resistors could load down nodes and will contribute to higher overall power dissipation.

9 Device and Documentation Support

9.1 Documentation Support

9.1.1 Related Documentation

9.2 Related Links

The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

Table 1. Related Links

PARTS	PRODUCT FOLDER	SAMPLE & BUY	TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY
LMH6642Q-Q1	Click here	Click here	Click here	Click here	Click here
LMH6643Q-Q1	Click here	Click here	Click here	Click here	Click here

9.3 Trademarks

All trademarks are the property of their respective owners.

9.4 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

9.5 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

10 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LMH6642QMF/NOPB	ACTIVE	SOT-23	DBV	5	1000	RoHS & Green	SN	Level-1-260C-UNLIM	-40 to 85	A64Q	Samples
LMH6642QMF/NOPB	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	SN	Level-1-260C-UNLIM	-40 to 85	A64Q	Samples
LMH6643QMM/NOPB	ACTIVE	VSSOP	DGK	8	1000	RoHS & Green	SN	Level-1-260C-UNLIM	-40 to 85	643Q	Samples
LMH6643QMM/NOPB	ACTIVE	VSSOP	DGK	8	3500	RoHS & Green	SN	Level-1-260C-UNLIM	-40 to 85	643Q	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSELETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "-" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMH6642QMF/NOPB	SOT-23	DBV	5	1000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LMH6642QMF/NOPB	SOT-23	DBV	5	3000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LMH6643QMM/NOPB	VSSOP	DGK	8	1000	178.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LMH6643QMM/NOPB	VSSOP	DGK	8	3500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMH6642QMF/NOPB	SOT-23	DBV	5	1000	208.0	191.0	35.0
LMH6642QMFX/NOPB	SOT-23	DBV	5	3000	208.0	191.0	35.0
LMH6643QMM/NOPB	VSSOP	DGK	8	1000	208.0	191.0	35.0
LMH6643QMMX/NOPB	VSSOP	DGK	8	3500	367.0	367.0	35.0

DBV0005A



PACKAGE OUTLINE

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



4214839/K 08/2024

NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. Reference JEDEC MO-178.
4. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25 mm per side.
5. Support pin may differ or may not be present.

EXAMPLE BOARD LAYOUT

DBV0005A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:15X



SOLDER MASK DETAILS

4214839/K 08/2024

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

DBV0005A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE:15X

4214839/K 08/2024

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

DGK0008A



PACKAGE OUTLINE

VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



4214862/A 04/2023

NOTES:

PowerPAD is a trademark of Texas Instruments.

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-187.

EXAMPLE BOARD LAYOUT

DGK0008A

™ VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 15X



4214862/A 04/2023

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
8. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.
9. Size of metal pad may vary due to creepage requirement.

EXAMPLE STENCIL DESIGN

DGK0008A

™ VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



SOLDER PASTE EXAMPLE
SCALE: 15X

4214862/A 04/2023

NOTES: (continued)

11. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
12. Board assembly site may have different recommendations for stencil design.

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