BQ25882 I²C Controlled, 2-Cell, 2-A Boost-Mode Battery Charger For USB Input

1 Features

- High-efficiency 2-A, 1.5-MHz switch mode boost charger
  - 92.5% Charge efficiency at 7.6-V battery, 1-A charge
  - Optimized for USB input and 2-cell Li-Ion output
  - Selectable low power PFM mode for light load operation with out-of-audio option
- USB On-the-Go (OTG) with adjustable output from 4.5 V to 5.5 V
  - Buck converter with up-to 2-A output
  - 94.5% Efficiency at 5-V, 1-A output
  - Accurate constant current (CC) limit
  - Output short protection
  - Selectable low power PFM mode for light load operation with out-of-audio option
- Single Input to support USB input adapters
  - Supports 3.9-V to 6.2-V input voltage range with 20-V absolute maximum input voltage rating
  - Input current limit (500 mA to 3.3 A with 100-mA resolution) to support USB2.0, USB3.0 standard adapters
  - Maximum power tracking by input voltage limit up-to 5.5 V
  - Auto-detect USB SDP, CDP, DCP, and non-standard adapters
- Input current optimizer (ICO) to maximize input power without overloading adapters
- Highest battery discharge efficiency with 15-mΩ battery discharge MOSFET
- Integrated ADC for system monitoring (BUS voltage and current, BAT voltage, charge current, SYS voltage, and NTC and die temperature)
- Narrow VDC (NVDC) power path management
  - Instant-on works with no battery or deeply discharged battery
  - Ideal diode operation in battery supplement mode
- Flexible autonomous and I²C mode for optimal system performance
- High integration includes all MOSFETs, current sensing and loop compensation
- High accuracy
  - ±0.4% Charge voltage regulation
  - ±7.5% Charge current regulation
  - ±7.5% Input current regulation
- Safety
  - Battery temperature sensing in charge and OTG buck mode
  - Thermal regulation and thermal shutdown

2 Applications

- Wireless speaker
- Digital camera (DSC, DVC)
- Mobile printer
- Tablet
- Eletronic point of sales (ePOS)
- Portable electronic devices

3 Description

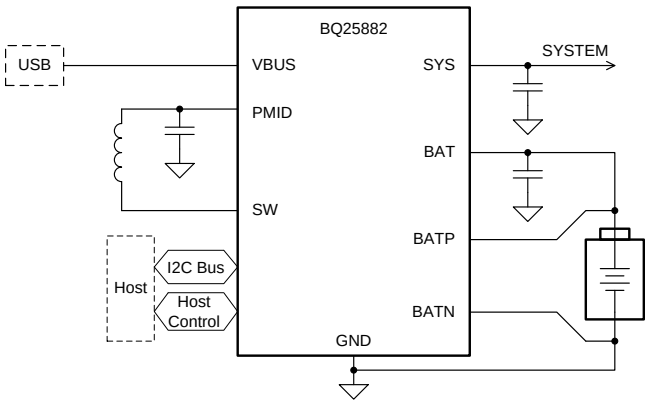
The BQ25882 is a highly-integrated 2-A switch-mode battery charge management and system power path management device for dual-cell Li-Ion and Li-polymer batteries. The I²C Serial interface with charging and system settings makes the device a truly flexible solution.

Device Information(1)

PART NUMBER	PACKAGE	BODY SIZE (NOM)
BQ25882	DSBGA (25)	2.10 x 2.10 mm²

(1) For all available packages, see the orderable addendum at the end of the data sheet.

Simplified Schematic



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4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision B (December 2018) to Revision C — **Page**

- Added Device Comparison Table section ... 3

Changes from Revision A (August 2018) to Revision B — **Page**

- Deleted bq25880 references from Functional Block Diagram ... 17
- Added ADC is enabled to list of conditions to enable buck operation in Buck Mode Operation from Battery (OTG) section ... 20

Changes from Original (February 2018) to Revision A — **Page**

- Changed from Advance Information to Production Data ... 1

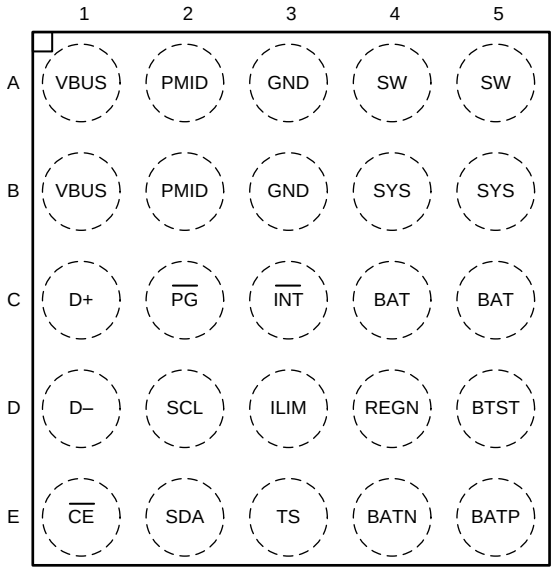
5 Device Comparison Table

Table 1. Device Comparison

PRT NUMBER	BQ25882	BQ25883	BQ25886	BQ25887
VBUS Operating Range	3.9 to 6.2 V	3.9 to 6.2 V	4.3 to 6.2 V	3.9 to 6.2 V
USB Detection	D+/D-	D+/D-	D+/D-	PSEL
Power Path	Yes	Yes	Yes	No
Cell Balancing	No	No	No	Yes
OTG	Up to 2 A	Up to 2 A	Up to 2 A	No OTG
16 bit ADC	Yes	Yes	No	Yes
Control Interface	I2C	I2C	Standalone	I2C
Status Pin	/PG	STAT, /PG	STAS, /PG	STAT, /PG
Package	2.1x2.1 WCSP-25	4x4 QFN-24	4x4 QFN-24	4x4 QFN-24

6 Pin Configuration and Functions

YFF (I^2C controlled)
25-Pin DSBGA
Top View



Top View = Xray through a soldered down part with A1 starting in upper left corner

Pin Functions

PIN		I/O	DESCRIPTION
NAME	NO.		
BAT	C4	P	**Battery Power Connection –** The internal BATFET is connected between SYS and BAT. Connect a 10µF ceramic capacitor closely to the BAT pin and GND.
	C5		
BATN	E4	AI	**Negative Battery Sense Terminal –** Kelvin connect as close as possible to negative battery terminal
BATP	E5	AI	**Positive Battery Sense Terminal –** Kelvin connect as close as possible to positive battery terminal
BTST	D5	P	**PWM High-side Driver Supply –** Internally, BTST is connected to the cathode of the boot-strap diode. Connect a 0.047µF bootstrap capacitor from SW to BTST.
$\overline{CE}$	E1	DI	**Active Low Charge Enable Pin –** Battery charging is enabled when EN_CHG bit is 1 and $\overline{CE}$ pin is LOW. $\overline{CE}$ pin must be pulled HIGH or LOW, do not leave floating.
D+	C1	AIO	**Positive USB data line –** D+/D– based USB host/charging port detection. The detection includes data contact detection (DCD) and secondary detection in BC1.2.
D–	D1	AIO	**Negative USB data line –** D+/D– based USB host/charging port detection. The detection includes data contact detection (DCD) and secondary detection in BC1.2.
GND	A3	—	**Ground Return**
	B3		
ILIM	D3	AI	**Input Current Limit –** ILIM pin sets the maximum input current and can be used to monitor input current. IINDPM loop regulates ILIM pin voltage at 0.8 V. When ILIM pin is less than 0.8 V, the input current limit can be calculated by IIN = KILIM x VILIM / (R_{ILIM} x 0.8 V). A resistor connected from ILIM pin to ground sets the current limit as IINMAX = KILIM / R_{ILIM}. The actual input current limit is the lower limit set by ILIM pin (when EN_ILIM bit is HIGH) or IINDPM register bits. Input current limit less than 500mA is not supported on ILIM pin. The ILIM pin function can be disabled when EN_ILIM bit is 0.
$\overline{INT}$	C3	DO	**Open Drain Active Low Interrupt Output –** Connect /INT to the logic rail via a 10-kΩ resistor. The $\overline{INT}$ pin sends active low, 256-µs pulse to the host to report charger device status and fault.
$\overline{PG}$	C2	DO	**Open Drain Active Low Power Good Indicator –** Connect to the pull up rail via 10-kΩ resistor. LOW indicates a good input source if the input voltage is within VVBUS_OP, and can provide more than IPOORSRC.
PMID	A2	P	**Blocking MOSFET Connection –** Given the total input capacitance, place 1µF on VBUS, and the rest on PMID, as close to the IC as possible. Typical value: 10µF ceramic capacitor
	B2		

Pin Functions (continued)

PIN		I/O	DESCRIPTION
NAME	NO.		
REGN	D4	P	**Gate Drive Supply –** Bias supply for internal MOSFETs driver and IC. Bypass REGN to GND with a 4.7µF ceramic capacitor.
SCL	D2	DI	**I2C Interface Clock –** Connect SCL to the logic rail through a 10-kΩ resistor.
SDA	E2	DIO	**I2C Interface Data –** Connect SDA to the logic rail through a 10-kΩ resistor.
SW	A4	P	**Inductor Connection –** Connect to the switched side of the external inductor.
	A5		
SYS	B4	P	**System Connection –** The internal BATFET is connected between SYS and BAT. When the battery falls below the minimum system voltage, switch-mode converter keeps SYS above the minimum system voltage. Connect a 44µF ceramic capacitor closely to the SYS pin and GND.
	B5		
TS	E3	AI	**Temperature Qualification Voltage –** Connect a negative temperature coefficient thermistor. Program temperature window with a resistor divider from REGN to TS to GND. Charge suspends when TS pin is out of range. Recommend 103AT-2 thermistor.
VBUS	A1	P	**Input Supply –** VBUS is connected to the external DC supply. Bypass VBUS to GND with at least 1µF ceramic capacitor, placed as close to the IC as possible.
	B1		

7 Specifications

7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1)

		MIN	MAX	UNIT
Voltage Range (with respect to GND unless otherwise specified)	VBUS (converter not switching)	–0.3	20	V
	PMID (converter not switching)	–0.3	8.5	V
	BAT, SYS (converter not switching)	–0.3	12	V
	SW	–0.6(2)	13	V
	BTST	–0.3	19	V
	BATP	–0.3	12	V
	BATN, REGN, SDA, SCL, $\overline{INT}$, $\overline{CE}$, TS, D+, D–, $\overline{PG}$	–0.3	6	V
	ILIM	–0.3	5	V
	BTST to SW	–0.3	6	V
	SYS to BAT	–0.3	8.5	V
Output Sink Current	$\overline{INT}$, $\overline{PG}$		6	mA
Junction Temperature, T_J		–40	150	°C
Storage temperature, T_{stg}		–40	150	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) –2V for 50ns

7.2 ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001(1)	±2000	V
		Charged device model (CDM), per JEDEC specification JESD22-C101(2)	±250	V

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V_{VBUS}	Input voltage	3.9		6.2	V
I_{VBUS}	Average input current (VBUS)			3.3	A
I_{BAT}	Average charge current (BAT)			2.2	A
V_{BAT}	Battery voltage (BATP - BATN)			9.2(1)	V
I_{BAT_RMS}	RMS discharging current with internal MOSFET			4	A
I_{BAT_PK}	Peak discharging current with internal MOSFET			8	A
T_A	Operating free-air temperature	–40		85	°C

(1) The inherent switching noise voltage spikes should not exceed the absolute maximum rating on SW pins. A tight layout minimizes switching noise.

7.4 Thermal Information

over operating free-air temperature range (unless otherwise noted)

THERMAL METRIC(1)		BQ25882 YFF (DSBGA) 25-BALL	UNIT
$R_{\Theta JA}$	Junction-to-ambient thermal resistance (EVM(2))	24	°C/W
$R_{\Theta JA}$	Junction-to-ambient thermal resistance (JEDEC (1))	64.4	°C/W
$R_{\Theta JC(top)}$	Junction-to-case (top) thermal resistance	0.4	°C/W
$R_{\Theta JB}$	Junction-to-board thermal resistance	14.8	°C/W
ψ_{JT}	Junction-to-top characterization parameter	0.2	°C/W
ψ_{JB}	Junction-to-board characterization parameter	14.9	°C/W
$R_{\Theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	N/A	°C/W

(1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report.
(2) Measured on 35µm thick copper, 4-layer board.

7.5 Electrical Characteristics

$V_{VBUS\_UVLOZ} < V_{VBUS} < V_{VBUS\_OV}$, T_J = -40°C to+125°C, and T_J = 25°C for typical values (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
QUIESCENT CURRENTS						
I_{BAT}	Battery discharge current (BATP, BAT, SYS)	VBAT = 9 V, No VBUS, SCL, SDA = 0 V or 1.8 V, ADC Disabled, T_J = 25℃		11.5	14	µA
		VBAT = 9 V, No VBUS, SCL, SDA = 0 V or 1.8 V, ADC Disabled, T_J < 85℃		11.5	20	µA
I_{VBUS_HIZ}	Input supply current (VBUS) in HIZ	VBUS = 5 V, High-Z Mode, no battery, ADC Disabled, T_J = 25℃		30	35	µA
		VBUS = 5 V, High-Z Mode, no battery, ADC Disabled, T_J < 85℃		30	40	µA
I_{VBUS}	Input supply current (VBUS)	VBUS = 5 V, V_{BAT} = 7.6 V, converter not switching		1.5	3	mA
		VBUS = 5 V, V_{BAT} = 7.6 V, converter switching, I_{SYS} = 0A		3		mA
I_{BAT_OTG}	Battery discharge current in OTG mode	VBAT = 8.4 V, OTG Buck Mode, I_{VBUS} = 0A, converter switching		3		mA
VBUS/VBAT POWER UP						
V_{VBUS_OP}	VBUS operating range		3.9		6.2	V
V_{VBUS_UVLOZ}	VBUS rising for active I2C, no battery	VBUS rising		3.3	3.6	V
$V_{VBUS\_PRESENT}$		VBUS rising		3.65	3.9	V
V_{VBUS_OV}	VBUS over-voltage rising threshold	VBUS rising	6.2		6.6	V
	VBUS over-voltage falling threshold	VBUS falling	5.9		6.4	V
V_{BAT_UVLOZ}	Battery for active I2C	VBAT rising	3.7	4.0	4.335	V
$V_{POORSRC}$	Bad adapter detection threshold			3.7		V
$I_{POORSRC}$	Bad adapter detection current source			15		mA
POWER-PATH						
V_{SYS}	Typical System Regulation Voltage	ISYS = 0A, VBAT = 8.80 V > SYS_MIN[3:0], Charge Disabled (EN_CHG = 0). Offset above VBAT		100		mV
V_{SYS}	Typical System Regulation Voltage	ISYS = 0A, VBAT < SYS_MIN[3:0], Charge Disabled (EN_CHG = 0). Offset above SYS_MIN		200		mV
V_{SYS_MIN}	System Regulation Voltage	VBAT < SYS_MIN[3:0] = 1010, Charge Disabled (EN_CHG = 0)	7	7.2		V
R_{ON_QBLK} (Q1)	Blocking MOSFET on-resistance between VBUS and PMID (QB)	T_J = 25°C		25	36	mΩ
		T_J = – 40°C - 125°C		25	52	mΩ

Electrical Characteristics (continued)

V_{VBUS_UVLOZ} < V_{VBUS} < V_{VBUS_OV}, T_J = -40°C to+125°C, and T_J = 25°C for typical values (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
R_{ON_QHS} (Q2)	High-side switching MOSFET on-resistance between SW and SYS (Q2)	T_J = 25°C		30	40	mΩ
		T_J = – 40°C - 125°C		30	55	mΩ
R_{ON_QLS} (Q3)	Low-side switching MOSFET on-resistance between SW and GND (Q3)	T_J = 25°C		40	59	mΩ
		T_J = – 40°C - 125°C		40	80	mΩ
BATTERY CHARGER						
V_{REG_RANGE}	Typical charge voltage regulation range		6.8		9.2	V
V_{REG_STEP}	Typical charge voltage step			10		mV
V_{REG_ACC}	Charge voltage accuracy	VREG = 8.40 V, T_J = – 40°C - 85°C	8.3664	8.4	8.4336	V
		VREG = 8.70 V, T_J = – 40°C - 85°C	8.6652	8.7	8.7348	V
		VREG = 8.80 V, T_J = – 40°C - 85°C	8.7648	8.8	8.8352	V
I_{CHG_RANGE}	Charge current regulation range		0		2200	mA
I_{CHG_STEP}	Charge current regulation step			50		mA
I_{CHG_ACC}	Fast Charge current regulation accuracy	ICHG = 250 mA, VBAT = 6.2 V or 7.6 V, T_J = – 20°C - 85°C	-25		25	%
		ICHG = 500 mA, VBAT = 6.2 V or 7.6 V, T_J = – 20°C - 85°C	-10		10	%
		ICHG = 1000 mA, VBAT = 6.2 V or 7.6 V, T_J = – 20°C - 85°C	-7.5		7.5	%
I_{PRECHG_RANGE}	Precharge current range		50		800	mA
I_{PRECHG_STEP}	Typical precharge current step			50		mA
I_{PRECHG_ACC}	Precharge current accuracy	VBAT = 5.2 V, IPRECHG = 200 mA, T_J = – 20°C - 85°C	-20		20	%
I_{TERM_RANGE}	Termination current range		50		800	mA
I_{TERM_STEP}	Typical termination current step			50		mA
I_{TERM_ACC}	Termination current accuracy	ICHG = 1.5A, ITERM = 50 mA, T_J = – 40°C - 85°C	-40		40	%
		ICHG = 1.5A, ITERM = 150 mA, T_J = – 40°C - 85°C	-20		20	%
V_{BAT_SHORT}	Short Battery Voltage rising threshold to start pre-charging	VBAT rising	4.1	4.4	4.7	V
$V_{BAT\_SHORT\_HYS}$	Short Battery Voltage falling threshold to stop pre-charging	VBAT falling	3.7	4.0	4.3	V
I_{BAT_SHORT}	Short Battery Voltage trickle charging current	VBAT < 4.4 V		100		mA
V_{BAT_LOWV}	VBAT LOWV Rising threshold to start fast-charging	VBAT rising, VBATLOW = 6.0 V	5.7	6	6.3	V
	VBAT LOWV Falling threshold to stop fast-charging	VBAT falling, VBATLOW = 6.0 V	5.3	5.6	5.9	V
V_{RECHG}	Recharge threshold below V_{REG}	VBAT falling, VRECHG[1:0] = 01		200		mV
		VBAT falling, VRECHG[1:0] = 10		300		mV
R_{ON_QBAT} (Q4)	MOSFET on-resistance between SYS and BAT (Q4)	T_J = 25°C		15	18	mΩ
		T_J = – 40°C - 125°C		15	26	mΩ
R_{BATP}	BATP Input resistance	VBAT = 8 V, VBUS = 5 V, EN_HIZ = 1, ADC Disabled		2.7		MΩ
R_{BATN}	BATN Input resistance	VBAT = 8 V, VBUS = 5 V, EN_HIZ = 1, ADC Disabled		2.7		MΩ
INPUT VOLTAGE / CURRENT REGULATION						
V_{INDPM_RANGE}	Input voltage regulation range		3.9		5.5	V

Electrical Characteristics (continued)

$V_{VBUS\_UVLOZ} < V_{VBUS} < V_{VBUS\_OV}$, T_J = -40°C to+125°C, and T_J = 25°C for typical values (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{INDPM_STEP}	Input voltage regulation step			100		mV
V_{INDPM}	Input voltage limit	VINDPM = 3.9 V	3.783	3.9	4.017	V
		VINDPM = 4.4 V	4.268	4.4	4.532	V
I_{INDPM_RANGE}	Input current regulation range		500		3300	mA
I_{INDPM_STEP}	Input current regulation step			100		mA
I_{INDPM_ACC}	Input current regulation limit	IINDPM = 500 mA	445	470	500	mA
		IINDPM = 900 mA	765	832.5	900	mA
		IINDPM = 2500 mA	2125	2312.5	2500	mA
		IINDPM = 3000 mA	2550	2775	3000	mA
K_{ILIM}		$I_{INMAX} = K_{ILIM}/R_{ILIM}$, Input Current regulation by ILIM pin = 1.5A	1000	1085	1170	A x Ω
D+/D- DETECTION						
$V_{D+D-\_600MVSRC}$	D+/D- Voltage Source (600 mV)		500	600	700	mV
$I_{D+\_10UASRC}$	D+ Current Source (10 µA)		7	10	14	µA
$I_{D+D-\_100UASNK}$	D+/D- Current Sink (100 µA)		50	100	150	µA
V_{D+D-_0P325}	D+/D- Comparator Threshold for Secondary Detection		250		400	mV
V_{D+_0P8}	D+Comparator Threshold for Data Contact Detection				800	mV
R_{D-_19K}	D- Resistor to Ground (19 kΩ)		14.25		24.8	kΩ
V_{D+D-_1P2}	D+/D- Threshold for Non-standard adapter		1.05		1.35	V
V_{D+D-_2P0}	D+/D- Threshold for Non-standard adapter		1.85		2.15	V
V_{D+D-_2P8}	D+/D- Threshold for Non-standard adapter		2.55		2.85	V
I_{D+D-_LKG}	D+/D- Leakage Current	HiZ	-1		1	µA
BATTERY OVER-VOLTAGE PROTECTION						
V_{BAT_OVP}	Battery over-voltage rising threshold	VBAT rising, as percentage of VREG	103	104	105	%
	Battery over-voltage falling threshold	VBAT falling, as percentage of VREG	101	102	103	%
THERMAL REGULATION AND THERMAL SHUTDOWN						
T_{REG}	Junction temperature regulation accuracy	TREG = 120°C		120		°C
T_{SHUT}	Thermal Shutdown Rising threshold	Temperature Increasing		150		°C
	Thermal Shutdown Falling threshold	Temperature Decreasing		120		°C
JEITA THERMISTOR COMPARATOR (BOOST MODE)						
V_{T1}	T1 (0°C) threshold, Charge suspended below this temperature.	As Percentage to REGN	72.75	73.25	73.75	%
V_{T1_HYS}	Charge re-enabled to ICHG/2 and VREG above this temperature	As Percentage to REGN		1.3		%
V_{T2}	T2 (10°C) threshold, Charge back to ICHG/2 and VREG below this temperature	As Percentage to REGN	67.75	68.25	68.75	%
V_{T2_HYS}	Charge back to ICHG and VREG above this temperature	As Percentage to REGN		1.3		%
V_{T3}	T3 (45°C) threshold, Charge back to ICHG and 8.1 V above this temperature.	As Percentage to REGN	44.25	44.75	45.25	%
V_{T3_HYS}	Charge back to ICHG and VREG below this temperature	As Percentage to REGN		1		%

Electrical Characteristics (continued)

$V_{VBUS\_UVLOZ} < V_{VBUS} < V_{VBUS\_OV}$, T_J = -40°C to+125°C, and T_J = 25°C for typical values (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{T5}	T5 (60°C) threshold, charge suspended above this temperature.	As Percentage to REGN	33.875	34.375	34.875	%
V_{T5_HYS}	Charge back to ICHG and 8.1 V below this temperature	As Percentage to REGN		1.3		%
COLD/HOT THERMISTOR COMPARATOR (OTG BUCK MODE)						
V_{BCOLD0}	Cold Temperature Threshold 0, TS pin Voltage Rising Threshold	As Percentage to REGN, BCOLD = 0 (Approx. – 10°C w/ 103AT)	76.5	77	77.5	%
V_{BCOLD0_HYS}	Cold Temperature Threshold 0, TS pin Voltage Falling Threshold	As Percentage to REGN		1		%
V_{BCOLD1}	Cold Temperature Threshold 1, TS pin Voltage Rising Threshold	As Percentage to REGN, BCOLD = 1 (Approx. – 20°C w/ 103AT)	79.5	80	80.5	%
V_{BCOLD1_HYS}	Cold Temperature Threshold 1, TS pin Voltage Falling Threshold	As Percentage to REGN		1		%
V_{BHOT0}	Hot Temperature Threshold 0, TS pin Voltage Falling Threshold	As Percentage to REGN, BHOT[1:0] = 01 (Approx. 55°C w/ 103AT)	37.25	37.75	38.25	%
V_{BHOT0_HYS}	Hot Temperature Threshold 0, TS pin Voltage Rising Threshold	As Percentage to REGN		3		%
V_{BHOT1}	Hot Temperature Threshold 1, TS pin Voltage falling Threshold	As Percentage to REGN, BHOT[1:0] = 00 (Approx. 60°C w/ 103AT)	33.875	34.375	34.875	%
V_{BHOT1_HYS}	Hot Temperature Threshold 1, TS pin Voltage rising Threshold	As Percentage to REGN		3		%
V_{BHOT2}	Hot Temperature Threshold 2, TS pin Voltage falling Threshold	As Percentage to REGN, BHOT[1:0] = 10 (Approx. 65°C w/ 103AT)	30.75	31.25	31.75	%
V_{BHOT1_HY2}	Hot Temperature Threshold 2, TS pin Voltage rising Threshold	As Percentage to REGN		3		%
BOOST MODE CONVERTER						
F_{SW}	PWM switching frequency	Oscillator frequency	1.35	1.5	1.65	MHz
OTG BUCK MODE CONVERTER						
V_{OTG_BAT}	Battery voltage exiting OTG mode	BAT falling	5.85	6	6.15	V
V_{OTG_RANGE}	Typical OTG Buck mode voltage regulation range		4.5		5.5	V
V_{OTG_STEP}	Typical OTG Buck mode voltage regulation step			100		mV
V_{OTG_ACC}	OTG Buck mode voltage regulation accuracy	IVBUS = 0A, OTG_VLIM = 5 V	-3		3	%
I_{OTG_RANGE}	Typical OTG Buck mode current regulation range		0.5		2	A
I_{OTG_STEP}	Typical OTG Buck mode current regulation step			100		mA
I_{OTG_ACC}	OTG Buck mode current regulation accuracy	OTG_ILIM = 1A	-15	-7.5	0.05	%
V_{OTG_OVP}	OTG Buck mode over-voltage threshold		5.8	6		V
REGN LDO						
V_{REGN}	REGN LDO output voltage	V_{VBUS} = 5 V, I_{REGN} = 20 mA	4.7	4.8		V
I_{REGN}	REGN LDO current limit	V_{VBUS} = 5 V, V_{REGN} = 3.8 V	50			mA
Analog-to-Digital Converter (ADC)						
t_{ADC_CONV}	Conversion time, each measurement	ADC_SAMPLE[1:0] = 00		24		ms
		ADC_SAMPLE[1:0] = 01		12		ms
		ADC_SAMPLE[1:0] = 10		6		ms
		ADC_SAMPLE[1:0] = 11		3		ms

Electrical Characteristics (continued)

V_{VBUS_UVLOZ} < V_{VBUS} < V_{VBUS_OV}, T_J = -40°C to+125°C, and T_J = 25°C for typical values (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
ADCRES	Effective resolution	ADC_SAMPLE[1:0] = 00	14	15		bits
		ADC_SAMPLE[1:0] = 01	13	14		bits
		ADC_SAMPLE[1:0] = 10	12	13		bits
		ADC_SAMPLE[1:0] = 11	10	11		bits
ADC MEASUREMENT RANGES AND LSB						
$I_{BUS\_ADC\_RANGE}$	ADC BUS current range		0		4	A
$I_{BUS\_ADC\_LSB}$	ADC BUS current LSB			1		mA
$I_{BAT\_ADC\_RANGE}$	ADC BAT current range		0		4	A
$I_{BAT\_ADC\_LSB}$	ADC BAT current LSB			1		mA
$V_{BUS\_ADC\_RANGE}$	ADC BUS voltage range		0		6.5	V
$V_{BUS\_ADC\_LSB}$	ADC BUS voltage LSB			1		mV
$V_{SYS\_ADC\_RANGE}$	ADC SYS voltage range		0		10	V
$V_{SYS\_ADC\_LSB}$	ADC SYS voltage LSB			1		mV
$V_{BAT\_ADC\_RANGE}$	ADC BAT voltage range		0		10	V
$V_{BAT\_ADC\_LSB}$	ADC BAT voltage LSB			1		mV
$V_{TS\_ADC\_RANGE}$	ADC TS voltage range		20		80	%
$V_{TS\_ADC\_LSB}$	ADC TS voltage LSB			0.098		%
$V_{TDIE\_ADC\_RANGE}$	ADC Die temperature range		0		150	°C
$V_{TDIE\_ADC\_LSB}$	ADC Die temperature LSB			1		°C
I2C INTERFACE (SCL, SDA)						
V_{IH}	Input high threshold level, SDA and SCL	Pull-up rail 1.8 V	1.3			V
V_{IL}	Input low threshold level	Pull-up rail 1.8 V			0.4	V
V_{OL}	Output low threshold level	Sink current = 5 mA			0.4	V
I_{BIAS}	High level leakage current	Pull-up rail 1.8 V			1	µA
LOGIC I/O PIN (/CE, PSEL)						
V_{IH}	Input high threshold level		1.3			V
V_{IL}	Input low threshold level				0.4	V
I_{IN_BIAS}	High level leakage current	Pull-up rail 1.8 V			1	µA
LOGIC O PIN (/INT, /PG, STAT)						
V_{OL}	Output low threshold level	Sink current = 5 mA			0.4	V
I_{OUT_BIAS}	High level leakage current	Pull-up rail 1.8 V			1	µA

7.6 Timing Requirements

PARAMETER		TEST CONDITIONS	MIN	NOM	MAX	UNIT
VBUS/BAT POWER UP						
t_{VBUS_OV}	VBUS OVP reaction time	VBUS rising above V_{BUS_OV} threshold to converter turn off		200		ns
$t_{POORSRC}$	Bad adapter detection duration			30		ms
BATTERY CHARGER						
t_{TERM_DGL}	Deglitch time for charge termination	Charge current falling below I_{TERM}		250		ms
t_{RECGH_DGL}	Deglitch time for recharge threshold	BAT voltage falling below VRECHG = 100 mV		250		ms
$t_{BAT\_OVP\_DGL}$	Deglitch time for battery over-voltage to disable charge			1		µs
t_{TOP_OFF}	Typical Top-Off Timer Accuracy	TOP_OFF_TIMER[1:0] = 30 min	24	30	36	min
t_{SAFETY}	Charge Safety Timer Accuracy	CHG_TIMER[1:0] = 12 hours	10.8	12	13.2	hr

Timing Requirements (continued)

PARAMETER		TEST CONDITIONS	MIN	NOM	MAX	UNIT
I2C INTERFACE						
f_{SCL}	SCL clock frequency				1000	kHZ
DIGITAL CLOCK AND WATCHDOG TIMER						
f_{LPDIG}	Digital low power clock	REGN LDO disabled	18	30	45	kHZ
f_{DIG}	Digital clock	REGN LDO enabled	1.35	1.5	1.65	MHz
t_{WDT}	Watchdog Reset time	WATCHDOG[1:0] = 160 s, REGN LDO disabled	100	160		sec
t_{WDT}	Watchdog Reset time	WATCHDOG[1:0] = 160 s, REGN LDO enabled	136	160		sec

7.7 Typical Characteristics

C_{VBUS} = 1µF, C_{PMID}= 10µF, C_{SYS} = 44µF, C_{BAT} = 10µF, L = 1uH (DFE252012F-1R0) (unless otherwise specified)

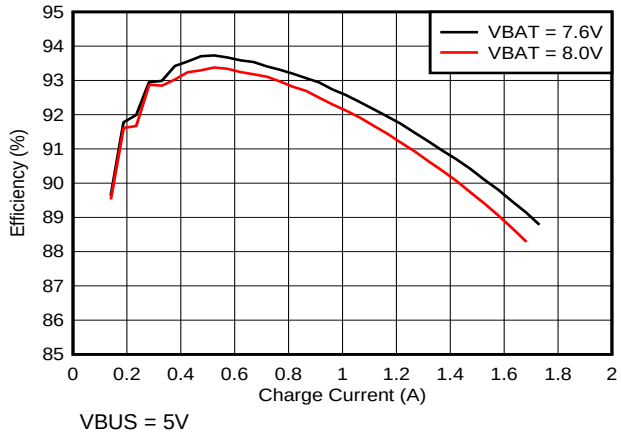


Figure 1. Charge Efficiency vs. Charge Current

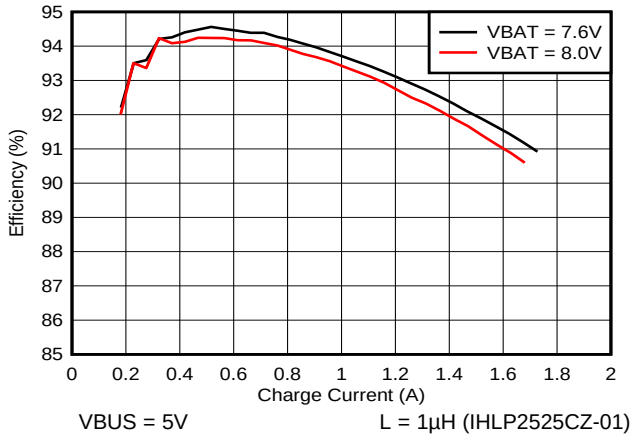


Figure 2. Charge Efficiency vs. Charge Current

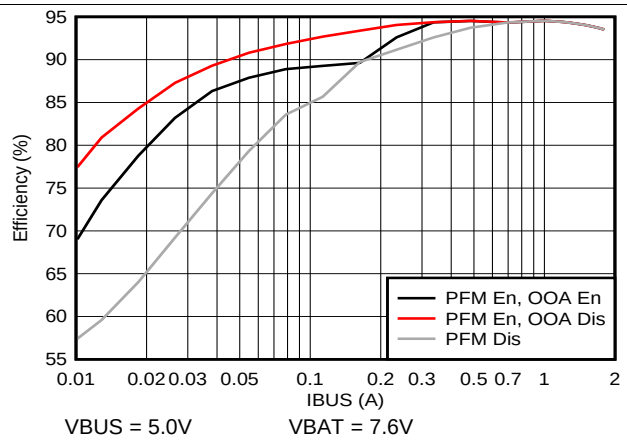


Figure 3. OTG Efficiency vs. VBUS Output Current

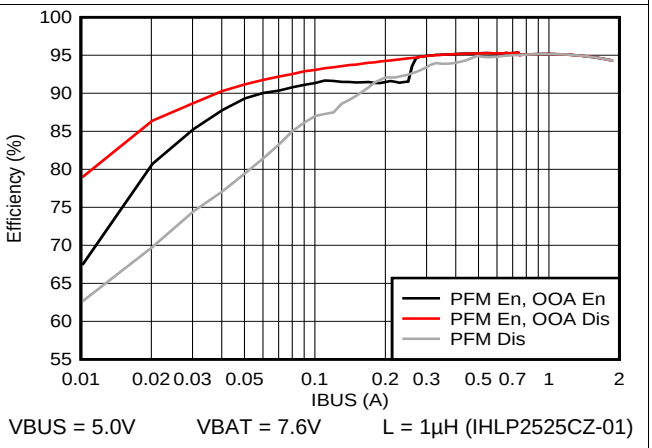


Figure 4. OTG Efficiency vs. VBUS Output Current

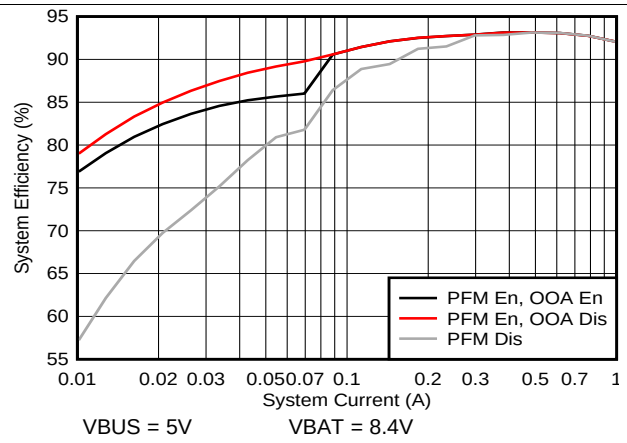


Figure 5. System Efficiency vs. System Current

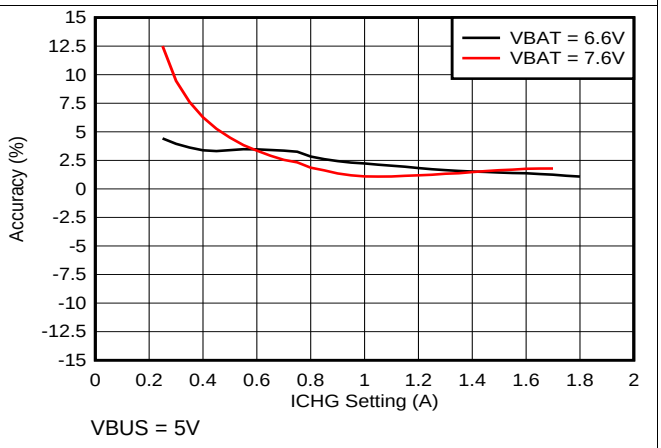


Figure 6. Charge Current Accuracy vs. ICHG Setting

Typical Characteristics (continued)

C_{VBUS} = 1µF, C_{PMID}= 10µF, C_{SYS} = 44µF, C_{BAT} = 10µF, L = 1uH (DFE252012F-1R0) (unless otherwise specified)

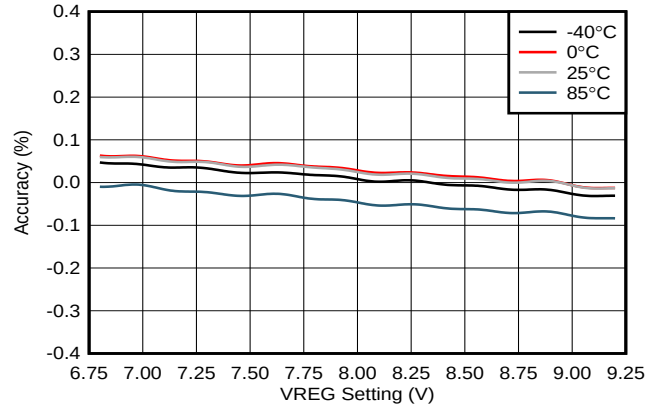


Figure 7. Battery Voltage Accuracy vs. VREG Setting

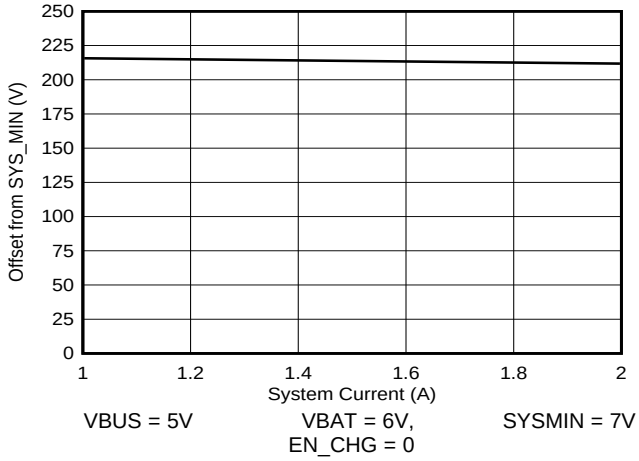


Figure 8. SYSMIN Load Regulation

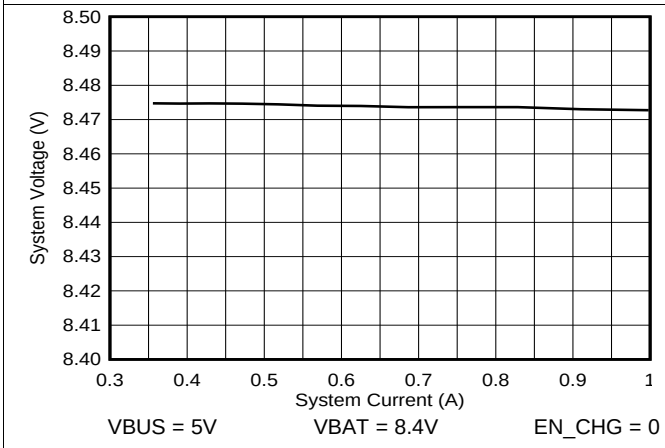


Figure 9. System Load Regulation After Charge Done

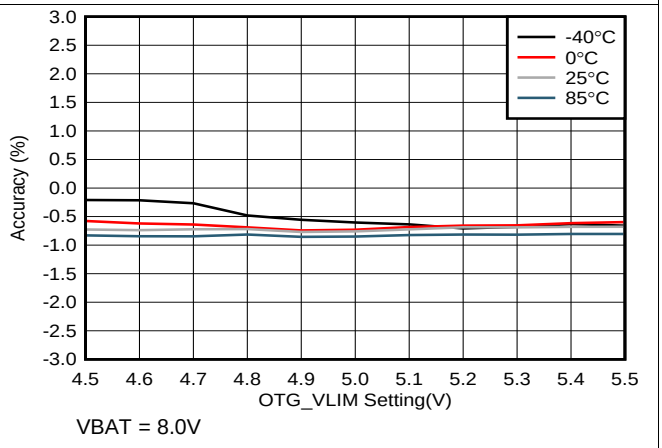


Figure 10. OTG VBUS Regulation vs. OTG_VLIM Setting

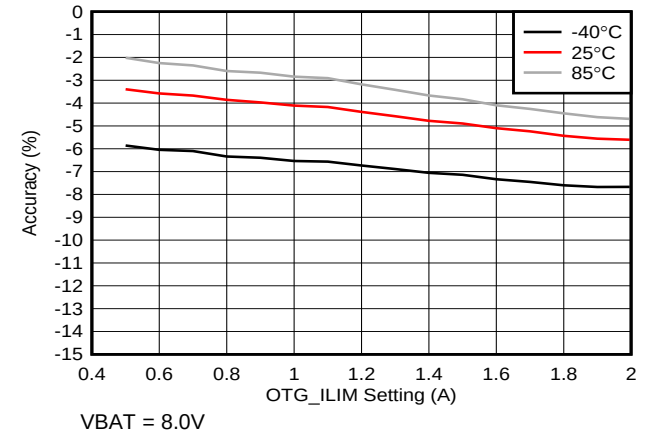


Figure 11. OTG IBUS Limit vs. OTG_ILIM Setting

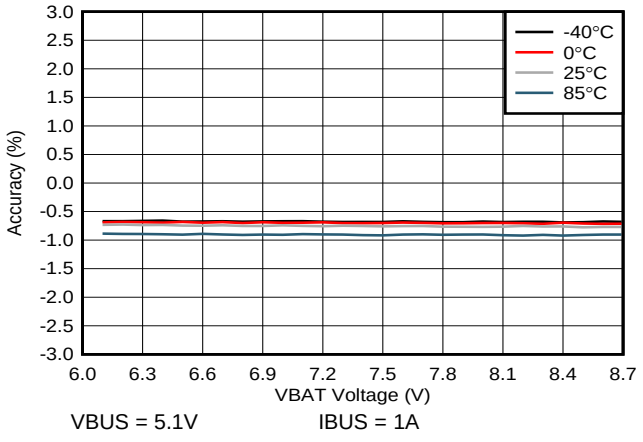
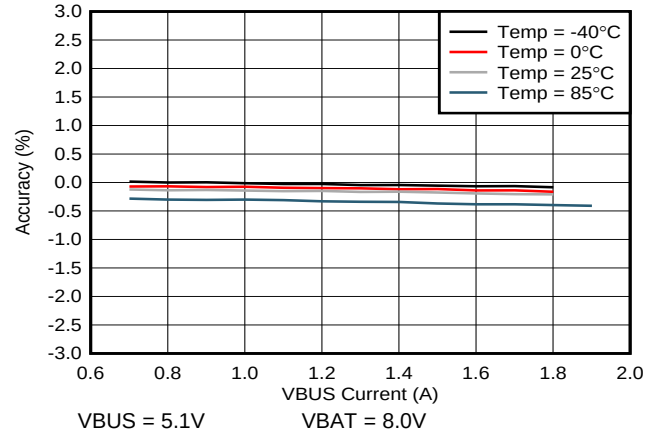


Figure 12. OTG Voltage Regulation vs. VBAT Voltage

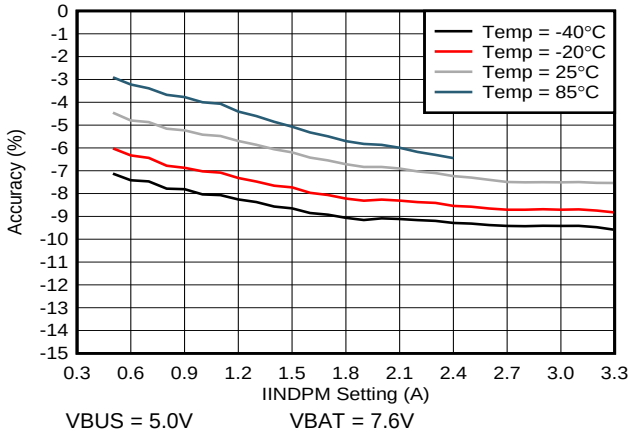
Typical Characteristics (continued)

C_{VBUS} = 1µF, C_{PMID}= 10µF, C_{SYS} = 44µF, C_{BAT} = 10µF, L = 1uH (DFE252012F-1R0) (unless otherwise specified)



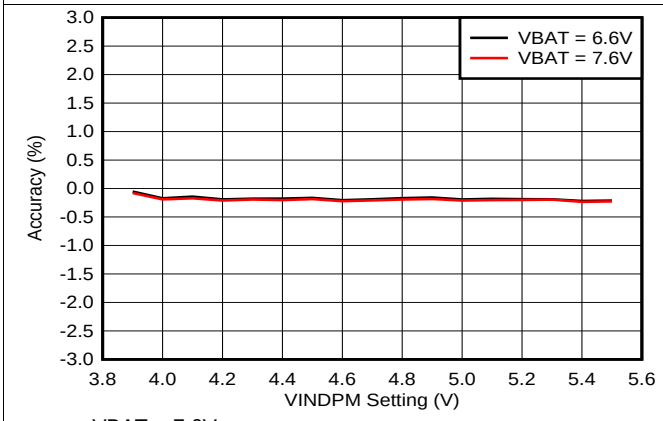
VBUS = 5.1V VBAT = 8.0V

Figure 13. OTG Voltage Regulation vs. OTG BUS Current



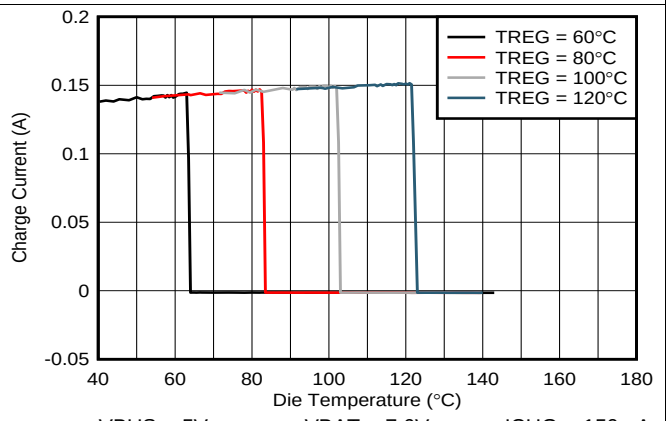
VBUS = 5.0V VBAT = 7.6V

Figure 14. Input Current Limit Accuracy vs. IINDPM Setting



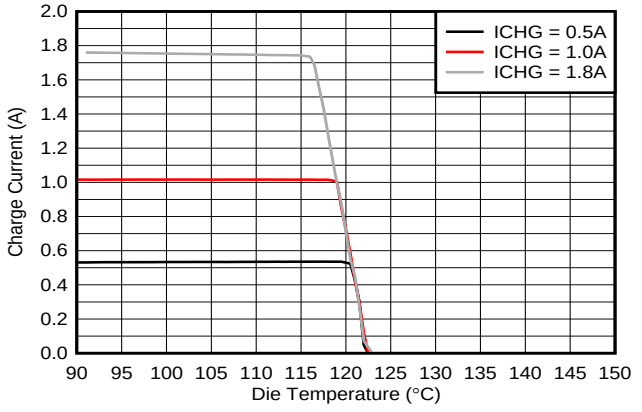
VBAT = 7.6V

Figure 15. Input Voltage Limit Accuracy vs. VINDPM Setting



VBUS = 5V VBAT = 7.6V ICHG = 150mA

Figure 16. TREG Profiles



VBUS = 5V VBAT = 7.6V ICHG = 1A, 1.5A, 2.2A

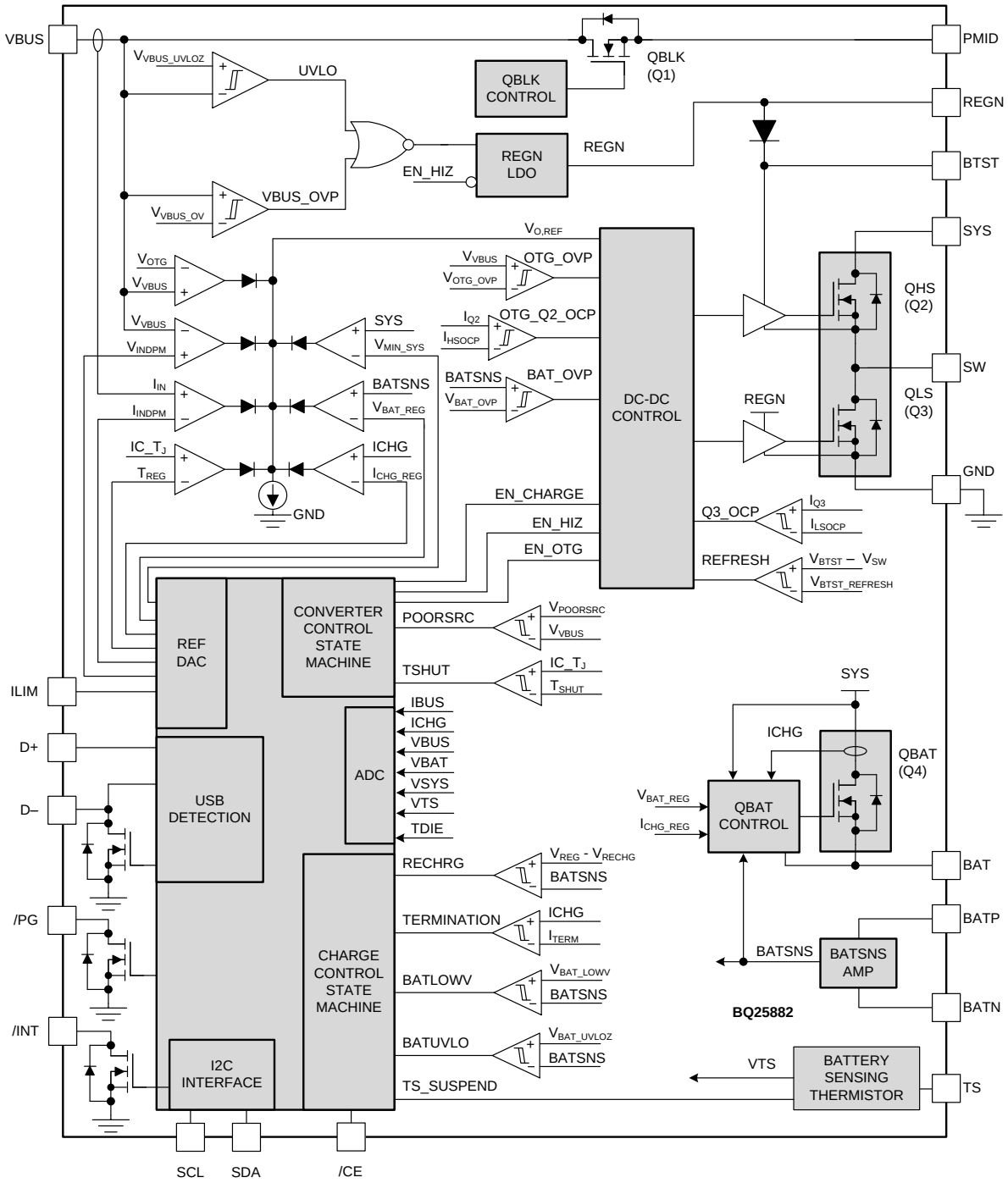
Figure 17. Max Current Temperature Profile

8 Detailed Description

8.1 Overview

The device is a highly integrated 2-A switch-mode battery charger for dual cell Li-Ion and Li-polymer battery. It integrates the input blocking FET (Q1, QBLK), high-side switching FET (Q2, QHS), low-side switching FET (Q3, QLS), and battery FET (Q4, QBAT). The device also integrates the boot-strap diode for high-side gate drive.

8.2 Functional Block Diagram



8.3 Feature Description

8.3.1 Device Power-On-Reset

The internal bias circuits are powered from either VBAT or VBUS. When VBUS rises above V_{VBUS_UVLOZ} or BAT rises above V_{BAT_UVLOZ}, the BATFET driver is active. I^2C interface is ready for communication and all the registers are reset to default value. The host can access all the registers after POR.

8.3.2 Device Power Up from Battery without Input Source

If only battery is present and the voltage is above UVLO threshold (V_{BAT_UVLOZ}), the BATFET turns on and connects battery to system. The REGN LDO stays off to minimize the quiescent current. The low $R_{DS(ON)}$ of BATFET and the low quiescent current on BAT minimize the conduction loss and maximize the battery run time. The device always monitors discharge current through BATFET (Supplement Mode).

8.3.3 Device Power Up from Input Source

When an input source is plugged in, the device checks the input source voltage to turn on REGN LDO and all the bias circuits. It detects and sets the input current limit before the boost converter is started. The power up sequence from input source is as listed:

1. Poor Source Qualification
2. Input Source Type Detection based on D+/D– to set default Input Current Limit (IINDPM) register and input source type
3. Power Up REGN LDO
4. Converter Power-up

8.3.3.1 Poor Source Qualification

After valid VBUS is plugged in, the device checks the current capability of the input source. The input source has to meet the following requirements in order to start the boost converter.

1. VBUS voltage below V_{VBUS_OVP}
2. VBUS voltage above $V_{POORSRC}$ when pulling $I_{POORSRC}$

If VBUS_OVP is detected (condition 1 above), the device automatically retries detection once the over-voltage fault goes away . If a poor source is detected (condition 2 above), the device repeats poor source qualification routine every 2 seconds. After 7 consecutive failures, the device sets EN_HIZ = 1, and goes to HIZ mode. The battery powers up the system when the device is in HIZ. Adapter re-plugin and/or EN_HIZ bit toggle is required to restart device operation. The EN_HIZ bit is cleared automatically when the adapter is plugged in.

8.3.3.2 Input Source Type Detection

After the input source is qualified, the charger device runs Input Source Type Detection when AUTO_INDET_EN bit is set.

The BQ25882 follows the USB Battery Charging Specification 1.2 (BC1.2) to detect input source (SDP/CDP/DCP) and non-standard adapter through USB D+/D- lines. After input source type detection, the following registers and pins are changed:

1. Input Current Limit (IINDPM) register is changed to set current limit
2. Input Voltage Limit (VINDPM) register is changed to set default limit (if EN_VINDPM_RST = 1, otherwise VINDPM value remains unchanged)
3. VBUS_STAT bits change to reflect the detected source
4. $\overline{\text{INT}}$ pin pulses to notify the host
5. $\overline{\text{PG}}$ pin is pulled LOW, and PG_STAT bit is set to '1'

After detection is completed, the host can over-write IINDPM or VINDPM registers to change the input current, or input voltage limit if needed. The charger input current is always limited by the lower of IINDPM register or ILIM pin at all-times regardless of Input Current Optimizer (ICO) setting.

Feature Description (continued)

When AUTO_INDET_EN is disabled, the Input Source Type Detection is bypassed, and the Input Current Limit (IINDPM) register remains unchanged from previous value. When EN_VINDPM_RST is disabled, the Input Voltage Limit (VINDPM) register remains unchanged from previous value.

8.3.3.2.1 D+/D– Detection Sets Input Current Limit

The BQ25882 contains a D+/D- based input source detection to program the input current limit. The D+/D- detection has three major steps: Data Contact Detect (DCD), Primary Detection, and Secondary Detection.

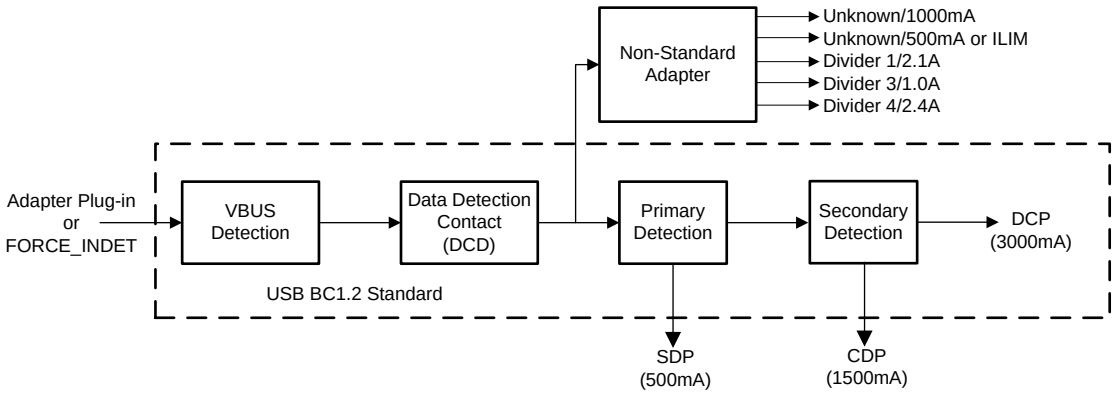


Figure 18. D+/D– Detection Flow

Table 2. Non-Standard Adapter Detection

NON-STANDARD ADAPTER	D+ THRESHOLD	D– THRESHOLD	INPUT CURRENT LIMIT
Divider 1	V_{D+} within V_{D+D-_2P8}	V_{D-} within V_{D+D-_2P0}	2.1 A
Divider 3	V_{D+} within V_{D+D-_2P0}	V_{D-} within V_{D+D-_2P8}	1 A
Divider 4	V_{D+} within V_{D+D-_2P8}	V_{D-} within V_{D+D-_2P8}	2.4 A
Unknown 2	VD+ = 1 MΩ to 0 V	VD- = 3.3 V	1.0 A

After the Input Source Type Detection is done, an INT pulse is asserted to the host. In addition, the following registers including Input Current Limit register (IINDPM), and VBUS_STAT are updated as below:

Table 3. Input Current Limit Setting from D+/D– Detection

D+/D– DETECTION	INPUT CURRENT LIMIT (IINDPM)	VBUS_STAT
USB SDP (USB500)	500 mA	001
USB CDP	1.5 A	010
USB DCP	3.0 A	011
Divider 3	1 A	110
Divider 1	2.1 A	110
Divider 4	2.4 A	110
Unknown 5-V Adapter (1)	500 mA	101
Unknown 5-V Adapter (2)	1000 mA	101

8.3.3.2.2 Force Input Current Limit Detection

In host mode, the host can force the device to run Input Current Limit Detection by setting FORCE_INDET bit. After the detection is completed, FORCE_INDET bit returns to 0 by itself and Input Result is updated. After the detection is completed, the input current limit (IINDPM), and the VBUS_STAT bits may be changed by the device due to the detection result.

8.3.3.3 Power up REGN Regulator (LDO)

The REGN LDO supplies internal bias circuits as well as the HSFET and LSFET gate drive. The LDO also provides bias rail to TS external resistors. The pull-up rail of /PG can be connected to REGN as well. The REGN is enabled when all the below conditions are valid.

1. VBUS above V_{VBUS_UVLOZ} in boost mode or VBUS below V_{VBUS_UVLOZ} in buck mode
2. Poor Source Qualification detects a valid input source
3. Input Source Type Detection completes and sets appropriate input current limit
4. After 220 ms delay is complete

If one of the above conditions is not valid, the device is in high impedance mode (HIZ) with REGN LDO off. The device draws less than IVBUS_HIZ from VBUS during HIZ state. The battery powers up the system when the device is in HIZ.

8.3.3.4 Converter Power Up

After the input current limit is set, the $\overline{PG}$ pin is pulled LOW, the PG_STAT and VBUS_STAT bits are changed, and the converter is enabled, allowing the HSFET and LSFET to start switching. If battery charging is disabled, BATFET turns off. Otherwise, BATFET stays on to charge the battery. The device provides soft-start when system rail is ramped up.

Before charging begins, the battery discharge source (IBAT_DISCHG) is enabled automatically to detect the presence of battery. The host can enable IBAT_DISCHG via the EN_BAT_DISCHG bit at any point during operation, including in Battery Only or HIZ modes.

As a battery charger, the device deploys a highly efficient 1.5-MHz step-up switching regulator. The fixed frequency oscillator keeps tight control of the switching frequency under all conditions of input voltage, battery voltage, charge current and temperature, simplifying filter design.

In order to improve light-load efficiency, the device switches to PFM control at light load when battery is below minimum system voltage setting or charging is disabled. During the PFM operation, the switching duty cycle is set by the ratio of SYS and VBUS. PFM operation may be disabled by the host using the PFM_DIS bit. PFM operation also includes an out-of-audio (OOA) feature to prevent the converter from switching within the audible range (< 20 kHz) at no load conditions. This feature may be disabled by the host using the PFM_OOA_DIS bit.

8.3.4 Input Current Optimizer (ICO)

The device provides innovative Input Current Optimizer (ICO) to identify maximum power point without overloading the input source. The algorithm automatically identifies maximum input current limit of a power source without staying in VINDPM to avoid input source overload.

This feature is enabled by default (EN_ICO=1) and can be disabled by setting EN_ICO bit to 0. After DCP type input source is detected based on the procedures describe above (Input Source Type Detection), the algorithm runs automatically when EN_ICO bit is set. The algorithm can also be forced to execute by setting FORCE_ICO bit regardless of input source type detected (EN_ICO = 1 is required for FORCE_ICO to work).

Table 4. Input Current Optimizer Automatic Operation

DEVICE	INPUT SOURCE	INPUT CURRENT LIMIT (IINDPM)	AUTOMATIC START ICO ALGORITHM WHEN EN_ICO = 1
BQ25882 (D+/D–)	USB SDP (USB500)	500mA	Disable
	USB CDP	1.5A	Disable
	USB DCP	3.0A	Enable
	Divider 3	1A	Disable
	Divider 1	2.1A	Disable
	Divider 4	2.4A	Disable
	Unknown 5V Adapter (1)	500mA	Disable
	Unknown 5V Adapter (2)	1000mA	Disable

The actual input current limit used by the Dynamic Power Management is reported in ICO_ILIM register while Input Current Optimizer is enabled (EN_ICO = 1) or set by IINDPM register when the algorithm is disabled (EN_ICO = 0). In addition, the current limit is clamped by ILIM pin unless EN_ILIM bit is 0 to disable ILIM pin function.

When the algorithm is enabled, it runs continuously to adjust input current limit of Dynamic Power Management (IINDPM) using ICO_ILIM register until ICO_STAT[1:0] and ICO_FLAG bits are set (the ICO_FLAG bit indicates any change in ICO_STAT[1:0] bits). The algorithm operates depending on battery voltage:

1. When battery voltage is below SYS_MIN, the algorithm starts ICO_ILIM register with IINDPM which is the maximum input current limit allowed by system
2. When battery voltage is above SYS_MIN, the algorithm starts ICO_ILIM register with 500 mA which is the minimum input current limit to minimize adapter overload

When optimal input current is identified, the ICO_STAT[1:0] and ICO_FLAG bits are set to indicate input current limit in ICO_ILIM register would not be changed until the algorithm is forced to run by the following event (these events also reset the ICO_STAT[1:0] bits to '01'):

1. A new input source is plugged-in, or EN_HIZ bit is toggled
2. IINDPM register is changed
3. VINDPM register is changed
4. FORCE_ICO bit is set to 1
5. VBUS_OVP event

8.3.5 Buck Mode Operation from Battery (OTG)

The device supports buck converter operation to deliver power from the battery to other portable devices through USB port. The buck mode output current rating meets the USB On-The-Go 500-mA (OTG_ILIM bits = 000) output requirement. The maximum output current is up to 2.0 A. The buck operation can be enabled if the following conditions are valid:

1. BAT above V_{OTG_BAT}
2. VBUS less than $V_{VBUS\_PRESENT}$
3. ADC is enabled
4. Buck mode operation is enabled (EN_OTG = 1)
5. Voltage at TS (thermistor) pin is within range configured by Buck Mode Temperature Monitor as configured by BHOT and BCOLD register bits

In buck mode, the device employs 1.5-MHz step-down switching regulator based on system requirements. During buck mode, the status register VBUS_STAT bits are set to 111, the VBUS output is 5.1 V by default (selectable via OTG_VLIM register bits) and the output current can reach up to 2.0 A, selected via I^2C (OTG_ILIM bits). The buck output is maintained when BAT is above V_{OTG_BAT} threshold, and VBUS is above $V_{VBUS\_PRESENT}$ threshold.

In order to improve light-load efficiency, the device switches to PFM control at light load. During the PFM operation, the switching duty cycle is set by the ratio of SYS and VBUS. PFM operation may be disabled by the host using the PFM_DIS bit. PFM operation also includes an out-of-audio (OOA) feature to prevent the converter from switching within the audible range (< 20 kHz) at no load conditions. This feature may be disabled by the host using the PFM_OOA_DIS bit.

8.3.6 Power Path Management

The device accommodates a wide range of input sources from USB, to wall adapter, to power bank. The device provides automatic power path selection to supply the system (SYS) from input source (VBUS), battery (BAT), or both.

8.3.6.1 Narrow VDC Architecture

The device deploys Narrow VDC architecture (NVDC) with BATFET separating system from battery. The minimum system voltage is set by SYS_MIN bits. Even with a fully depleted battery, the system is regulated above the minimum system voltage (default 7.0 V).

When the battery is below minimum system voltage setting, the BATFET operates in linear mode (LDO mode), and the system is typically 200 mV above the minimum system voltage setting. As the battery voltage rises above the minimum system voltage, BATFET is fully on and the voltage difference between the system and battery is the V_{DS} of BATFET.

When the battery charging is disabled and VBAT is above minimum system voltage setting or charging is terminated, the system is always regulated at typically 100 mV above battery voltage. The status register VSYS_STAT bit goes high when the system is in minimum system voltage regulation.

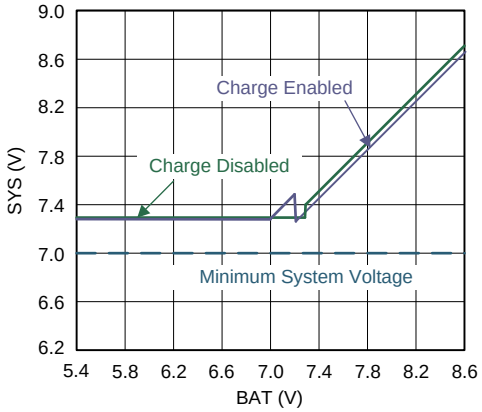


Figure 19. System Voltage vs. Battery Voltage

8.3.6.2 Dynamic Power Management

To meet maximum current limit in USB spec and avoid over loading the adapter, the device features Dynamic Power Management (DPM), which continuously monitors the input current and input voltage. When input source is over-loaded, either the current exceeds the input current limit (IINDPM or ICO_ILIM or ILIM pin setting) or the voltage falls below the input voltage limit (VINDPM). The device then reduces the charge current until the input current falls below the input current limit and the input voltage rises above the input voltage limit.

When the charge current is reduced to zero, but the input source is still overloaded, the system voltage starts to drop. Once the system voltage falls below the battery voltage, the device automatically enters the Supplement Mode where the BATFET turns on and battery starts discharging so that the system is supported from both the input source and battery.

During DPM mode, the status register bits VINDPM_STAT (VINDPM) and/or IINDPM_STAT (IINDPM) go high. The figure shows the DPM response with 5-V/3-A adapter, 6.4-V battery, 1.5-A charge current and 6.8-V minimum system voltage setting.

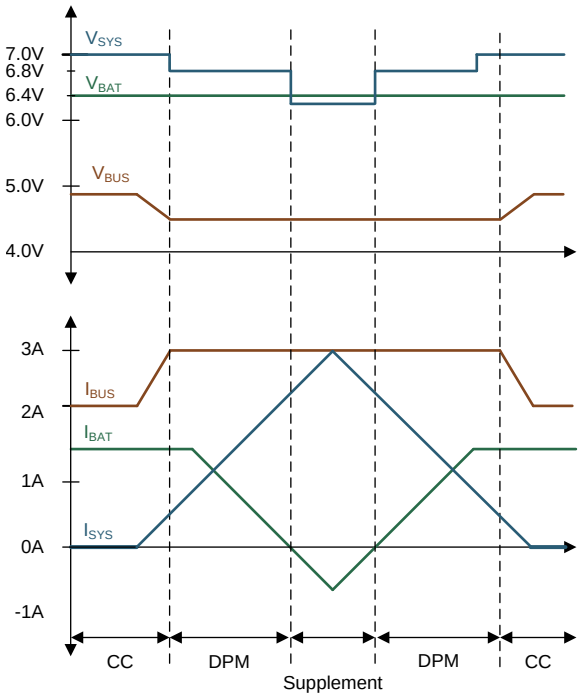


Figure 20. DPM Response

8.3.6.3 Supplement Mode

When the voltage falls below the battery voltage, the BATFET turns on.

As the discharge current increases, the BATFET gate is regulated with a higher voltage to reduce R_{DSON} until the BATFET is in full conduction. At this point onwards, the BATFET V_{DS} linearly increases with discharge current. The figure shows the V-I curve of the BATFET gate regulation operation. BATFET turns off to exit Supplement Mode when the battery is below battery depletion threshold (V_{BAT_DPL}).

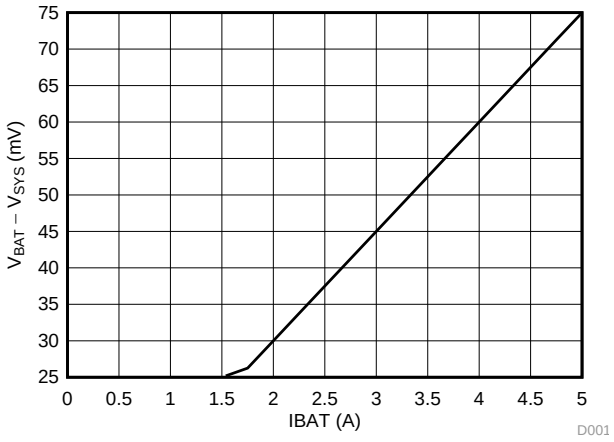


Figure 21. BATFET I-V Curve

8.3.7 Battery Charging Management

The device charges 2-cell Li-Ion battery with up to 2.2-A charge current for high capacity battery. The low $R_{DS(ON)}$ BATFET improves charging efficiency and minimize the voltage drop during discharging.

8.3.7.1 Autonomous Charging Cycle

When battery charging is enabled (EN_CHG bit =1 and /CE pin is LOW), the device autonomously completes a charging cycle without host involvement. The device default charging parameters are listed in the table below. The host can always control the charging operation and optimize the charging parameters by writing to the corresponding registers through I^2C.

Table 5. Charging Parameter Default Settings

DEFAULT MODE	BQ25882
Charging Voltage	8.4 V
Charging Current	1.00 A
Pre-Charge Current	150 mA
Termination Current	150 mA
Temperature Profile	JEITA
Safety Timer	12 hours

A new charge cycle starts when the following conditions are valid:

- Converter starts
- Battery charging is enabled by I^2C register bit (EN_CHG = 1 and $\overline{CE}$ pin is LOW and ICHG register is not 0 mA)
- No thermistor fault on TS
- No safety timer fault

The charger device automatically terminates the charging cycle when the charging current is below termination threshold, charge voltage is above recharge threshold, and device is not in DPM mode or thermal regulation. When a full battery voltage is discharged below recharge threshold (threshold selectable via VRECHG[1:0] bits), the device automatically starts a new charging cycle. After the charge is done, toggle either $\overline{CE}$ pin or EN_CHG bit can initiate a new charging cycle.

The status register (CHRG_STAT) indicates the different charging phases as:

- 000 – Not Charging
- 001 – Trickle Charge (VBAT < V_{BAT_SHORT})
- 010 – Pre-charge (V_{BAT_SHORT} < VBAT < V_{BAT_LOWV})
- 011 – Fast-charge (CC mode)
- 100 – Taper Charge (CV mode)
- 101 – Top-off Timer Active Charging
- 110 – Charge Termination Done

When the charger transitions to any of these states, including when charge cycle is completed, an INT is asserted to notify the host.

8.3.7.2 Battery Charging Profile

The device charges the battery in five phases: trickle charge, pre-charge, constant current, constant voltage, and top-off trickle charging (optional). At the beginning of a charging cycle, the device checks the battery voltage and regulates current/voltage accordingly.

Table 6. Default Charging Current Setting

VBAT	CHARGING CURRENT	REGISTER DEFAULT SETTING	CHRG_STAT
< V_{BAT_SHORT}	I_{BAT_SHORT}	100 mA	001
V_{BAT_SHORT} to V_{BAT_LOWV}	I_{PRECHG}	150 mA	010
> V_{BAT_LOWV}	ICHG	1500 mA	011

If the charger device is in DPM regulation or thermal regulation during charging, the actual charging current will be less than the programmed value. In this case, termination is temporarily disabled and the charging safety timer is counted at half the clock rate, as explained in the Charging Safety Timer section.

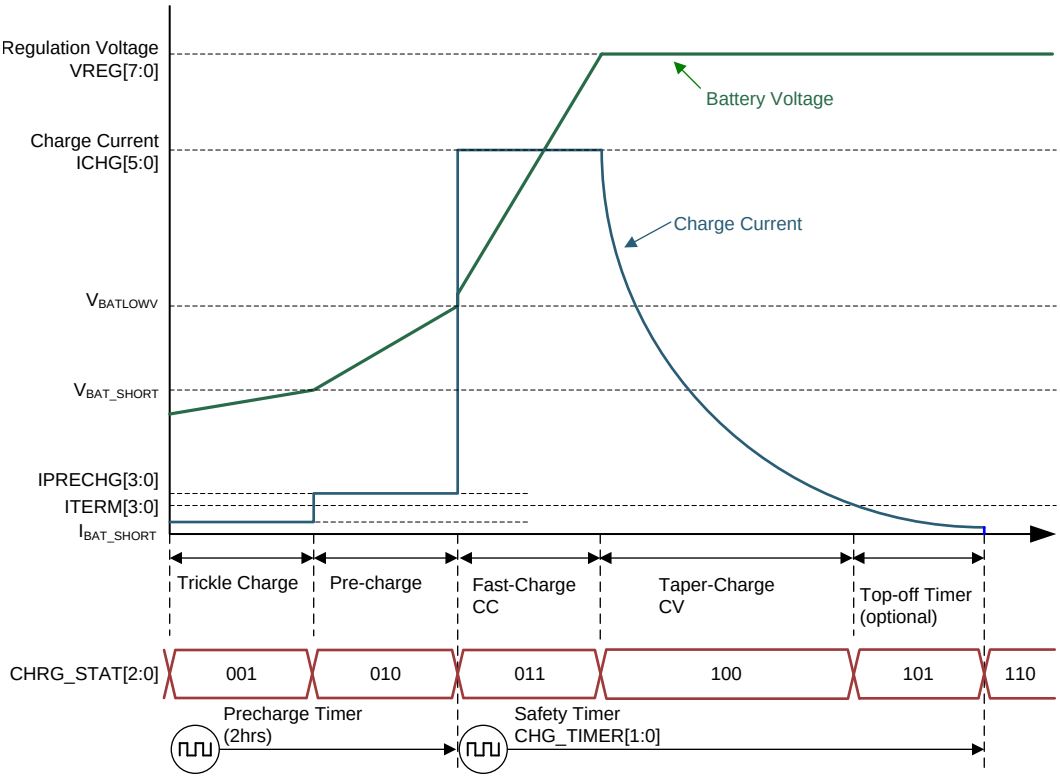


Figure 22. Battery Charging Profile

8.3.7.3 Charging Termination

The device terminates a charge cycle when the battery voltage is above recharge threshold, and the current is below termination current. After the charging cycle is completed, the BATFET turns off. The converter keeps running to power the system, and BATFET can turn on again to engage Supplement Mode.

When termination occurs, the status register CHRG_STAT is set to 110, and an INT pulse is asserted to the host. Termination is temporarily disabled when the charger device is in input current, voltage or thermal regulation. Termination can be permanently disabled by writing 0 to EN_TERM bit prior to charge termination.

At low termination currents (50 mA to100 mA), due to the comparator offset, the actual termination current may be up to 40% higher than the termination target. In order to compensate for comparator offset, a programmable top-off timer (default disabled) can be applied after termination is detected. The top-off timer will follow safety timer constraints, such that if safety timer is suspended, so will the top-off timer. Similarly, if safety timer is doubled, so will the top-off timer. CHRG_STAT reports whether the top off timer is active via the 101 code. Once the Top-Off timer expires, the CHRG_STAT register is set to 110 and an INT pulse is asserted to the host.

Top-off timer gets reset (set to 0 and counting resumes when appropriate) for any of the following conditions:

1. Charge disable to enable
2. Termination status low to high
3. REG_RST register bit is set (disables top-off timer)

The top-off timer settings are read in once termination is detected by the charger. Programming a top-off timer value after termination will have no effect until a new charge cycle is initiated. An INT is asserted to the host when entering top-off timer segment as well as when top-off timer expires. All charge cycle related INT pulses (including top-off timer INT pulses) can be masked by CHRG_MASK bit.

8.3.7.4 Thermistor Qualification

The charger device provides a single thermistor input for battery temperature monitor.

8.3.7.4.1 JEITA Guideline Compliance in Charge Mode

To improve the safety of charging Li-ion batteries, JEITA guideline was released on April 20, 2007. The guideline emphasized the importance of avoiding a high charge current and high charge voltage at certain low and high temperature ranges.

To initiate a charge cycle, the voltage on TS pin must be within the VT1 to VT5 thresholds. If TS voltage exceeds the T1-T5 range, the controller suspends charging and waits until the battery temperature is within the T1 to T5 range. At cool temperature (T1-T2), JEITA recommends the charge current to be reduced to half of the charge current or lower. At warm temperature (T3-T5), JEITA recommends charge voltage less than 4.1 V / cell.

The charger provides flexible voltage/current settings beyond the JEITA requirement. The voltage setting at warm temperature (T3-T5) can be VREG, 8.0 V, 8.3 V, or charge suspend (configured by JEITA_VSET [1:0]). The fast charge current setting at warm temperature (T3-T5) can be 100%, or 40% of fast charge current, ICHG (configured by JEITA_ISETH). The fast charge current setting at cool temperature (T1-T2) can be 100%, 40%, or 20% of fast charge current, ICHG, or charge suspend (configured by JEITA_ISETC[1:0]). Whenever the charger detects "warm" or "cool" temperature, TERMINATION is automatically disabled regardless of JEITA_VSET, JEITA_ISETH and JEITA_ISETC register bit settings.

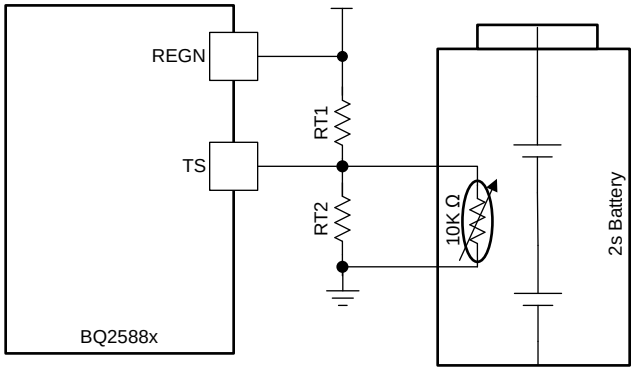


Figure 23. TS Resistor Network

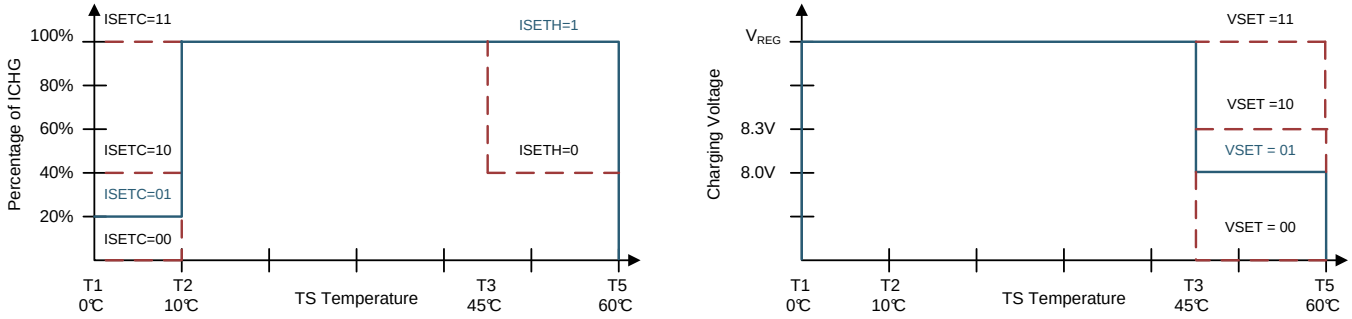


Figure 24. TS Charging Values

Assuming a 103AT NTC thermistor on the battery pack as shown above, the value of RT1 and RT2 can be determined by:

$$RT2 = \frac{V_{REGN} \times RTH_{COLD} \times RTH_{HOT} \times \left(\frac{1}{VT1} - \frac{1}{VT5}\right)}{RTH_{HOT} \times \left(\frac{V_{REGN}}{VT5} - 1\right) - RTH_{COLD} \times \left(\frac{V_{REGN}}{VT1} - 1\right)} \quad (1)$$

$$RT1 = \frac{\frac{V_{REGN}}{VT1} - 1}{\frac{1}{RT2} + \frac{1}{RTH_{COLD}}} \quad (2)$$

Select 0°C to 60°C range for Li-ion or Li-polymer battery:

RTH_{T1} = 27.28 kΩ

RTH_{T5} = 3.02 kΩ

RT1 = 5.24 kΩ

RT2 = 30.31 kΩ

8.3.7.4.2 Cold/Hot Temperature Window in OTG Buck Mode

For battery protection during OTG buck mode, the device monitors the battery temperature to be within the VBCOLD to VBHOT thresholds. When temperature is outside of the temperature thresholds, the OTG mode is suspended. In addition, VBUS_STAT bits are set to 000 and corresponding TS_STAT is reported. Once temperature returns within thresholds, the OTG mode is recovered and TS_STAT is cleared.

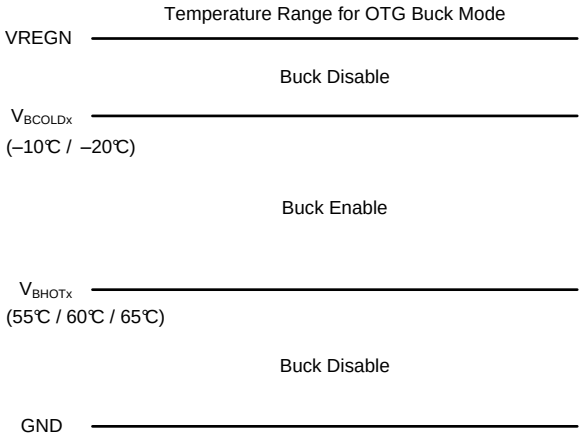


Figure 25. TS Pin Thermistor Sense Threshold in OTG Buck Mode

8.3.7.5 Charging Safety Timer

The device has built-in safety timer to prevent extended charging cycle due to abnormal battery conditions. The user can program fast charge safety timer through I^2C (CHG_TIMER bits). When safety timer expires, the fault register TMR_STAT bit is set to 1, and an INT pulse is asserted to the host. The safety timer feature can be disabled by clearing EN_TIMER bit.

During input voltage, current or thermal regulation, the safety timer counts at half clock rate as the actual charge current is likely to be below the register setting. For example, if the charger is in input current regulation (IINDPM_STAT=1) throughout the whole charging cycle, and the safety timer is set to 5 hours, then the timer will expire in 10 hours. This half clock rate feature can be disabled by setting TMR2X_EN = 0. Changing the TMR2X_EN bit while the device is running has no effect on the safety timer count, other than forcing the timer to count at half the rate under the conditions dictated above.

During faults which disable charging, or supplement mode, timer is suspended. Since the timer is not counting in this state, the TMR2X_EN bit has no effect. Once the fault goes away, safety timer resumes. If the charging cycle is stopped and started again, the timer gets reset (toggle $\overline{CE}$ pin or EN_CHG bit restarts the timer).

The safety timer is reset for the following events:

1. Charging cycle stop and restart (toggle $\overline{CE}$ pin, EN_CHG bit, or charged battery falls below recharge threshold)
2. BAT voltage changes from pre-charge to fast-charge or vice versa (in host-mode or default mode)

The precharge safety timer (fixed 2hr counter that runs when VBAT < V_{BAT_SHORT}), follows the same rules as the fast-charge safety timer in terms of getting suspended, reset, and counting at half-rate when TMR2X_EN is set.

8.3.8 Integrated 16-Bit ADC for Monitoring

The device includes a 16-bit ADC to monitor critical system information based on the device's modes of operation. The control of the ADC is done through the ADC Control Register (Address = 15h) [reset = 30h] register. The ADC_EN bit provides the ability to enable and disable the ADC to conserve power. The ADC_RATE bit allows continuous conversion or one-shot behavior. After a 1-shot conversion finishes, the ADC_EN bit is cleared, and must be re-asserted to start a new conversion.

To enable the ADC, the ADC_EN bit must be set to '1'. The ADC is allowed to operate if either the V_{VBUS}>V_{VBUS_UVLOZ} or VBAT>V_{BAT_UVLOZ} is valid. If no adapter is present, and the VBAT is less than V_{BAT_UVLOZ}, the device will not perform an ADC measurement, nor update the ADC read-back values in REG17 through REG24. Additionally, the device will immediately reset ADC_EN bit without sending any interrupt. The same will happen if the ADC is enabled when all ADC channels are disabled. It is recommended to read back ADC_EN after setting it to '1' to ensure ADC is running a conversion. If the charger changes mode (for example, if adapter is connected, EN_HIZ goes to '1', or EN_OTG goes to '1') while an ADC conversion is running, the conversion is interrupted. Once the mode change is complete, the ADC resumes conversion, starting with the channel where it was interrupted.

The ADC_SAMPLE bits control the sample speed of the ADC, with conversion time of t_{ADC_CONV}. The integrated ADC has two rate conversion options: a 1-shot mode and a continuous conversion mode set by the ADC_RATE bit. By default, all ADC parameters will be converted in 1-shot or continuous conversion mode unless disabled in the ADC Function Disable Register (Address = 16h) [reset = 00h]. If an ADC parameter is disabled by setting the corresponding bit in REG16, then the read-back value in the corresponding register will be from the last valid ADC conversion or the default POR value (all zeros if no conversions have taken place). If an ADC parameter is disabled in the middle of an ADC measurement cycle, the device will finish the conversion of that parameter, but will not convert the parameter starting the next conversion cycle. Even though no conversion takes place when all ADC measurement parameters are disabled, the ADC circuitry is active and ready to begin conversion as soon as one of the bits in the ADC Function Disable register is set to '0'. If all channels are disabled in 1-shot conversion mode, the ADC_EN bit is cleared.

The ADC_DONE_STAT and ADC_DONE_FLAG bits signal when a conversion is complete in 1-shot mode only. This event produces an INT pulse, which can be masked with ADC_DONE_MASK. During continuous conversion mode, the ADC_DONE_STAT bit has no meaning and will be '0'. The ADC_DONE_FLAG bit will remain unchanged in continuous conversion mode.

ADC conversion operates independently of the faults present in the device. ADC conversion will continue even after a fault has occurred (such as one that causes the power stage to be disabled), and the host must set ADC_EN = '0' to disable the ADC. ADC conversion is interrupted upon adapter plug-in, and will only resume until after Input Source Type Detection is complete. ADC readings are only valid for DC states and not for transients. When host writes ADC_EN = 0, the ADC stops immediately, and ADC measurement values correspond to last valid ADC reading.

If the host wants to exit ADC more gracefully, it is possible to do either of the following:

1. Write ADC_RATE to one-shot, and the ADC will stop at the end of a complete cycle of conversions, or
2. Disable all ADC conversion channels, and the ADC will stop at the end of the current measurement.

8.3.9 Status Outputs ($\overline{PG}$, and $\overline{INT}$)

8.3.9.1 Power Good Indicator ($\overline{PG}$)

The PG_STAT bit goes HIGH and the $\overline{PG}$ pin goes LOW to indicate a good input source when:

1. VBUS above V_{VBUS_UVLOZ}
2. VBUS below V_{VBUS_OV} threshold
3. VBUS above $V_{POORSRC}$ when $I_{POORSRC}$ current is applied (not a poor source)
4. Input Source Type Detection is completed

8.3.9.2 Interrupt to Host ($\overline{INT}$)

In some applications, the host does not always monitor the charger operation. The $\overline{INT}$ pin notifies the system host on the device operation. By default, the following events will generate an active-low, 256-µs INT pulse.

1. Good input source detected
   - $V_{VBUS} < V_{VBUS\_OV}$ threshold
   - $V_{VBUS} > V_{POORSRC}$ when $I_{POORSRC}$ current is applied (not a poor source)
2. VBUS_STAT changes state (VBUS_STAT any bit change)
3. Good input source removed
4. Entering IINDPM regulation
5. Entering VINDPM regulation
6. Entering IC junction temperature regulation (TREG)
7. I^2C Watchdog timer expired
   - At initial power up, this INT gets asserted to signal I^2C is ready for communication
8. Charger status changes state (CHRG_STAT value change), including Charge Complete
9. TS_STAT changes state (TS_STAT any bit change)
10. VBUS overvoltage detected (VBUS_OVP)
11. Junction temperature shutdown (TSHUT)
12. Battery overvoltage detected (BATOVP)
13. Charge safety timer expired
14. A rising edge on any of the *_STAT bits

Each one of these INT sources can be masked off to prevent INT pulses from being sent out when they occur. Three bits exist for each one of these events:

- The STAT bit holds the *current status* of each INT source
- The FLAG bit holds information on which source produced an INT, regardless of the current status
- The MASK bit is used to prevent the device from sending out INT for each particular event

When one of the above conditions occurs (a rising edge on any of the *_STAT bits), the device sends out an INT pulse and keeps track of which source generated the INT via the FLAG registers. The FLAG register bits are automatically reset to zero after the host reads them, and a new edge on STAT bit is required to re-assert the FLAG.

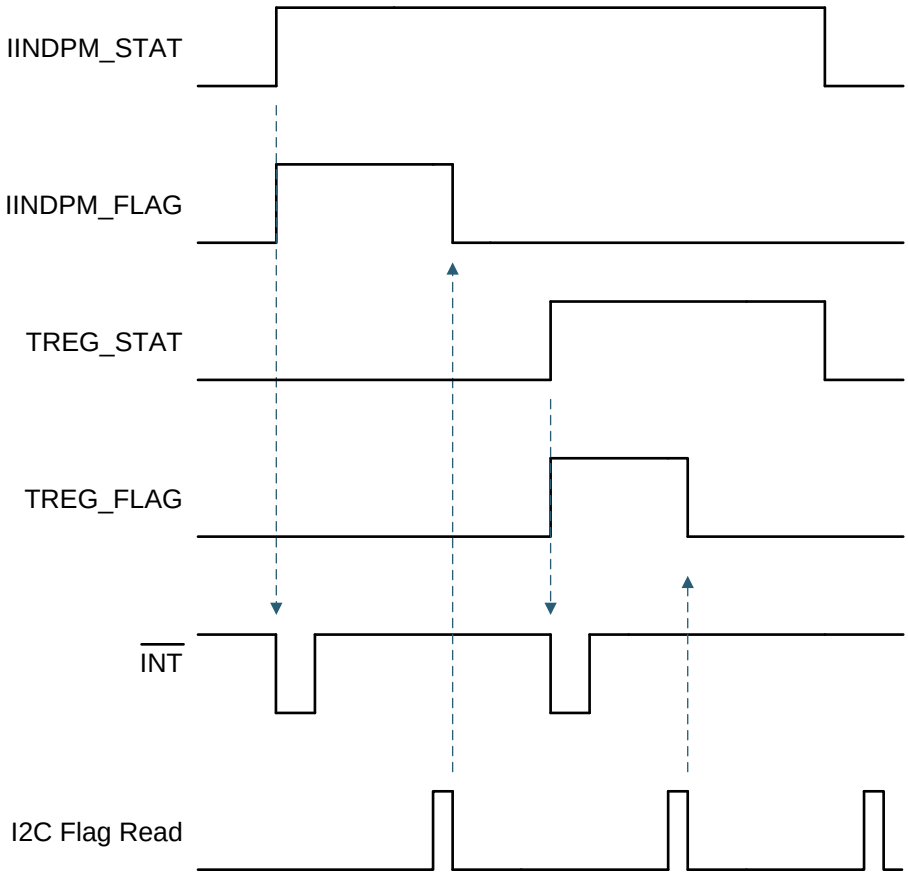


Figure 26. INT Generation Behavior Example

8.3.10 Input Current Limit on ILIM Pin

For safe operation, the device has an additional hardware pin on ILIM to limit maximum input current. The maximum input current is set by a resistor from ILIM pin to ground as:

$$I_{INMAX} = \frac{K_{ILIM}}{R_{ILIM}} \quad (3)$$

The actual input current limit is the lower value between ILIM pin setting and register setting (IINDPM). For example, if the register setting is 3.3 A (0x1C), and ILIM has a 820-Ω resistor to ground, the input current limit is 1.43 A (K_{ILIM} = 1170 max). ILIM pin can be used to set the input current limit rather than the register settings when EN_ILIM bit is set. The device regulates ILIM pin at 0.8 V. If ILIM voltage exceeds 0.8 V, the device enters input current regulation (refer to Dynamic Power Management section). Entering IINDPM through ILIM pin sets the IINDPM_STAT and FLAG bits, and produces an interrupt to host. The interrupt can be masked via the IINDPM_MASK bit.

The ILIM pin can also be used to monitor input current when EN_ILIM is set and the device is not in ILIM regulation. The voltage on ILIM pin is proportional to the input current. ILIM can be used to monitor input current with the following relationship:

$$I_{IN} = \frac{K_{ILIM} \times V_{ILIM}}{R_{ILIM} \times 0.8V} \quad (4)$$

For example, if ILIM pin is set with 820-Ω resistor, and the ILIM voltage 0.5 V, the actual input current is 0.762 A to 0.892 A (based on KILIM specified). If ILIM pin is open, the input current is limited to zero since ILIM voltage floats above 0.8 V. If ILIM pin is shorted, the input current limit is set by the register.

The ILIM pin function can be disabled by setting the EN_ILIM bit to 0. When the pin is disabled, both input current limit function and monitoring function are not available.

8.3.11 Voltage and Current Monitoring

The device closely monitors the input and system voltage, as well as internal FET currents for safe boost and buck mode operation.

8.3.11.1 Voltage and Current Monitoring in Boost Mode

8.3.11.1.1 Input Over-voltage Protection

The valid input voltage range for boost mode operation is V_{VBUS_OP}. If VBUS voltage exceeds V_{VBUS_OV}, the device stops switching immediately to protect the power FETs. During input over-voltage, an INT pulse is asserted to signal the host, and the VBUS_OVP_STAT and VBUS_OVP_FLAG fault registers get set. The device automatically starts switching again when the over-voltage condition goes away.

8.3.11.1.2 Input Under-Voltage Protection

The valid input voltage range for boost mode operation is V_{VBUS_OP}. If VBUS voltage falls below $V_{POORSRC}$ during operation, the device stops switching. During input under-voltage, an INT pulse is asserted to signal the host, and the PG_STAT bit gets cleared. The PG_FLAG bit will get set to signal this event. The device automatically attempts to restart switching when the under-voltage condition goes away.

8.3.11.1.3 System Over-Voltage Protection

The charger device clamps the system voltage during load transient so that the components connect to system would not be damaged due to high voltage. SYSOVP threshold is 350 mV above system regulation voltage. Upon SYSOVP, converter stops immediately to clamp the overshoot.

8.3.11.1.4 System Over-Current Protection

The charger device continually monitors and compares VBUS to VSYS to protect against a system short-circuit event. In the event that VSYS drops to within 250 mV of VBUS during operation, and the input current exceeds IINDPM threshold, a short circuit event is flagged and the converter stops switching. The SYS_SHORT_FLAG bit is set and an INT pulse is asserted to the host. The device attempts to recover from this condition automatically.

8.3.11.2 Voltage and Current Monitoring in OTG Buck Mode

The device closely monitors the VBUS voltage, as well as RBFET (Q1, QBLK) and LSFET (Q3, QLS) current to ensure safe buck mode operation.

8.3.11.2.1 VBUS Over-Voltage Protection

When the VBUS voltage rises above regulation target and exceeds V_{OTG_OVP}, the device enters over-voltage protection which stops switching, clears the EN_OTG bit and exits buck mode. During the over-voltage duration, the OTG_STAT and OTG_FLAG bits are set high to indicate a fault in buck mode operation. An INT is also asserted to the host.

8.3.11.2.2 VBUS Over-Current Protection

The device monitors output current to provide output short protection. The OTG buck mode has built-in constant current regulation to allow OTG to adapt to various types of loads. If short circuit is detected on VBUS, the OTG turns off and OTG is disabled with EN_OTG bit cleared. In addition OTG_STAT and OTG_FLAG bits are set high to indicate the fault, and an INT is asserted to the host.

8.3.12 Thermal Regulation and Thermal Shutdown

8.3.12.1 Thermal Protection in Boost Mode

The device monitors internal junction temperature, T_J, to avoid overheating and limits the IC surface temperature in boost mode. When the internal junction temperature exceeds the preset thermal regulation limit (TREG bits), the device reduces charge current. A wide thermal regulation range from 60°C to 120°C allows optimization for the system thermal performance.

During thermal regulation, the actual charging current is usually below the programmed value in ICHG registers. Therefore, termination is disabled, the safety timer runs at half the clock rate, the status register TREG_STAT bit goes high, and an INT is asserted to the host.

Additionally, the device has thermal shutdown to turn off the converter when IC surface temperature exceeds T_{SHUT}. The fault register bits TSHUT_STAT and TSHUT_FLAG are set and an INT pulse is asserted to the host. The converter turns back on when IC temperature is below T_{SHUT_HYS}.

8.3.12.2 Thermal Protection in OTG Buck Mode

The device monitors the internal junction temperature to provide thermal shutdown during OTG buck mode. When IC surface temperature exceeds T_{SHUT}, the buck mode is disabled (converter is turned off) by setting EN_OTG bit low. The fault register bits TSHUT_STAT and TSHUT_FLAG are set and an INT pulse is asserted to the host. When IC surface temperature is below T_{SHUT_HYS} (typ. 30°C), the host can set EN_OTG bit to '1' to recover.

8.3.13 Battery Protection

8.3.13.1 Battery Overvoltage Protection (BATOVP)

The battery over-voltage limit is clamped at 4% above the battery regulation voltage while charging. When battery overvoltage occurs, the charger device immediately disables charge. The fault register BATOVP_STAT bit goes high and an INT pulse is asserted to signal the host.

8.3.13.2 Battery Over-Discharge Protection

When the battery is discharged below V_{BAT_UVLOZ}, the BATFET is turned off to protect battery from over-discharge. To recover from over-discharge, an input source is required at VBUS. When an input source is plugged in, the BATFET turns on. The battery is charged with I_{BAT_SHORT} current when the VBAT < V_{BAT_SHORT}, or pre-charge current as set in IPRECHG registers when the battery voltage is between V_{BAT_SHORT} and V_{BAT_LOWV}.

8.3.14 Serial Interface

The device uses I^2C compatible interface for flexible charging parameter programming and instantaneous device status reporting. I^2C is a bi-directional 2-wire serial interface. Only two open-drain bus lines are required: a serial data line (SDA), and a serial clock line (SCL). Devices can be considered as masters or slaves when performing data transfers. A master is a device which initiates a data transfer on the bus and generates the clock signals to permit that transfer. At that time, any device addressed is considered a slave.

The device operates as a slave device with address 0x6B, receiving control inputs from the master device like micro-controller or digital signal processor through REG00 – REG25. Register read beyond REG25 (0x25), returns 0xFF. The I^2C interface supports both standard mode (up to 100 kbits/s), and fast mode (up to 400 kbits/s). When the bus is free, both lines are HIGH. The SDA and SCL pins are open drain and must be connected to the positive supply voltage via a current source or pull-up resistor.

8.3.14.1 Data Validity

The data on the SDA line must be stable during the HIGH period of the clock. The HIGH or LOW state of the data line can only change when the clock signal on SCL line is LOW. One clock pulse is generated for each data bit transferred.

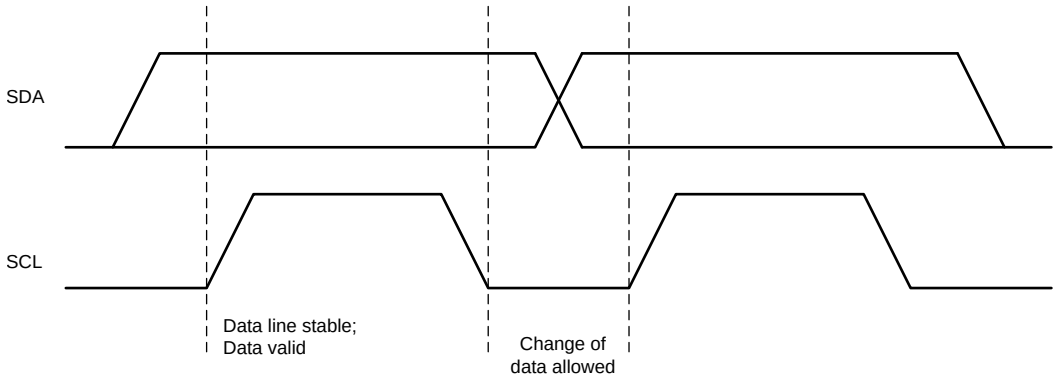


Figure 27. Bit Transfers on the I^2C bus

8.3.14.2 START and STOP Conditions

All transactions begin with a START (S) and are terminated with a STOP (P). A HIGH to LOW transition on the SDA line while SCL is HIGH defines a START condition. A LOW to HIGH transition on the SDA line when the SCL is HIGH defines a STOP condition.

START and STOP conditions are always generated by the master. The bus is considered busy after the START condition, and free after the STOP condition.

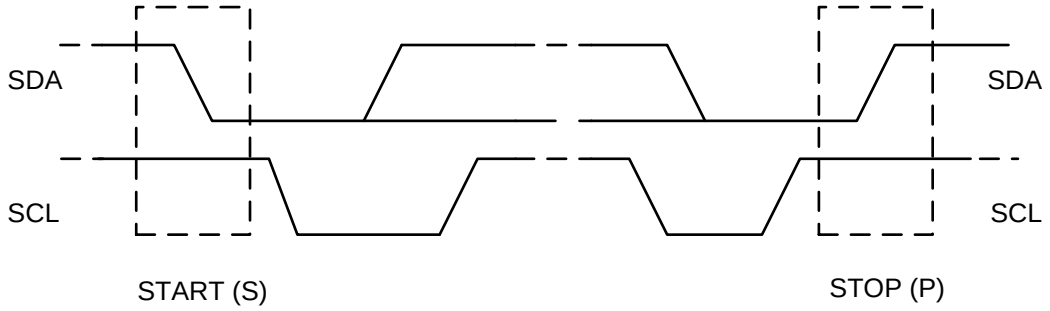


Figure 28. START and STOP Conditions on the I²C Bus

8.3.14.3 Byte Format

Every byte on the SDA line must be 8 bits long. The number of bytes to be transmitted per transfer is unrestricted. Each byte has to be followed by an ACKNOWLEDGE (ACK) bit. Data is transferred with the Most Significant Bit (MSB) first. If a slave cannot receive or transmit another complete byte of data until it has performed some other function, it can hold the SCL line low to force the master into a wait state (clock stretching). Data transfer then continues when the slave is ready for another byte of data and releases the SCL line.

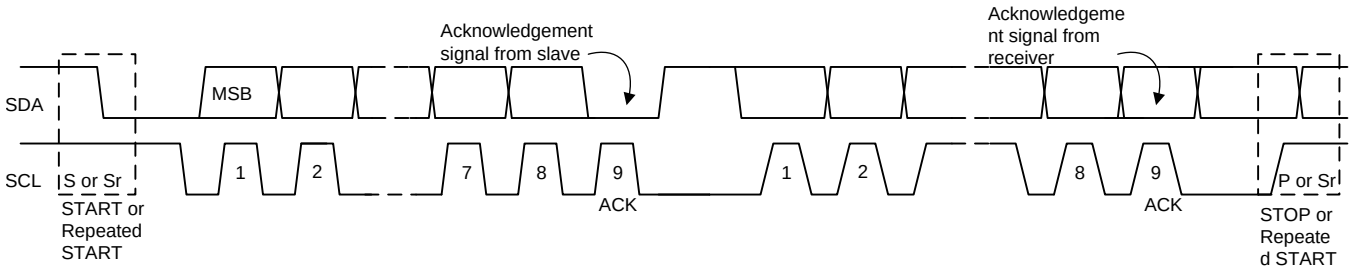


Figure 29. Data Transfer on the I²C Bus

8.3.14.4 Acknowledge (ACK) and Not Acknowledge (NACK)

The ACK signaling takes place after byte. The ACK bit allows the receiver to signal the transmitter that the byte was successfully received and another byte may be sent. All clock pulses, including the acknowledge 9th clock pulse, are generated by the master.

The transmitter releases the SDA line during the acknowledge clock pulse so the receiver can pull the SDA line LOW and it remains stable LOW during the HIGH period of this 9th clock pulse.

A NACK is signaled when the SDA line remains HIGH during the 9th clock pulse. The master can then generate either a STOP to abort the transfer or a repeated START to start a new transfer.

8.3.14.5 Slave Address and Data Direction Bit

After the START signal, a slave address is sent. This address is 7 bits long, followed by the 8 bit as a data direction bit (bit R/$\overline{W}$). A zero indicates a transmission (WRITE) and a one indicates a request for data (READ). The device 7-bit address is defined as 1101 011' (0x6B) by default. The address bit arrangement is shown below.

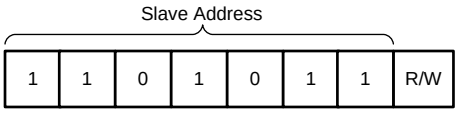


Figure 30. 7-Bit Addressing (0x6B)

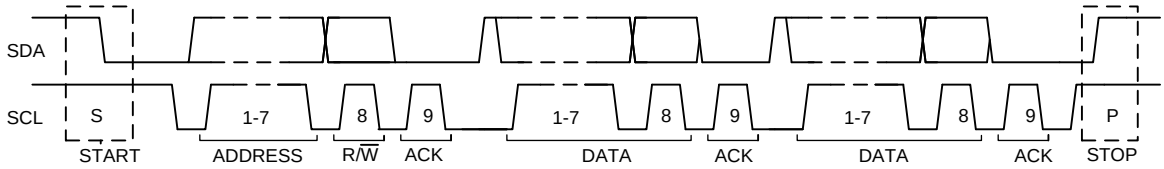


Figure 31. Complete Data Transfer on the I²C Bus

8.3.14.6 Single Write and Read

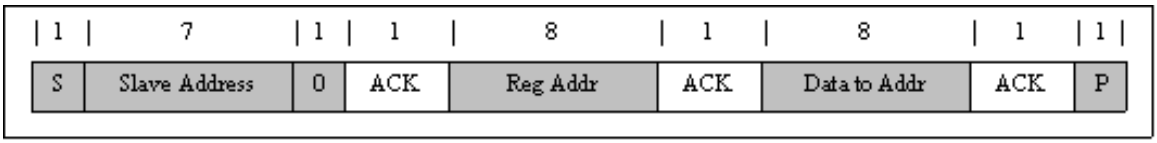


Figure 32. Single Write

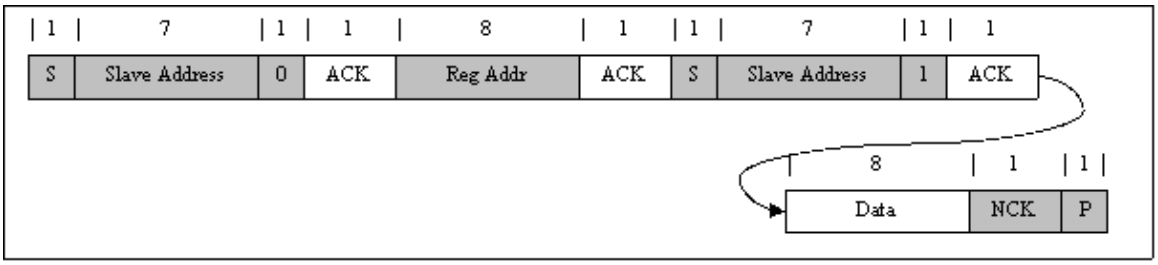


Figure 33. Single Read

If the register address is not defined, the charger IC sends back NACK and returns to the idle state.

8.3.14.7 Multi-Write and Multi-Read

The charger device supports multi-read and multi-write of all registers.

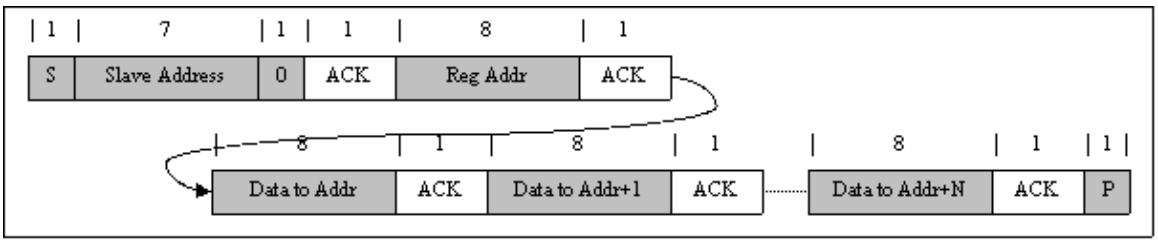


Figure 34. Multi-Write

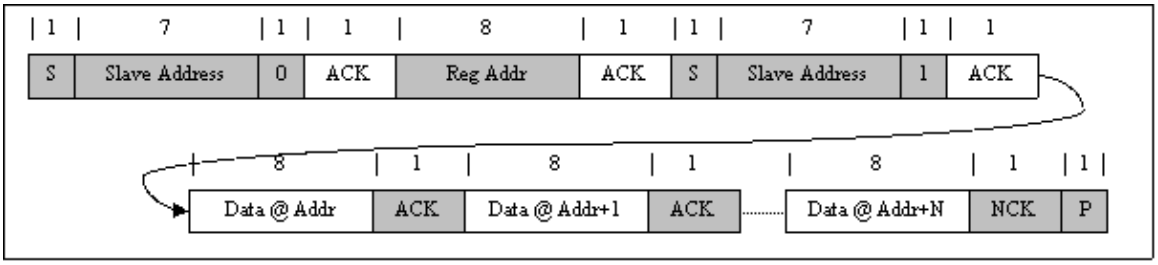


Figure 35. Multi-Read

8.4 Device Functional Modes

8.4.1 Host Mode and Default Mode

The device is a host controlled charger, but it can operate in default mode without host management. In default mode, the device can be used as an autonomous charger with no host or while host is in sleep mode. When the charger is in default mode, WD_STAT bit is HIGH. When the charger is in host mode, WD_STAT bit is LOW.

After power-on-reset, the device starts in default mode with watchdog timer expired, or default mode. All the registers are in the default settings.

In default mode, the device keeps charging the battery with default 10-hour fast charging safety timer. At the end of the 10-hour, the charging is stopped and the boost converter continues to operate to supply system load.

A write to any I^2C register transitions the charger from default mode to host mode, and initiates the watchdog timer. All the device parameters can be programmed by the host. To keep the device in host mode, the host has to reset the watchdog timer by writing 1 to WD_RST bit before the watchdog timer expires (WD_STAT bit is set), or disable watchdog timer by setting WATCHDOG bits = 00.

When the watchdog timer (WD_STAT bit = 1) is expired, the device returns to default mode and all registers are reset to default values except as detailed in the Register Maps section.

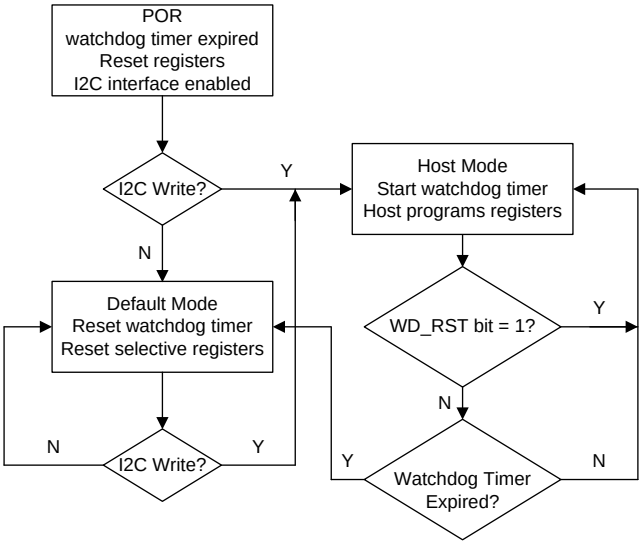


Figure 36. Watchdog Timer Flow Chart

8.5 Register Maps

Default I²C Slave Address: 0x6B (1101 011B + R/$\overline{W}$)

Table 7. I²C Registers

Address	Access Type	Acronym	Register Name	Section
00h	R/W	REG00	Battery Voltage Limit	Go
01h	R/W	REG01	Charge Current Limit	Go
02h	R/W	REG02	Input Voltage Limit	Go
03h	R/W	REG03	Input Current Limit	Go
04h	R/W	REG04	Precharge and Termination Control	Go
05h	R/W	REG05	Charger Control 1	Go
06h	R/W	REG06	Charger Control 2	Go
07h	R/W	REG07	Charger Control 3	Go
08h	R/W	REG08	Charger Control 4	Go
09h	R/W	REG09	OTG Control	Go
0Ah	R	REG0A	ICO Current Limit	Go
0Bh	R	REG0h	Charger Status 1	Go
0Ch	R	REG0C	Charger Status 2	Go
0Dh	R	REG0D	NTC Status	Go
0Eh	R	REG0E	FAULT Status	Go
0Fh	R	REG0F	Charger Flag 1	Go
10h	R	REG10	Charger Flag 2	Go
11h	R	REG11	Fault Flag	Go
12h	R/W	REG12	Charger Mask 1	Go
13h	R/W	REG13	Charger Mask 2	Go
14h	R/W	REG14	Fault Mask	Go
15h	R/W	REG15	ADC Control	Go
16h	R/W	REG16	ADC Function Disable	Go
17h	R	REG17	IBUS ADC1	Go
18h	R	REG18	IBUS ADC0	Go
19h	R	REG19	ICHG ADC1	Go
1Ah	R	REG1A	ICHG ADC0	Go
1Bh	R	REG1B	VBUS ADC1	Go
1Ch	R	REG1C	VBUS ADC0	Go
1Dh	R	REG1D	VBAT ADC1	Go
1Eh	R	REG1E	VBAT ADC0	Go
1Fh	R	REG1F	VSYS ADC1	Go
20h	R	REG20	VSYS ADC0	Go
21h	R	REG21	TS ADC1	Go
22h	R	REG22	TS ADC0	Go
23h	R	REG23	TDIE ADC1	Go
24h	R	REG24	TDIE ADC0	Go
25h	R/W	REG25	Part Information	Go

Complex bit access types are encoded to fit into small table cells. Table 8 shows the codes that are used for access types in this section.

Table 8. I²C Access Type Codes

ACCESS TYPE	CODE	DESCRIPTION
Read Type		
R	R	Read
Write Type		
W	W	Write
Reset Value		
-n		Value after reset
-X		Undefined value

8.5.1 Battery Voltage Regulation Limit Register (Address = 00h) [reset = A0h]

REG00 is shown in Figure 37 and described in Table 9.

Return to Summary Table.

Figure 37. REG00 Register

Bit	7	6	5	4	3	2	1	0
Reset	1	0	1	0	0	0	0	0
Field	VREG[7:0]							

Table 9. REG00 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	VREG[7]	R/W	Yes	Yes	1280 mV	Charge voltage limit Offset: 6.80 V Range: 6.80 V to 9.20 V Default 8.40 V
6	VREG[6]	R/W	Yes	Yes	640 mV	
5	VREG[5]	R/W	Yes	Yes	320 mV	
4	VREG[4]	R/W	Yes	Yes	160 mV	
3	VREG[3]	R/W	Yes	Yes	80 mV	
2	VREG[2]	R/W	Yes	Yes	40 mV	
1	VREG[1]	R/W	Yes	Yes	20 mV	
0	VREG[0]	R/W	Yes	Yes	10 mV	

8.5.2 Charger Current Limit Register (Address = 01h) [reset = 54h]

REG01 is shown in Figure 38 and described in Table 10.

Return to Summary Table.

Figure 38. REG01 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	1	0	1	0	1	0	0
Field	EN_HIZ	EN_ILIM	ICHG[5:0]					

Table 10. REG01 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	EN_HIZ	R/W	Yes	Yes	Enable HIZ Mode: 0 – Disable (default) 1 – Enable	
6	EN_ILIM	R/W	Yes	Yes	Enable ILIM Pin Function: 0 – Disable 1 – Enable (default)	
5	ICHG[5]	R/W	Yes	Yes	1600 mA	Fast Charge Current Limit Offset: 0 mA Range: 100 mA – 2200 mA Default: 1000 mA Note: ICHG > 2.2 A (2Ch) clamped to 2.2 A. ICHG < 100 mA (01h) clamped at 100 mA
4	ICHG[4]	R/W	Yes	Yes	800 mA	
3	ICHG[3]	R/W	Yes	Yes	400 mA	
2	ICHG[2]	R/W	Yes	Yes	200 mA	
1	ICHG[1]	R/W	Yes	Yes	100 mA	
0	ICHG[0]	R/W	Yes	Yes	50 mA	

8.5.3 Input Voltage Limit Register (Address = 02h) [reset = 85h]

REG02 is shown in Figure 39 and described in Table 11.

Return to Summary Table.

Figure 39. REG02 Register

Bit	7	6	5	4	3	2	1	0
Reset	1	0	0	0	0	1	0	1
Field	EN_VINDPM_RST	EN_BAT_DISCHG	PFM_OOA_DIS	VINDPM[4:0]				

Table 11. REG02 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	EN_VINDPM_RST	R/W	Yes	Yes	Enable VINDPM automatic reset upon adapter plugin: 0 – Disable VINDPM reset when adapter is plugged in 1 – Enable VINDPM reset when adapter is plugged in (VINDPM resets to default value after Input Source Type Detection)	
6	EN_BAT_DISCHG	R/W	Yes	Yes	Enable BAT pin discharge load (I_{BAT_DISCHG}): 0 – Disable load (Default) 1 – Enable BAT discharge load	
5	PFM_OOA_DIS	R/W	Yes	No	PFM Out-of-Audio (OOA) Mode Disable: 0 – Out-of-audio mode enabled while in converter is in PFM (Default) 1 – Out-of-audio mode disabled while in converter is in PFM	
4	VINDPM[4]	R/W	Yes	No	1600 mV	Absolute Input Voltage Limit: Offset: 3.9 V Range: 3.9 V – 5.5 V Default: 4.4 V Note: VINDPM > 5.5 V (10h) clamped to 5.5 V. VINDPM register is reset upon adapter plug-in if EN_VINDPM_RST = 1.
3	VINDPM[3]	R/W	Yes	No	800 mV	
2	VINDPM[2]	R/W	Yes	No	400 mV	
1	VINDPM[1]	R/W	Yes	No	200 mV	
0	VINDPM[0]	R/W	Yes	No	100 mV	

8.5.4 Input Current Limit Register (Address = 03h) [reset = 39h]

REG03 is shown in Figure 40 and described in Table 12.

Return to Summary Table.

Figure 40. REG03 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	1	1	1	0	0	1
Field	FORCE_ICO	FORCE_INDET	EN_ICO	IINDPM[4:0]				

Table 12. REG03 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	FORCE_ICO	R/W	Yes	Yes	Force Start Input Current Optimizer (ICO): 0 – Do not force ICO (default) 1 – Force ICO start Note: This bit can only be set and always returns 0 after ICO starts. This bit only valid when EN_ICO = 1	
6	FORCE_INDET	R/W	Yes	Yes	Force D+/D– Detection: 0 – Not in D+/D– detection (default) 1 – Force D+/D– detection	
5	EN_ICO	R/W	Yes	No	Input Current Optimization (ICO) Algorithm Control: 0 – Disable ICO 1 – Enable ICO (default)	
4	IINDPM[4]	R/W	Yes	No	1600 mA	Input Current Limit: Offset: 500 mA Range: 500 mA – 3300 mA Default: 3000 mA Note: IINDPM > 3300 mA (1Ch) clamped to 3300 mA. Actual input current limit is lower of I^2C or ILIM pin.
3	IINDPM[3]	R/W	Yes	No	800 mA	
2	IINDPM[2]	R/W	Yes	No	400 mA	
1	IINDPM[1]	R/W	Yes	No	200 mA	
0	IINDPM[0]	R/W	Yes	No	100 mA	

8.5.5 Precharge and Termination Current Limit Register (Address = 04h) [reset = 22h]

REG04 is shown in Figure 41 and described in Table 13.

Return to Summary Table.

Figure 41. REG04 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	1	0	0	0	1	0
Field	IPRECHG[3:0]				ITERM[3:0]			

Table 13. REG04 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	IPRECHG[3]	R/W	Yes	Yes	400 mA	Precharge Current Limit: Offset: 50 mA Range: 50 mA – 800 mA Default: 150 mA
6	IPRECHG[2]	R/W	Yes	Yes	200 mA	
5	IPRECHG[1]	R/W	Yes	Yes	100 mA	
4	IPRECHG[0]	R/W	Yes	Yes	50 mA	
3	ITERM[3]	R/W	Yes	Yes	400 mA	Termination Current Limit: Offset: 50 mA Range: 50 mA – 800 mA Default: 150 mA
2	ITERM[2]	R/W	Yes	Yes	200 mA	
1	ITERM[1]	R/W	Yes	Yes	100 mA	
0	ITERM[0]	R/W	Yes	Yes	50 mA	

8.5.6 Charger Control 1 Register (Address = 05h) [reset = 9Dh]

REG05 is shown in Figure 42 and described in Table 14.

Return to Summary Table.

Figure 42. REG05 Register

Bit	7	6	5	4	3	2	1	0
Reset	1	0	0	1	1	1	0	1
Field	EN_TERM	RESERVED	WATCHDOG[1:0]		EN_TIMER	CHG_TIMER[1:0]		TMR2X_EN

Table 14. REG05 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	EN_TERM	R/W	Yes	Yes	Termination Control: 0 – Disable termination 1 – Enable termination (default)
6	RESERVED	R/W	Yes	Yes	Reserved bit always reads 0
5	WATCHDOG[1]	R/W	Yes	Yes	I2C Watchdog Timer Settings: 00 – Disable WD Timer 01 – 40s (default) 10 – 80s 11 – 160s
4	WATCHDOG[0]	R/W	Yes	Yes	
3	EN_TIMER	R/W	Yes	Yes	Charging Safety Timer Enable 0 – Disable 1 – Enable (Default)
2	CHG_TIMER[1]	R/W	Yes	Yes	Fast Charge Timer Setting 00 – 5 hrs 01 – 8 hrs 10 – 12 hrs (Default) 11 – 20 hrs
1	CHG_TIMER[0]	R/W	Yes	Yes	
0	TMR2X_EN	R/W	Yes	Yes	Safety Timer during DPM or TREG 0 – Safety timer always count normally 1 – Safety timer slowed by 2X during input DPM or TREG (Default)

System Note: When the WATCHDOG bits are changed (writing the same value does not change WATCHDOG bit), the internal counter is reset. The same applies for the CHG_TIMER bits. Only changing the value in the register will reset the CHG_TIMER

8.5.7 Charger Control 2 Register (Address = 06h) [reset = 7Dh]

REG06 is shown in Figure 43 and described in Table 15.

Return to Summary Table.

Figure 43. REG06 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	1	1	1	1	1	0	1
Field	EN_OTG	AUTO_INDET_EN	TREG[1:0]		EN_CHG	BATLOWV	VRECHG[1:0]	

Table 15. REG06 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	EN_OTG	R/W	Yes	Yes	Buck (OTG) Mode control: 0 – Disable OTG (default) 1 – Enable OTG Note: If EN_OTG and EN_CHG are set simultaneously, EN_CHG takes priority
6	AUTO_INDET_EN	R/W	Yes	Yes	Automatic Input Source Detection Enable: 0 – Disable D+/D– detection when VBUS plugs in 1 – Enable D+/D– detection when VBUS plugs in (default)
5	TREG[1]	R/W	Yes	Yes	Thermal Regulation Threshold 00 – 60°C 01 – 80°C 10 – 100°C 11 – 120°C (Default)
4	TREG[0]	R/W	Yes	Yes	
3	EN_CHG	R/W	Yes	Yes	Charger Enable Configuration 0 – Charge Disable 1 – Charge Enable (default) Note: If EN_OTG and EN_CHG are set simultaneously, EN_CHG takes priority
2	BATLOWV	R/W	Yes	Yes	Battery precharge to fast-charge threshold: 0 – 5.6 V 1 – 6.0 V (default)
1	VRECHG[1]	R/W	Yes	No	200 mV Battery Recharge Threshold Offset (below VREG): Offset: 100 mV Range: 100 mV – 400 mV Default: 200 mV
0	VRECHG[0]	R/W	Yes	No	100 mV

8.5.8 Charger Control 3 Register (Address = 07h) [reset = 0Ah]

REG07 is shown in Figure 44 and described in Table 16.

Return to Summary Table.

Figure 44. REG07 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	1	0	1	0
Field	PFM_DIS	WD_RST	TOPOFF_TIMER[1:0]		SYS_MIN[3:0]			

Table 16. REG07 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	PFM_DIS	R/W	Yes	No	PFM Mode Disable control: 0 – Enable PFM operation (default) 1 – Disable PFM operation	
6	WD_RST	R/W	Yes	Yes	I2C Watchdog Timer Reset: 0 – Normal 1 – Reset (Bit goes back to 0 after timer reset)	
5	TOPOFF_TIMER[1]	R/W	Yes	Yes	Top-off Timer Control : 00 – Disabled (default) 01 – 15 mins 10 – 30 mins 11 – 45 mins	
4	TOPOFF_TIMER[0]	R/W	Yes	Yes		
3	SYS_MIN[3]	R/W	Yes	No	800 mV	Minimum System Voltage Limit Offset: 6.0 V Range: 6.0 V – 7.5 V Default: 7.0 V
2	SYS_MIN[2]	R/W	Yes	No	400 mV	
1	SYS_MIN[2]	R/W	Yes	No	200 mV	
0	SYS_MIN[0]	R/W	Yes	No	100 mV	

8.5.9 Charger Control 4 Register (Address = 08h) [reset = 0Dh]

REG08 is shown in Figure 45 and described in Table 17.

Return to Summary Table.

Figure 45. REG08 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	1	1	0	1
Field	BHOT[1:0]		BCOLD	JEITA_VSET[1:0]		JEITA_ISETH	JEITA_ISETC[1:0]	

Table 17. REG08 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	BHOT[1]	R/W	Yes	Yes	OTG Mode TS HOT Temperature Threshold: 00 – VBHOT1 threshold (34.75%) (default) 01 – VBHOT0 threshold (37.75%) 10 – VBHOT2 threshold (31.25%) 11 – Disable OTG mode thermal protection
6	BHOT[0]	R/W	Yes	Yes	
5	BCOLD	R/W	Yes	Yes	OTG Mode TS COLD Temperature Threshold: 0 – VBCOLD0 threshold (77%) (default) 1 – VBCOLD1 threshold (80%)
4	JEITA_VSET[1]	R/W	Yes	Yes	JEITA High Temp. (45°C – 60°C) Voltage Setting: 00 – Charge Suspend 01 – Set VREG to 8.0 V (default) 10 – Set VREG to 8.3 V 11 – VREG unchanged
3	JEITA_VSET[0]	R/W	Yes	Yes	
2	JEITA_ISETH	R/W	Yes	Yes	JEITA High Temp. (45°C – 60°C) Current Setting (percentage with respect to ICHG REG01[5:0]): 0 – 40% of ICHG 1 – 100% of ICHG (default)
1	JEITA_ISETC[1]	R/W	Yes	Yes	JEITA Low Temp. (0°C – 10°C) Current Setting (percentage with respect to ICHG REG01[5:0]): 00 – Charge Suspend 01 – 20% of ICHG (default) 10 – 40% of ICHG 11 – 100% of ICHG
0	JEITA_ISETC[0]	R/W	Yes	Yes	

8.5.10 OTG Control Register (Address = 09h) [reset = F6h]

REG09 is shown in Figure 46 and described in Table 18.

Return to Summary Table.

Figure 46. REG09 Register

Bit	7	6	5	4	3	2	1	0
Reset	1	1	1	1	0	1	1	0
Field	OTG_ILIM[3:0]				OTG_VLIM[3:0]			

Table 18. REG09 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	OTG_ILIM[3]	R/W	Yes	Yes	800 mA	Buck (OTG) Mode Current Limit: Offset: 0.5 A Range: 0.5 A – 2.0 A Default: 2 A
6	OTG_ILIM[2]	R/W	Yes	Yes	400 mA	
5	OTG_ILIM[1]	R/W	Yes	Yes	200 mA	
4	OTG_ILIM[0]	R/W	Yes	Yes	100 mA	
3	OTG_VLIM[3]	R/W	Yes	Yes	800 mV	Buck (OTG) Mode Regulation Voltage: Offset: 4.5 V Range: 4.5 V – 5.5 V Default: 5.1 V Note: Values above 5.5 V (Ah) will be clamped to 5.5 V
2	OTG_VLIM[2]	R/W	Yes	Yes	400 mV	
1	OTG_VLIM[1]	R/W	Yes	Yes	200 mV	
0	OTG_VLIM[0]	R/W	Yes	Yes	100 mV	

8.5.11 ICO Current Limit Register (Address = 0Ah) [reset = XXh]

REG0A is shown in Figure 47 and described in Table 19.

Return to Summary Table.

Figure 47. REG0A Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	X	X	X	X	X
Field	RESERVED	RESERVED	RESERVED	ICO_ILIM[4:0]				

Table 19. REG0A Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	RESERVED	R	No	No	Reserved bit always reads 0	
6	RESERVED	R	No	No	Reserved bit always reads 0	
5	RESERVED	R	No	No	Reserved bit always reads 0	
4	ICO_ILIM[4]	R	No	No	1600 mA	Input Current Limit when ICO is enabled: Offset: 500 mA Range: 500 mA – 3300 mA
3	ICO_ILIM[3]	R	No	No	800 mA	
2	ICO_ILIM[2]	R	No	No	400 mA	
1	ICO_ILIM[1]	R	No	No	200 mA	
0	ICO_ILIM[0]	R	No	No	100 mA	

8.5.12 Charger Status 1 Register (Address = 0Bh) [reset = XXh]

REG0B is shown in Figure 48 and described in Table 20.

Return to Summary Table.

Figure 48. REG0B Register

Bit	7	6	5	4	3	2	1	0
Reset	X	X	X	X	X	X	X	X
Field	ADC_DONE_STAT	IINDPM_STAT	VINDPM_STAT	TREG_STAT	WD_STAT	CHRG_STAT[2:0]		

Table 20. REG0B Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	ADC_DONE_STAT	R	No	No	ADC Conversion Status (in one-shot mode only): 0 – Conversion not complete 1 – Conversion complete Note: Always reads 0 in continuous mode
6	IINDPM_STAT	R	No	No	IINDPM Status: 0 – Normal 1 – In IINDPM Regulation (ILIM pin or IINDPM register)
5	VINDPM_STAT	R	No	No	VINDPM Status: 0 – Normal 1 – In VINDPM Regulation
4	TREG_STAT	R	No	No	IC Thermal regulation Status: 0 – Normal 1 – In Thermal Regulation
3	WD_STAT	R	No	No	I2C Watchdog Timer Status bit: 0 – Normal 1 – WD Timer expired
2	CHRG_STAT[2]	R	No	No	Charge Status bits: 000 – Not Charging 001 – Trickle Charge (VBAT < VBAT_SHORT) 010 – Pre-charge (VBAT_SHORT < VBAT < VBAT_LOWV) 011 – Fast-charge (CC mode) 100 – Taper Charge (CV mode) 101 – Top-off Timer Charging 110 – Charge Termination Done 111 – Reserved
1	CHRG_STAT[1]	R	No	No	
0	CHRG_STAT[0]	R	No	No	

8.5.13 Charger Status 2 Register (Address = 0Ch) [reset = XXh]

REG0C is shown in Figure 49 and described in Table 21.

Return to Summary Table.

Figure 49. REG0C Register

Bit	7	6	5	4	3	2	1	0
Reset	X	X	X	X	0	X	X	X
Field	PG_STAT	VBUS_STAT[2:0]			RESERVED	ICO_STAT[1]	ICO_STAT[0]	VSYS_STAT

Table 21. REG0C Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	PG_STAT	R	No	No	Power Good Status: 0 – Not Power Good 1 – Power Good
6	VBUS_STAT[2]	R	No	No	VBUS Detection Status 000 – No Input 001 – USB Host SDP 010 – USB CDP (1.5 A) 011 – USB DCP (3.0 A) 100 – POORSRC detected 7 consecutive times 101 – Unknown Adapter (500 mA) 110 – Non-standard Adapter (1 A/2 A/2.1 A/2.4 A) 111 – OTG
5	VBUS_STAT[1]	R	No	No	
4	VBUS_STAT[0]	R	No	No	
3	RESERVED	R	No	No	Reserved bit always reads 0
2	ICO_STAT[1]	R	No	No	Input Current Optimizer (ICO) Status: 00 – ICO Disabled 01 – ICO Optimization is in progress 10 – Maximum input current detected 11 – Reserved
1	ICO_STAT[0]	R	No	No	
0	VSYS_STAT	R	No	No	VSYS Regulation Status: 0 – Not in SYS_MIN regulation (BAT > VSYS_MIN) 1 – In SYS_MIN regulation (BAT < VSYS_MIN)

8.5.14 NTC Status Register (Address = 0Dh) [reset = 0Xh]

REG0D is shown in Figure 50 and described in Table 22.

Return to Summary Table.

Figure 50. REG0D Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	X	X	X
Field	RESERVED	RESERVED	RESERVED	RESERVED	RESERVED	TS_STAT[2:0]		

Table 22. REG0D Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	RESERVED	R	No	No	Reserved bit always reads 0
6	RESERVED	R	No	No	Reserved bit always reads 0
5	RESERVED	R	No	No	Reserved bit always reads 0
4	RESERVED	R	No	No	Reserved bit always reads 0
3	RESERVED	R	No	No	Reserved bit always reads 0
2	TS_STAT[2]	R	No	No	NTC (TS) Status: 000 – Normal 010 – TS Warm 011 – TS Cool 101 – TS Cold 110 – TS Hot
1	TS_STAT[1]	R	No	No	
0	TS_STAT[0]	R	No	No	

8.5.15 FAULT Status Register (Address = 0Eh) [reset = XXh]

REG0E is shown in Figure 51 and described in Table 23.

Return to Summary Table.

Figure 51. REG0E Register

Bit	7	6	5	4	3	2	1	0
Reset	X	X	X	X	0	0	0	0
Field	VBUS_OVP_STAT	TSHUT_STAT	BATOVP_STAT	TMR_STAT	RESERVED	RESERVED	RESERVED	RESERVED

Table 23. REG0E Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	VBUS_OVP_STAT	R	No	No	Input overvoltage Status: 0 – Normal 1 – Device in overvoltage protection
6	TSHUT_STAT	R	No	No	IC Temperature shutdown Status: 0 – Normal 1 – Device in thermal shutdown protection
5	BATOVP_STAT	R	No	No	Battery over-voltage Status: 0 – Normal 1 – BATOVP (VBAT > VBATOVP)
4	TMR_STAT	R	No	No	Charge Safety timer Status: 0 – Normal 1 – Charge Safety timer expired
3	RESERVED	R	No	No	Reserved bit always reads 0
2	RESERVED	R	No	No	Reserved bit always reads 0
1	RESERVED	R	No	No	Reserved bit always reads 0
0	RESERVED	R	No	No	Reserved bit always reads 0

8.5.16 Charger Flag 1 Register (Address = 0Fh) [reset = 00h]

REG0F is shown in Figure 52 and described in Table 24.

Return to Summary Table.

Figure 52. REG0F Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	ADC_DONE_FLAG	IINDPM_FLAG	VINDPM_FLAG	TREG_FLAG	WD_FLAG	RESERVED	RESERVED	CHRG_FLAG

Table 24. REG0F Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	ADC_DONE_FLAG	R	Yes	No	ADC Conversion Flag (only 1-shot mode): 0 – Conversion not complete 1 – Conversion complete **Note:** Always reads 0 in continuos mode
6	IINDPM_FLAG	R	Yes	No	IINDPM Regulation INT Flag: 0 – Normal 1 – IINDPM signal rising edge detected
5	VINDPM_FLAG	R	Yes	No	VINDPM regulation INT Flag: 0 – Normal 1 – VINDPM signal rising edge detected
4	TREG_FLAG	R	Yes	No	IC Temperature Regulation INT Flag: 0 – Normal 1 – TREG signal rising edge detected
3	WD_FLAG	R	Yes	No	I2C Watchdog INT Flag: 0 – Normal 1 – WD_STAT signal rising edge detected
2	RESERVED	R	Yes	No	Reserved bit always reads 0
1	RESERVED	R	Yes	No	Reserved bit always reads 0
0	CHRG_FLAG	R	Yes	No	Charge Status INT Flag: 0 – Normal 1 – CHRG_STAT[2:0] bits changed (transition to any state)

8.5.17 Charger Flag 2 Register (Address = 10h) [reset = 00h]

REG10 is shown in Figure 53 and described in Table 25.

Return to Summary Table.

Figure 53. REG10 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	PG_FLAG	RESERVED	RESERVED	VBUS_FLAG	RESERVED	TS_FLAG	ICO_FLAG	VSYS_FLAG

Table 25. REG10 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	PG_FLAG	R	Yes	No	Power Good INT Flag: 0 – Normal 1 – PG signal toggle detected
6	RESERVED	R	Yes	No	Reserved bit always reads 0
5	RESERVED	R	Yes	No	Reserved bit always reads 0
4	VBUS_FLAG	R	Yes	No	VBUS Status INT Flag: 0 – Normal 1 – VBUS_STAT[2:0] bits changed (transition to any state)
3	RESERVED	R	Yes	No	Reserved bit always reads 0
2	TS_FLAG	R	Yes	No	TS Status INT Flag: 0 – Normal 1 – TS_STAT[2:0] bits changed (transition to any state)
1	ICO_FLAG	R	Yes	No	Input Current Optimizer (ICO) INT Flag: 0 – Normal 1 – ICO_STAT[1:0] changed (transition to any state)
0	VSYS_FLAG	R	Yes	No	VSYS Regulation INT Flag: 0 – Normal 1 – Entered or exited SYS_MIN regulation

8.5.18 FAULT Flag Register (Address = 11h) [reset = 00h]

REG11 is shown in Figure 54 and described in Table 26.

Return to Summary Table.

Figure 54. REG11 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	VBUS_OVP_FLAG	TSHUT_FLAG	BATOVP_FLAG	TMR_FLAG	SYS_SHORT_FLAG	RESERVED	RESERVED	OTG_FLAG

Table 26. REG11 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	VBUS_OVP_FLAG	R	Yes	No	Input over-voltage INT Flag: 0 – Normal 1 – Entered VBUS_OVP Fault
6	TSHUT_FLAG	R	Yes	No	IC Temperature shutdown INT Flag: 0 – Normal 1 – Entered TSHUT Fault
5	BATOVP_FLAG	R	Yes	No	Battery over-voltage INT Flag: 0 – Normal 1 – Entered BATOVP Fault
4	TMR_FLAG	R	Yes	No	Charge Safety timer Fault INT Flag: 0 – Normal 1 – Charge Safety timer expired rising edge detected
3	SYS_SHORT_FLAG	R	Yes	No	System Short INT Flag: 0 – Normal 1 – Stopped switching due to boost converter overload
2	RESERVED	R	Yes	No	Reserved bit always reads 0
1	RESERVED	R	Yes	No	Reserved bit always reads 0
0	OTG_FLAG	R	Yes	No	OTG Buck Mode Fault INT Flag: 0 – Normal 1 – VBUS overloaded in OTG, or VBUS OVP, or battery below VOTG_BAT

8.5.19 Charger Mask 1 Register (Address = 12h) [reset = 00h]

REG12 is shown in Figure 55 and described in Table 27.

Return to Summary Table.

Figure 55. REG12 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	ADC_DONE_MASK	IINDPM_MASK	VINDPM_MASK	TREG_MASK	WD_MASK	RESERVED	RESERVED	CHRG_MASK

Table 27. REG12 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	ADC_DONE_MASK	R/W	Yes	No	ADC Conversion INT Mask Flag (only 1-shot mode) 0 – ADC_DONE does produce INT pulse 1 – ADC_DONE does produce not INT pulse
6	IINDPM_MASK	R/W	Yes	No	IINDPM Regulation INT Mask 0 – IINDPM entry produces INT pulse 1 – IINDPM entry does not produce INT pulse
5	VINDPM_MASK	R/W	Yes	No	VINDPM Regulation INT Mask 0 – VINDPM entry produces INT pulse 1 – VINDPM entry not produce INT pulse
4	TREG_MASK	R/W	Yes	No	IC Temperature Regulation INT Mask 0 – TREG entry produces INT pulse 1 – TREG entry produce INT pulse
3	WD_MASK	R/W	Yes	No	I2C Watchdog Timer INT Mask 0 – WD_STAT rising edge produces INT pulse 1 – WD_STAT rising edge does not produce INT
2	RESERVED	R/W	Yes	No	Reserved bit always reads 0
1	RESERVED	R/W	Yes	No	Reserved bit always reads 0
0	CHRG_MASK	R/W	Yes	No	Charge Status INT Mask 0 – CHRG_STAT[2:0] bit change produces INT 1 – CHRG_STAT[2:0] bit change does not produce INT pulse

8.5.20 Charger Mask 2 Register (Address = 13h) [reset = 00h]

REG13 is shown in Figure 56 and described in Table 28.

Return to Summary Table.

Figure 56. REG13 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	PG_MASK	RESERVED	RESERVED	VBUS_MASK	RESERVED	TS_MASK	ICO_MASK	VSYS_MASK

Table 28. REG13 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	PG_MASK	R/W	Yes	No	Power Good INT Mask: 0 – PG toggle produces INT pulse 1 – PG toggle does not produce INT pulse
6	RESERVED	R/W	Yes	No	Reserved bit always reads 0
5	RESERVED	R/W	Yes	No	Reserved bit always reads 0
4	VBUS_MASK	R/W	Yes	No	VBUS Status INT Mask: 0 – VBUS_STAT[2:0] bit change produces INT 1 – VBUS_STAT[2:0] bit change does not produces INT
3	RESERVED	R/W	Yes	No	Reserved bit always reads 0
2	TS_MASK	R/W	Yes	No	TS Status INT Mask: 0 – TS_STAT[2:0] bit change produces INT 1 – TS_STAT[2:0] bit change does not produces INT pulse
1	ICO_MASK	R/W	Yes	No	Input Current Optimizer (ICO) INT Mask: 0 – ICO_STAT rising edge produces INT 1 – ICO_STAT rising edge does not produce INT
0	VSYS_MASK	R/W	Yes	No	VSYS Regulation INT Mask: 0 – Entering or exiting SYS_MIN produces INT 1 – Entering or exiting SYS_MIN does not produce INT

8.5.21 FAULT Mask Register (Address = 14h) [reset = 00h]

REG14 is shown in Figure 57 and described in Table 29.

Return to Summary Table.

Figure 57. REG14 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	VBUS_OVP_MASK	TSHUT_MASK	BATOVP_MASK	TMR_MASK	SYS_SHORT_MASK	RESERVED	RESERVED	OTG_MASK

Table 29. REG14 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	VBUS_OVP_MASK	R/W	Yes	No	Input over-voltage INT Mask: 0 – VBUS_OVP rising edge produces INT pulse 1 – VBUS_OVP rising edge does not produce INT pulse
6	TSHUT_MASK	R/W	Yes	No	Thermal Shutdown INT Mask: 0 – TSHUT rising edge produces INT pulse 1 – TSHUT rising edge does not produce INT pulse
5	BATOVP_MASK	R/W	Yes	No	Battery overvoltage INT Mask: 0 – BATOVP rising edge produces INT pulse 1 – BATOVP rising edge does not produce INT pulse
4	TMR_MASK	R/W	Yes	No	Charge Safety Timer Fault INT Mask: 0 – Timer expired rising edge produces INT pulse 1 – Timer expired rising edge does not produce INT pulse
3	SYS_SHORT_MASK	R/W	Yes	No	System Short Fault INT Mask: 0 – System short rising edge produces INT pulse 1 – System short rising edge does not produce INT pulse
2	RESERVED	R/W	Yes	No	Reserved bit always reads 0
1	RESERVED	R/W	Yes	No	Reserved bit always reads 0
0	OTG_MASK	R/W	Yes	No	OTG Buck Mode Fault INT Mask: 0 – OTG_STAT event produces INT 1 – OTG_STAT event does not produce INT

8.5.22 ADC Control Register (Address = 15h) [reset = 30h]

REG15 is shown in Figure 58 and described in Table 30.

Return to Summary Table.

Figure 58. REG15 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	1	1	0	0	0	0
Field	ADC_EN	ADC_RATE	ADC_SAMPLE[1:0]		RESERVED	RESERVED	RESERVED	RESERVED

Table 30. REG15 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	ADC_EN	R/W	Yes	Yes	ADC Control: 0 – Disable ADC (default) 1 – Enable ADC
6	ADC_RATE	R/W	Yes	No	0 – Continuous conversion (default) 1 – One-shot conversion
5	ADC_SAMPLE[1]	R/W	Yes	No	Sample Speed of ADC: 00 – 15-bit effective resolution 01 – 14-bit effective resolution 10 – 13-bit effective resolution 11 – 12-bit effective resolution (default)
4	ADC_SAMPLE[0]	R/W	Yes	No	
3	RESERVED	R/W	Yes	No	Reserved bit always reads 0
2	RESERVED	R/W	Yes	No	Reserved bit always reads 0
1	RESERVED	R/W	Yes	No	Reserved bit always reads 0
0	RESERVED	R/W	Yes	No	Reserved bit always reads 0

8.5.23 ADC Function Disable Register (Address = 16h) [reset = 00h]

REG16 is shown in Figure 59 and described in Table 31.

Return to Summary Table.

Figure 59. REG16 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	IBUS_ADC_DIS	ICHG_ADC_DIS	VBUS_ADC_DIS	VBAT_ADC_DIS	VSYS_ADC_DIS	TS_ADC_DIS	RESERVED	TDIE_ADC_DIS

Table 31. REG16 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	IBUS_ADC_DIS	R/W	Yes	No	0 – Enable conversion 1 – Disable conversion
6	ICHG_ADC_DIS	R/W	Yes	No	0 – Enable conversion 1 – Disable conversion
5	VBUS_ADC_DIS	R/W	Yes	No	0 – Enable conversion 1 – Disable conversion
4	VBAT_ADC_DIS	R/W	Yes	No	0 – Enable conversion 1 – Disable conversion
3	VSYS_ADC_DIS	R/W	Yes	No	0 – Enable conversion 1 – Disable conversion
2	TS_ADC_DIS	R/W	Yes	No	0 – Enable conversion 1 – Disable conversion
1	RESERVED	R/W	Yes	No	Reserved bit always reads 0
0	TDIE_ADC_DIS	R/W	Yes	No	0 – Enable conversion 1 – Disable conversion

8.5.24 IBUS ADC 1 Register (Address = 17h) [reset = 00h]

REG17 is shown in Figure 60 and described in Table 32.

Return to Summary Table.

Figure 60. REG17 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	IBUS_ADC[15:8]							

Table 32. REG17 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	IBUS_ADC[15]	R	Yes	No	Sign bit: overall results reported in two's complement.	
6	IBUS_ADC[14]	R	Yes	No	16384 mA	
5	IBUS_ADC[13]	R	Yes	No	8192 mA	
4	IBUS_ADC[12]	R	Yes	No	4096 mA	
3	IBUS_ADC[11]	R	Yes	No	2048 mA	VBUS Current Reading (positive current flows into VBUS pin, negative current flows out ot VBUS pin): Range: 0 A – 4 A
2	IBUS_ADC[10]	R	Yes	No	1024 mA	
1	IBUS_ADC[9]	R	Yes	No	512 mA	
0	IBUS_ADC[8]	R	Yes	No	256 mA	

8.5.25 IBUS ADC 0 Register (Address = 18h) [reset = 00h]

REG18 is shown in Figure 61 and described in Table 33.

Return to Summary Table.

Figure 61. REG18 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	IBUS_ADC[7:0]							

Table 33. REG18 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	IBUS_ADC[7]	R	Yes	No	128 mA	VBUS Current Reading (positive current flows into VBUS pin, negative current flows out ot VBUS pin): Range: 0 A – 4 A
6	IBUS_ADC[6]	R	Yes	No	64 mA	
5	IBUS_ADC[5]	R	Yes	No	32 mA	
4	IBUS_ADC[4]	R	Yes	No	16 mA	
3	IBUS_ADC[3]	R	Yes	No	8 mA	
2	IBUS_ADC[2]	R	Yes	No	4 mA	
1	IBUS_ADC[1]	R	Yes	No	2 mA	
0	IBUS_ADC[0]	R	Yes	No	1 mA	

8.5.26 ICHG ADC 1 Register (Address = 19h) [reset = 00h]

REG19 is shown in Figure 62 and described in Table 34.

Return to Summary Table.

Figure 62. REG19 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	RESERVED	ICHG_ADC[14:8]						

Table 34. REG19 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	RESERVED	R	Yes	No	Reserved register always reads 0.	
6	ICHG_ADC[14]	R	Yes	No	16384 mA	
5	ICHG_ADC[13]	R	Yes	No	8192 mA	
4	ICHG_ADC[12]	R	Yes	No	4096 mA	
3	ICHG_ADC[11]	R	Yes	No	2048 mA	Charge Current Reading: Range: 0 A – 4 A
2	ICHG_ADC[10]	R	Yes	No	1024 mA	
1	ICHG_ADC[9]	R	Yes	No	512 mA	
0	ICHG_ADC[8]	R	Yes	No	256 mA	

8.5.27 ICHG ADC 0 Register (Address = 1Ah) [reset = 00h]

REG1A is shown in Figure 63 and described in Table 35.

Return to Summary Table.

Figure 63. REG1A Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	ICHG_ADC[7:0]							

Table 35. REG1A Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	ICHG_ADC[7]	R	Yes	No	128 mA	Charge Current Reading: Range: 0 A – 4 A
6	ICHG_ADC[6]	R	Yes	No	64 mA	
5	ICHG_ADC[5]	R	Yes	No	32 mA	
4	ICHG_ADC[4]	R	Yes	No	16 mA	
3	ICHG_ADC[3]	R	Yes	No	8 mA	
2	ICHG_ADC[2]	R	Yes	No	4 mA	
1	ICHG_ADC[1]	R	Yes	No	2 mA	
0	ICHG_ADC[0]	R	Yes	No	1 mA	

8.5.28 VBUS ADC 1 Register (Address = 1Bh) [reset = 00h]

REG1B is shown in Figure 64 and described in Table 36.

Return to Summary Table.

Figure 64. REG1B Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	VBUS_ADC[15:8]							

Table 36. REG1B Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	VBUS_ADC[15]	R	Yes	No	Sign bit: overall results reported in two's complement.	
6	VBUS_ADC[14]	R	Yes	No	16384 mV	
5	VBUS_ADC[13]	R	Yes	No	8192 mV	VBUS Voltage reading Range: 0 V – 10 V
4	VBUS_ADC[12]	R	Yes	No	4096 mV	
3	VBUS_ADC[11]	R	Yes	No	2048 mV	
2	VBUS_ADC[10]	R	Yes	No	1024 mV	
1	VBUS_ADC[9]	R	Yes	No	512 mV	
0	VBUS_ADC[8]	R	Yes	No	256 mV	

8.5.29 VBUS ADC 0 Register (Address = 1Ch) [reset = 00h]

REG1C is shown in Figure 65 and described in Table 37.

Return to Summary Table.

Figure 65. REG1C Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	VBUS_ADC[7:0]							

Table 37. REG1C Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	VBUS_ADC[7]	R	Yes	No	128 mV	VBUS Voltage Reading: Range: 0 V – 10 V
6	VBUS_ADC[6]	R	Yes	No	64 mV	
5	VBUS_ADC[5]	R	Yes	No	32 mV	
4	VBUS_ADC[4]	R	Yes	No	16 mV	
3	VBUS_ADC[3]	R	Yes	No	8 mV	
2	VBUS_ADC[2]	R	Yes	No	4 mV	
1	VBUS_ADC[1]	R	Yes	No	2 mV	
0	VBUS_ADC[0]	R	Yes	No	1 mV	

8.5.30 VBAT ADC 1 Register (Address = 1Dh) [reset = 00h]

REG1D is shown in Figure 66 and described in Table 38.

Return to Summary Table.

Figure 66. REG1D Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	VBAT_ADC[15:8]							

Table 38. REG1D Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	VBAT_ADC[15]	R	Yes	No	Sign bit: overall results reported in two's complement.	
6	VBAT_ADC[14]	R	Yes	No	16384 mV	
5	VBAT_ADC[13]	R	Yes	No	8192 mV	VBAT Voltage reading: Range: 0 V – 10 V
4	VBAT_ADC[12]	R	Yes	No	4096 mV	
3	VBAT_ADC[11]	R	Yes	No	2048 mV	
2	VBAT_ADC[10]	R	Yes	No	1024 mV	
1	VBAT_ADC[9]	R	Yes	No	512 mV	
0	VBAT_ADC[8]	R	Yes	No	256 mV	

8.5.31 VBAT ADC 0 Register (Address = 1Eh) [reset = 00h]

REG1E is shown in Figure 67 and described in Table 39.

Return to Summary Table.

Figure 67. REG1E Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	VBAT_ADC[7:0]							

Table 39. REG1E Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	VBAT_ADC[7]	R	Yes	No	128 mV	VBAT Voltage reading: Range: 0 V – 10 V
6	VBAT_ADC[6]	R	Yes	No	64 mV	
5	VBAT_ADC[5]	R	Yes	No	32 mV	
4	VBAT_ADC[4]	R	Yes	No	16 mV	
3	VBAT_ADC[3]	R	Yes	No	8 mV	
2	VBAT_ADC[2]	R	Yes	No	4 mV	
1	VBAT_ADC[1]	R	Yes	No	2 mV	
0	VBAT_ADC[0]	R	Yes	No	1 mV	

8.5.32 VSYS ADC 1 Register (Address = 1Fh) [reset = 00h]

REG1F is shown in Figure 68 and described in Table 40.

Return to Summary Table.

Figure 68. REG1F Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	VSYS_ADC[15:8]							

Table 40. REG1F Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	VSYS_ADC[15]	R	Yes	No	Sign bit: overall results reported in two's complement.	
6	VSYS_ADC[14]	R	Yes	No	16384 mV	
5	VSYS_ADC[13]	R	Yes	No	8192 mV	VSYS Voltage reading: Range: 0 V – 10 V
4	VSYS_ADC[12]	R	Yes	No	4096 mV	
3	VSYS_ADC[11]	R	Yes	No	2048 mV	
2	VSYS_ADC[10]	R	Yes	No	1024 mV	
1	VSYS_ADC[9]	R	Yes	No	512 mV	
0	VSYS_ADC[8]	R	Yes	No	256 mV	

8.5.33 VSYS ADC 0 Register (Address = 20h) [reset = 00h]

REG20 is shown in Figure 69 and described in Table 41.

Return to Summary Table.

Figure 69. REG20 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	VSYS_ADC[7:0]							

Table 41. REG20 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	VSYS_ADC[7]	R	Yes	No	128 mV	VSYS Voltage reading: Range: 0 V – 10 V
6	VSYS_ADC[6]	R	Yes	No	64 mV	
5	VSYS_ADC[5]	R	Yes	No	32 mV	
4	VSYS_ADC[4]	R	Yes	No	16 mV	
3	VSYS_ADC[3]	R	Yes	No	8 mV	
2	VSYS_ADC[2]	R	Yes	No	4 mV	
1	VSYS_ADC[1]	R	Yes	No	2 mV	
0	VSYS_ADC[0]	R	Yes	No	1 mV	

8.5.34 TS ADC 1 Register (Address = 21h) [reset = 00h]

REG21 is shown in Figure 70 and described in Table 42.

Return to Summary Table.

Figure 70. REG21 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	TS_ADC[15:8]							

Table 42. REG21 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	TS_ADC[15]	R	Yes	No	Sign bit: overall results reported in two's complement.	
6	TS_ADC[14]	R	Yes	No		
5	TS_ADC[13]	R	Yes	No		
4	TS_ADC[12]	R	Yes	No		
3	TS_ADC[11]	R	Yes	No		
2	TS_ADC[10]	R	Yes	No		
1	TS_ADC[9]	R	Yes	No	50.0%	TS as percentage of REGN reading: Range: 0% – 94.9%
0	TS_ADC[8]	R	Yes	No	25.0%	

8.5.35 TS ADC 0 Register (Address = 22h) [reset = 00h]

REG22 is shown in Figure 71 and described in Table 43.

Return to Summary Table.

Figure 71. REG22 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	TS_ADC[7:0]							

Table 43. REG22 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	TS_ADC[7]	R	Yes	No	12.50%	TS as percentage of REGN reading: Range: 0% – 94.9%
6	TS_ADC[6]	R	Yes	No	6.25%	
5	TS_ADC[5]	R	Yes	No	3.125%	
4	TS_ADC[4]	R	Yes	No	1.563%	
3	TS_ADC[3]	R	Yes	No	0.781%	
2	TS_ADC[2]	R	Yes	No	0.391%	
1	TS_ADC[1]	R	Yes	No	0.195%	
0	TS_ADC[0]	R	Yes	No	0.098%	

8.5.36 TDIE ADC 1 Register (Address = 23h) [reset = 00h]

REG23 is shown in Figure 72 and described in Table 44.

Return to Summary Table.

Figure 72. REG23 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	RESERVED	TDIE_ADC[14:8]						

Table 44. REG23 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	RESERVED	R	Yes	No	Reserved bit always reads 0	
6	TDIE_ADC[14]	R	Yes	No		
5	TDIE_ADC[13]	R	Yes	No		
4	TDIE_ADC[12]	R	Yes	No		
3	TDIE_ADC[11]	R	Yes	No		
2	TDIE_ADC[10]	R	Yes	No		
1	TDIE_ADC[9]	R	Yes	No		
0	TDIE_ADC[8]	R	Yes	No	128°C	TDIE (IC Temperature) reading: Range: 0°C – 128°C

8.5.37 TDIE ADC 0 Register (Address = 24h) [reset = 00h]

REG24 is shown in Figure 73 and described in Table 45.

Return to Summary Table.

Figure 73. REG24 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	0	0	0	0	0
Field	TDIE_ADC[7:0]							

Table 45. REG24 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description	
7	TDIE_ADC[7]	R	Yes	No	64°C	TDIE (IC Temperature) reading: Range: 0°C – 128°C
6	TDIE_ADC[6]	R	Yes	No	32°C	
5	TDIE_ADC[5]	R	Yes	No	16°C	
4	TDIE_ADC[4]	R	Yes	No	8°C	
3	TDIE_ADC[3]	R	Yes	No	4°C	
2	TDIE_ADC[2]	R	Yes	No	2°C	
1	TDIE_ADC[1]	R	Yes	No	1°C	
0	TDIE_ADC[0]	R	Yes	No	0.5°C	

8.5.38 Part Information Register (Address = 25h) [reset = 11h]

REG25 is shown in Figure 74 and described in Table 46.

Return to Summary Table.

Figure 74. REG25 Register

Bit	7	6	5	4	3	2	1	0
Reset	0	0	0	1	0	0	0	1
Field	REG_RST	PN[3:0]				DEV_REV[2:0]		

Table 46. REG25 Register Field Descriptions

Bit	Field	Type	Reset by REG_RST	Reset by WATCHDOG	Description
7	REG_RST	R	Yes	No	Register Reset: 0 – Keep current register settings 1 – Reset to default register value and reset safety timer (bit resets to 0 after register reset is complete)
6	PN[3]	R	Yes	No	0010: BQ25882
5	PN[2]	R	Yes	No	
4	PN[1]	R	Yes	No	
3	PN[0]	R	Yes	No	
2	DEV_REV[2]	R	Yes	No	Device revision: 001
1	DEV_REV[1]	R	Yes	No	
0	DEV_REV[0]	R	Yes	No	

9 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

A typical application consists of the device configured as an I^2C controlled power path management device and a dual-cell battery charger for Li-Ion and Li-polymer batteries used in a wide range of smartphones and other portable devices. It integrates an input blocking FET (QBLK, Q1), high-side switching FET (QHS, Q2), low-side switching FET (QLS, Q3), and battery FET (QBAT, Q4) between system and battery. The device also integrates a bootstrap diode for the high-side gate drive.

9.2 Typical Application

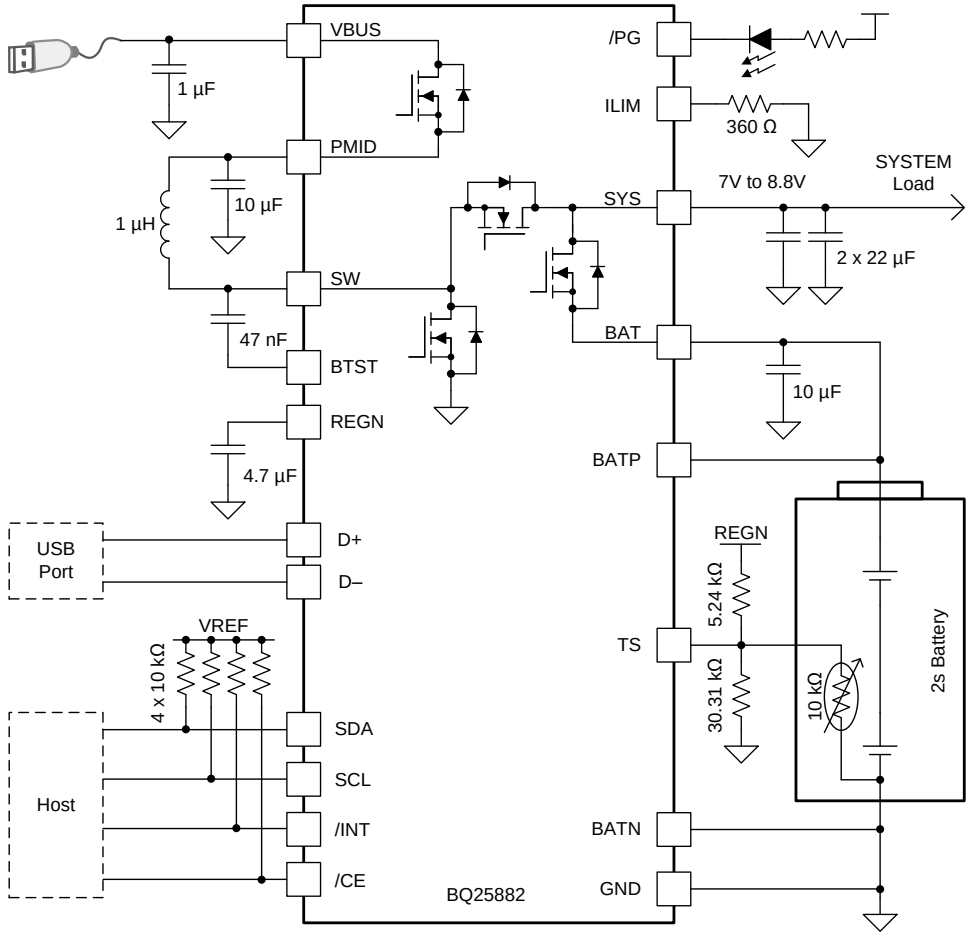


Figure 75. BQ25882 Typical Application Diagram

9.2.1 Design Requirements

For this design example, use the parameters shown in the table below.

Typical Application (continued)

Table 47. Design Parameters

PARAMETER	VALUE
VBUS voltage range	3.9 V to 6.2 V
Input current limit (IINDPM[4:0])	3.0 A
Fast charge current limit (ICHG[5:0])	1.5 A
Minimum system voltage (SYS_MIN[3:0])	7.0 V
Cell regulation voltage (VREG[7:0])	8.7 V

9.2.2 Detailed Design Procedure

9.2.2.1 Inductor Selection

The device has 1.5-MHz switching frequency to allow the use of small inductor and capacitor values. The inductor saturation current should be higher than the input current (I_{IN}) plus half the ripple current (I_{RIPPLE}):

$$I_{SAT} \geq I_{IN} + \frac{I_{RIPPLE}}{2} \tag{5}$$

The inductor ripple current (I_{RIPPLE}) depends on input voltage (V_{VBUS}), duty cycle (D = 1 – V_{BUS}/V_{BAT}), switching frequency (f_{SW}) and inductance (L):

$$I_{RIPPLE} = \frac{V_{BUS} \times (V_{SYS} - V_{BUS})}{V_{SYS} \times f_{SW} \times L} \tag{6}$$

The maximum inductor ripple current happens in the vicinity of D = 0.5. Usually inductor ripple is designed in the range of (20 – 40%) maximum charging current as a trade-off between inductor size and efficiency for a practical design.

9.2.2.2 Input (VBUS / PMID) Capacitor

Input capacitor should have enough ripple current rating to absorb input switching ripple current. The worst case RMS ripple current occurs when duty cycle is 0.5. If the converter does not operate at 50% duty cycle, then the worst case capacitor RMS current I_{PMID} occurs where the duty cycle is closest to 50% and can be estimated by

$$I_{PMID} = \frac{I_{RIPPLE}}{2 \times \sqrt{3}} \approx 0.29 \times I_{RIPPLE} \tag{7}$$

Low ESR ceramic capacitor such as X7R or X5R is preferred for input decoupling capacitor and should be placed close to the PMID and GND pins of the IC. Voltage rating of the capacitor must be higher than normal input voltage level. 25-V rating or higher capacitor is preferred for up to 5-V input voltage. 10-μF capacitor is suggested for up to 3.3-A input current.

9.2.2.3 Output (VSYS) Capacitor

SYS capacitor is the boost converter output capacitor and should also have enough ripple current rating to absorb output switching ripple current. The output capacitor RMS current I_{CSYS} is given:

$$I_{CSYS,\,rms} = I_{OUT} \times \sqrt{\frac{D}{1-D}} \tag{8}$$

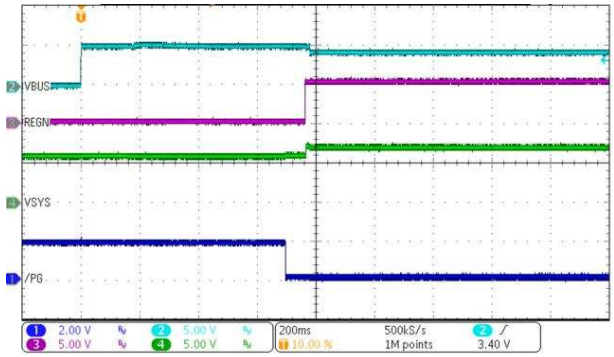
The output capacitor voltage ripple is a function of the boost output current (I_{OUT}), and can be calculated as follows:

$$\Delta V_{SYS} = \frac{I_{OUT} \times D}{f_{SW} \times C_{SYS}} \tag{9}$$

Low ESR ceramic capacitor such as X7R or X5R is preferred for SYS decoupling capacitor and should be placed close to the SYS and GND pins of the IC. Voltage rating of the capacitor must be higher than normal output voltage level. 16-V rating or higher capacitor is preferred. 40-μF capacitor is suggested for up to 2.2-A boost converter output current.

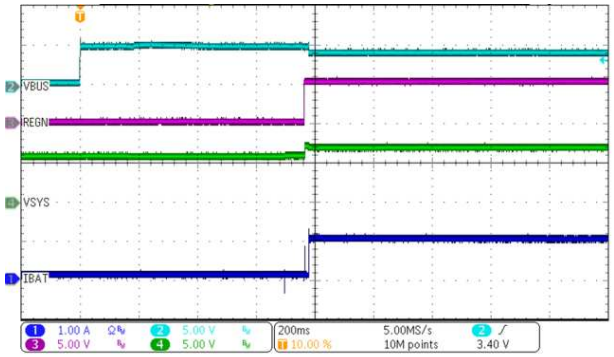
9.2.3 Application Curves

C_{VBUS} = 1µF, C_{PMID}= 10µF, C_{BAT} = 10µF, C_{SYS} = 44µF, L = DFE252012F-1R0 (1µH) (unless otherwise specified)



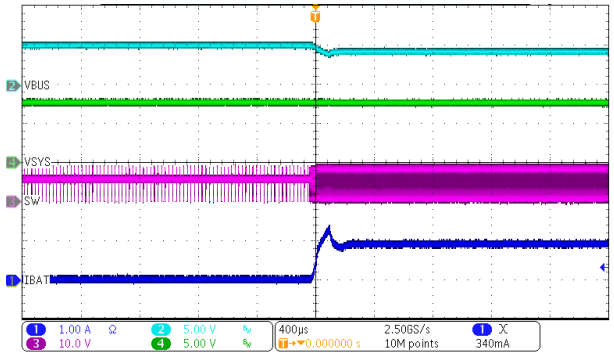
VBUS = 5V VBAT = 6.0V

Figure 76. Adapter Power Up With Charge Enabled



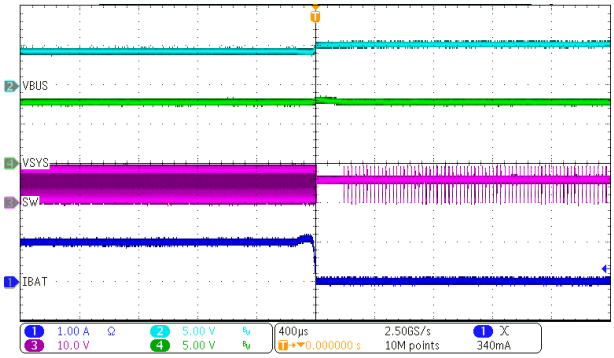
VBUS = 5V VBAT = 6.0V ICHG = 1A

Figure 77. Adapter Power Up With Charge Enabled



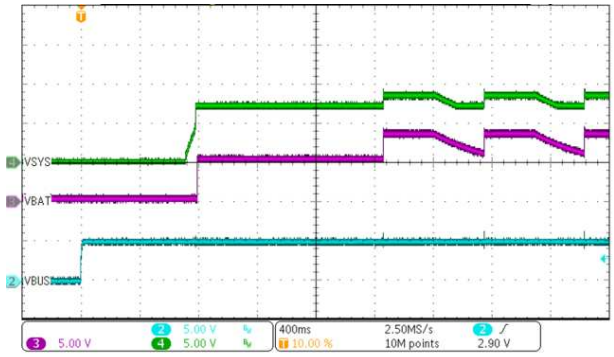
VBUS = 5V VBAT = 7.4V ICHG = 1A

Figure 78. Charge Enable



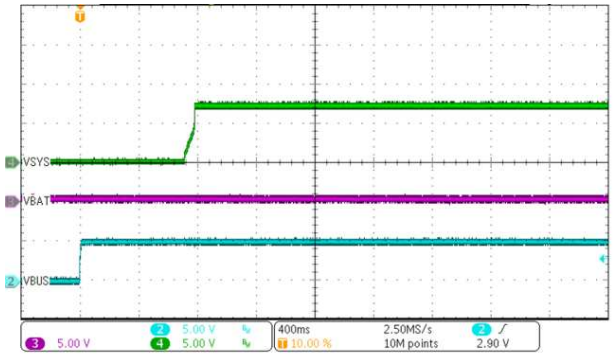
VBUS = 5V VBAT = 7.4V ICHG = 1A

Figure 79. Charge Disable



VBUS = 5V VBAT = Open Charge enabled

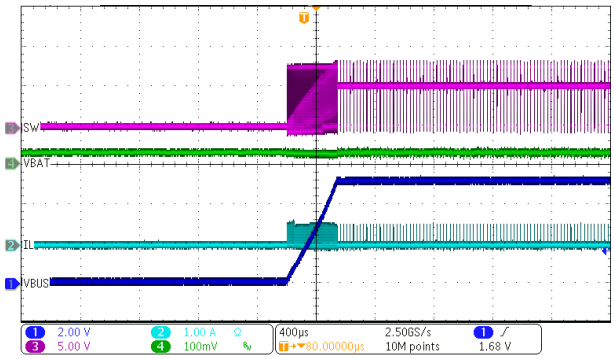
Figure 80. Adapter Plug-in with No Battery



VBUS = 5V VBAT = Open Charge disabled

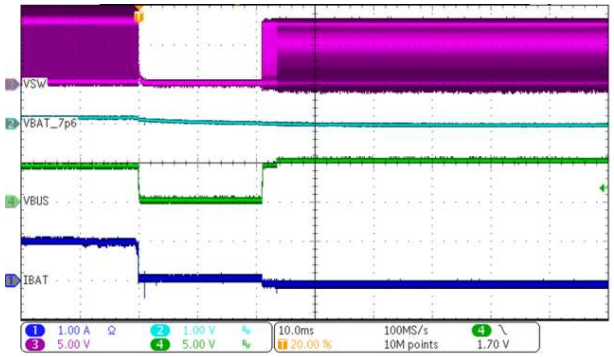
Figure 81. Adapter Plug-in with No Dead Battery

C_{VBUS} = 1µF, C_{PMID}= 10µF, C_{BAT} = 10µF, C_{SYS} = 44µF, L = DFE252012F-1R0 (1µH) (unless otherwise specified)



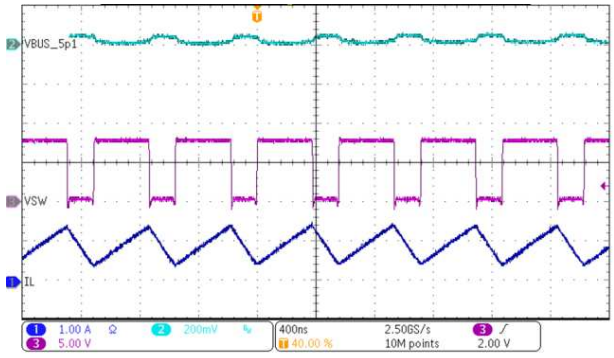
VBAT = 7.6V VBUS = 5.1V No IBUS load

Figure 82. Buck Mode (OTG) Startup



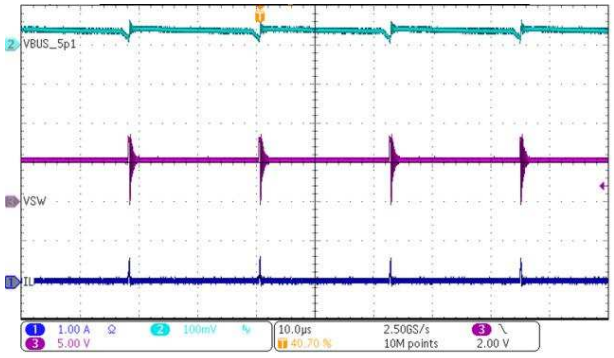
VBAT = 7.6V Adapter removed with EN_OTG = 1 R_{BUS} = 25Ω

Figure 83. Buck Mode Startup After Adapter Removal



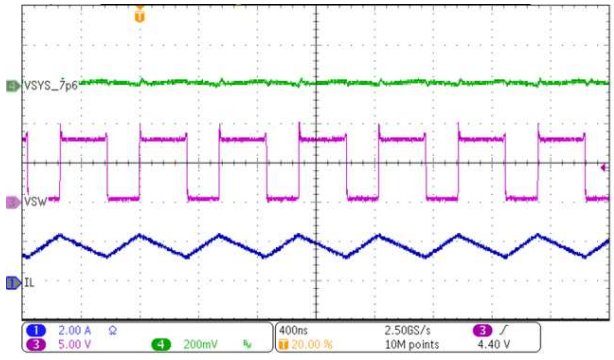
VBAT = 7.6V VBUS = 5.1V IBUS = 1A

Figure 84. Buck Mode (OTG) PWM Switching



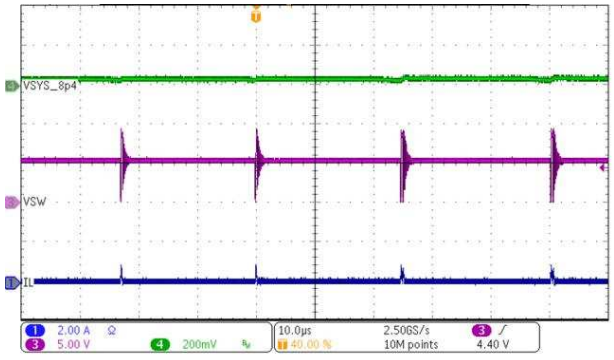
VBAT = 7.6V VBUS = 5.1V IBUS = 0mA

Figure 85. Buck Mode (OTG) PFM Switching



VBUS = 5V VBAT = 7.6V ICHG = 1A

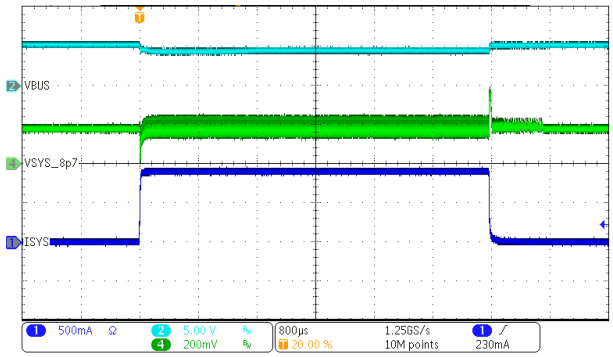
Figure 86. Boost Mode PWM Switching



VBUS = 5V VBAT = 8.4V Charge disabled

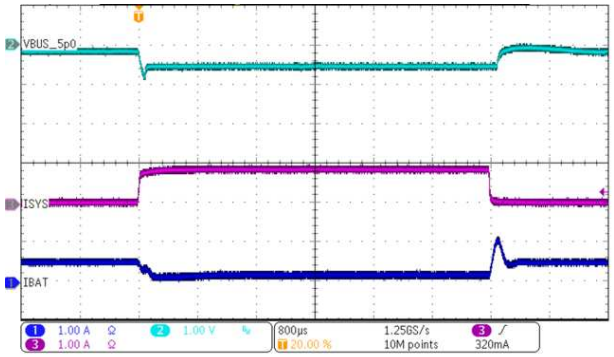
Figure 87. Boost Mode PFM Switching

C_{VBUS} = 1µF, C_{PMID}= 10µF, C_{BAT} = 10µF, C_{SYS} = 44µF, L = DFE252012F-1R0 (1µH) (unless otherwise specified)



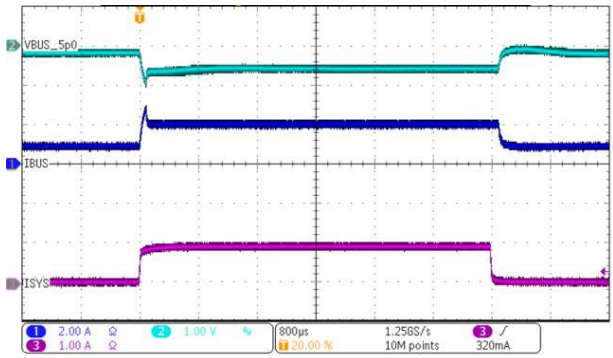
VBUS = 5V VBAT = 8.7V Charge disabled

Figure 88. SYSMIN Transient Response



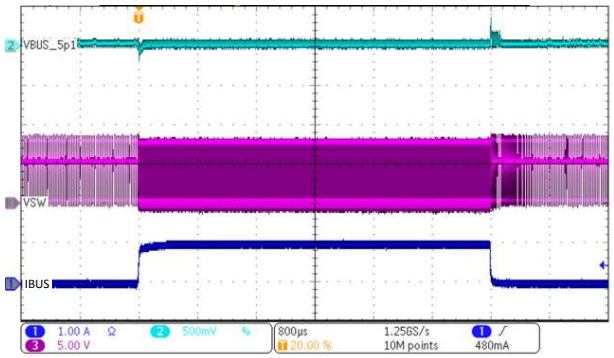
DCP Adapter VBAT = 8.0V Charge enabled

Figure 89. VINDPM Transient Response



DCP Adapter VBAT = 8.0V Charge enabled

Figure 90. IINDPM Transient Response



VBAT = 7.6V VBUS = 5.1V

Figure 91. Buck Mode (OTG) Load Transient Response

10 Power Supply Recommendations

In order to provide an output voltage on SYS, the device requires a power supply between 3.9 V and 6.2 V input with at least 500-mA current rating connected to VBUS or a dual-cell Li-Ion battery with voltage > VBAT_UVLO connected to BAT. The source current rating needs to be at least 3.3 A in order for the boost converter of the charger to provide maximum output power to SYS.

11 Layout

11.1 Layout Guidelines

The switching node rise and fall times should be minimized for minimum switching loss. Proper layout of the components to minimize high frequency current path loops is important to prevent electrical and magnetic field radiation and high frequency resonant problems. Here is a PCB layout priority list for proper layout. Layout PCB according to this specific order is essential.

1. Place SYS and BAT output capacitor as close to SYS, BAT and GND bumps as possible. Ground connections need to be tied to the IC ground with a short copper trace connection or GND plane.
2. Place PMID input capacitor as close as possible to PMID bumps and GND bumps and use shortest copper trace connection or GND plane.
3. Place inductor input terminal to SW bumps as close as possible. Minimize the copper area of this trace to lower electrical and magnetic field radiation but make the trace wide enough to carry the input current. Minimize parasitic capacitance from this area to any other trace or plane.
4. Decoupling capacitors should be placed next to the IC and make trace connection as short as possible.
5. Ensure that there are sufficient thermal vias directly under bumps of the power FETs, connecting to copper on other layers.
6. Via size and number should be enough for a given current path.
7. Route BATP and BATN away from switching nodes such as SW.

   Refer to the EVM design and the Layout Example below for the recommended component placement with trace and via locations.

11.2 Layout Example

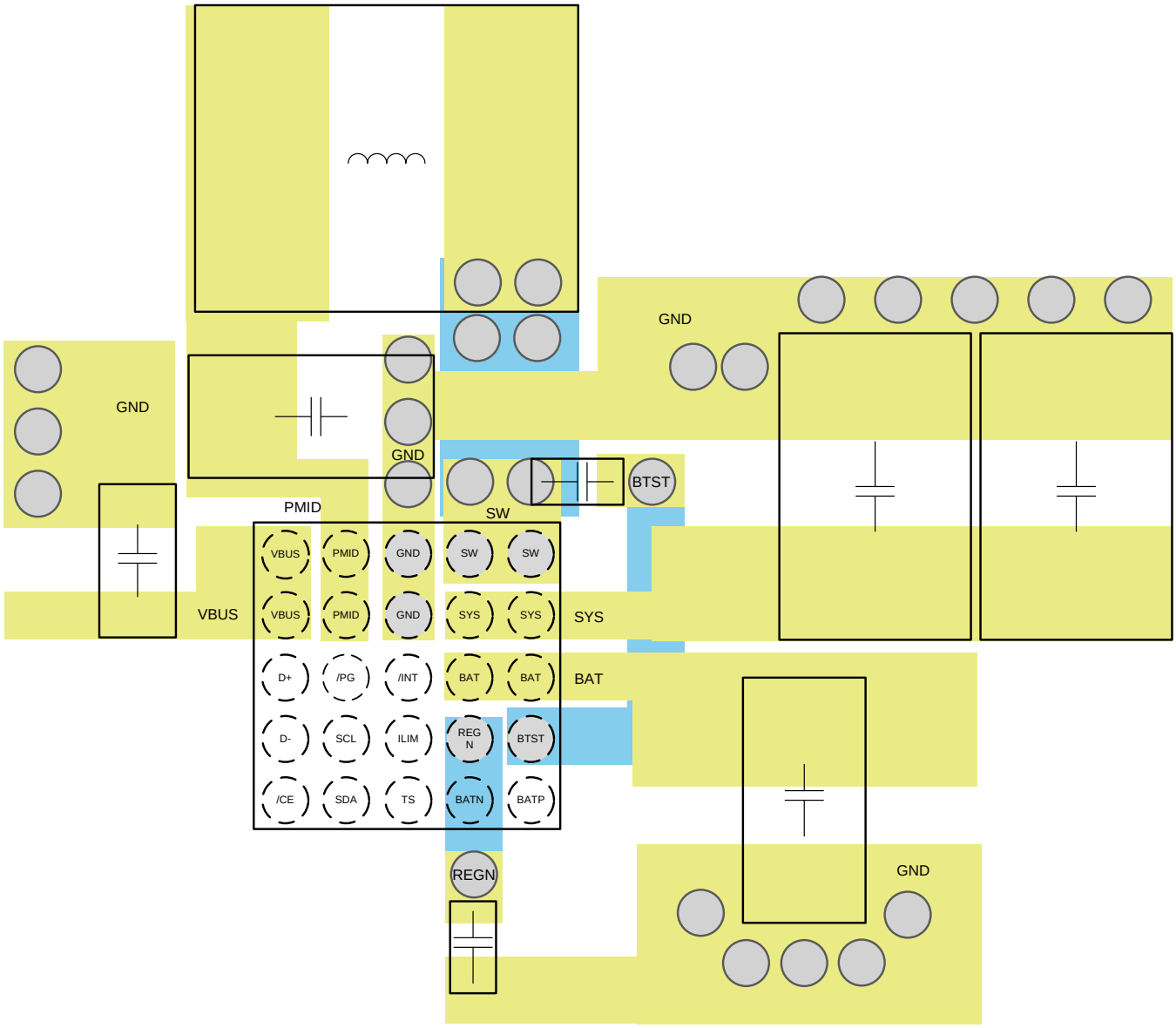


Figure 92. PCB Layout Example

12 Device and Documentation Support

12.1 Device Support

12.1.1 Third-Party Products Disclaimer

12.1.1.1 Third-Party Products Disclaimer

TI'S PUBLICATION OF INFORMATION REGARDING THIRD-PARTY PRODUCTS OR SERVICES DOES NOT CONSTITUTE AN ENDORSEMENT REGARDING THE SUITABILITY OF SUCH PRODUCTS OR SERVICES OR A WARRANTY, REPRESENTATION OR ENDORSEMENT OF SUCH PRODUCTS OR SERVICES, EITHER ALONE OR IN COMBINATION WITH ANY TI PRODUCT OR SERVICE.

12.2 Documentation Support

12.2.1 Related Documentation

For related documentation see the following:

- *BQ2588x Boosting Battery Chargers Evaluation Module User's Guide*

12.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

12.4 Community Resources

TI E2E™ support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

12.5 Trademarks

E2E is a trademark of Texas Instruments.
All other trademarks are the property of their respective owners.

12.6 Electrostatic Discharge Caution

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

12.7 Glossary

SLYZ022 — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
BQ25882YFFR	ACTIVE	DSBGA	YFF	25	3000	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	-40 to 85	BQ25882	Samples
BQ25882YFFT	ACTIVE	DSBGA	YFF	25	250	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	-40 to 85	BQ25882	Samples

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".
RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.
Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

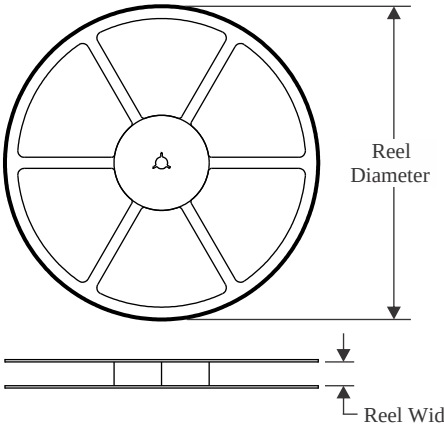
(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

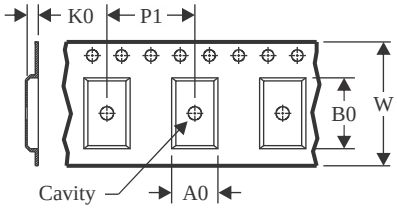
(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

TAPE AND REEL INFORMATION

REEL DIMENSIONS

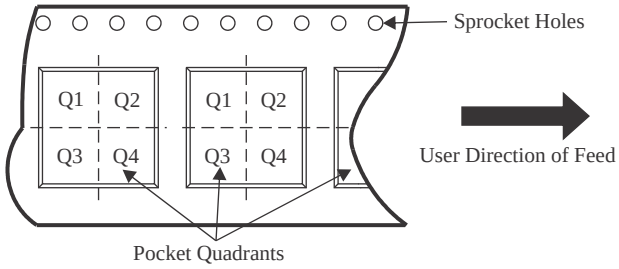


TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

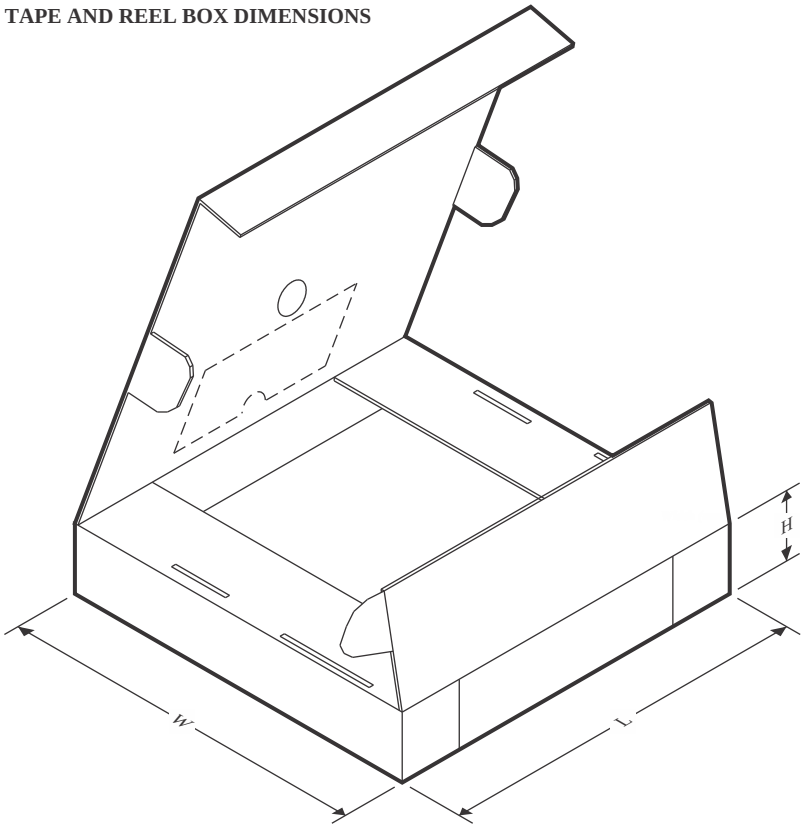
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
BQ25882YFFR	DSBGA	YFF	25	3000	180.0	8.4	2.28	2.28	0.73	4.0	8.0	Q1
BQ25882YFFT	DSBGA	YFF	25	250	180.0	8.4	2.28	2.28	0.73	4.0	8.0	Q1

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

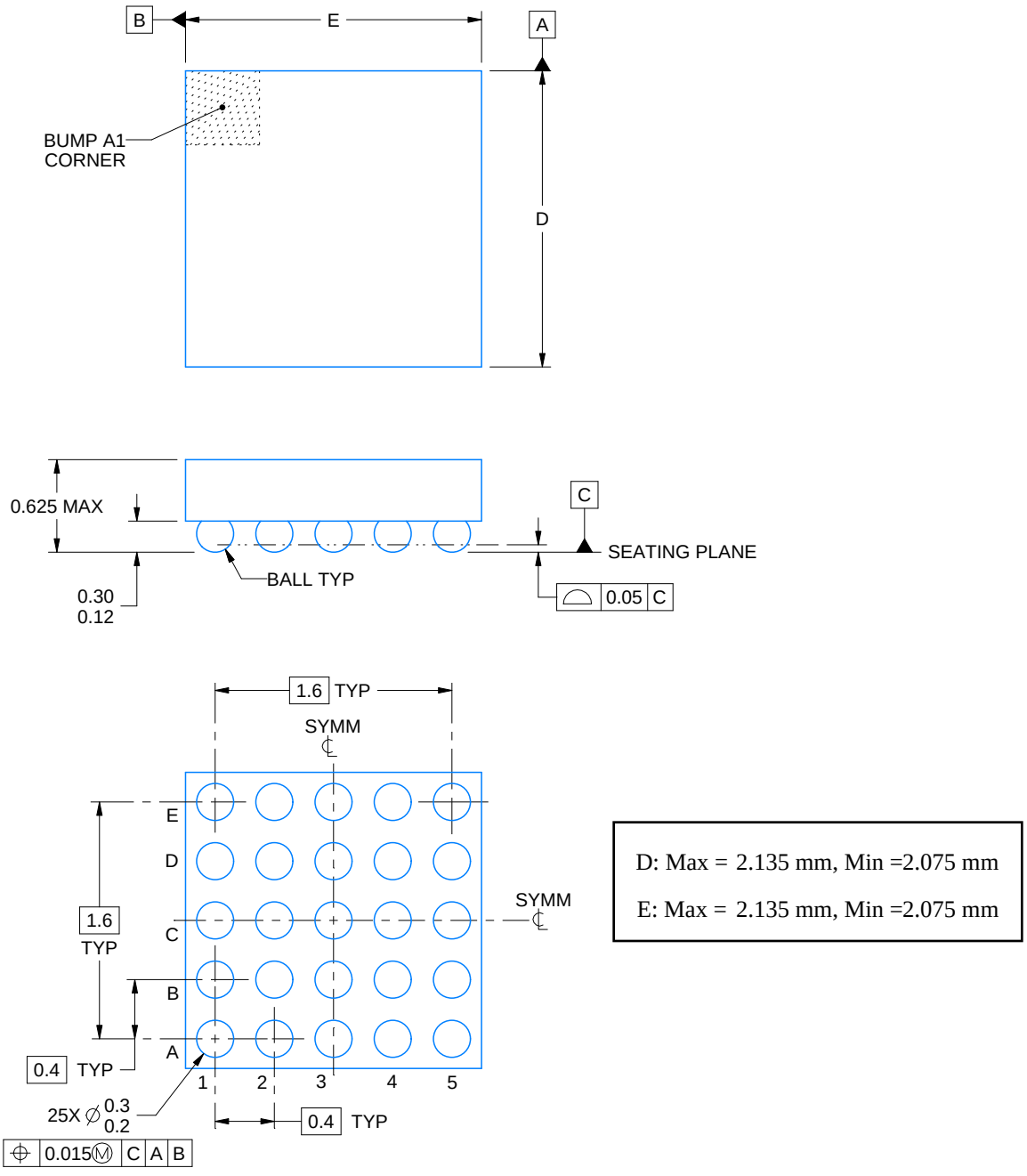
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
BQ25882YFFR	DSBGA	YFF	25	3000	182.0	182.0	20.0
BQ25882YFFT	DSBGA	YFF	25	250	182.0	182.0	20.0

PACKAGE OUTLINE

YFF0025

DSBGA - 0.625 mm max height

DIE SIZE BALL GRID ARRAY



D: Max = 2.135 mm, Min =2.075 mm

E: Max = 2.135 mm, Min =2.075 mm

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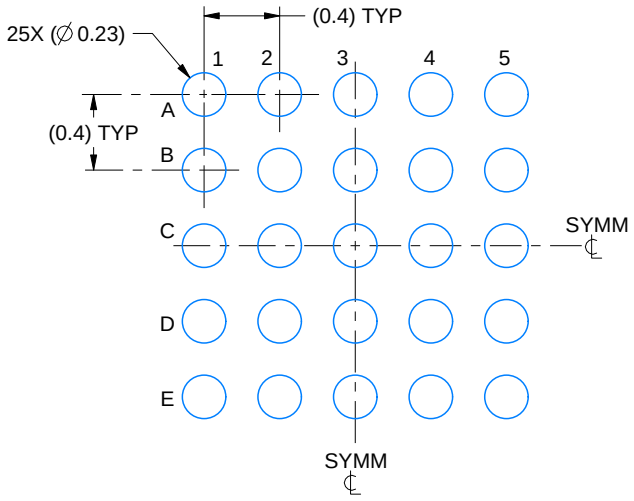
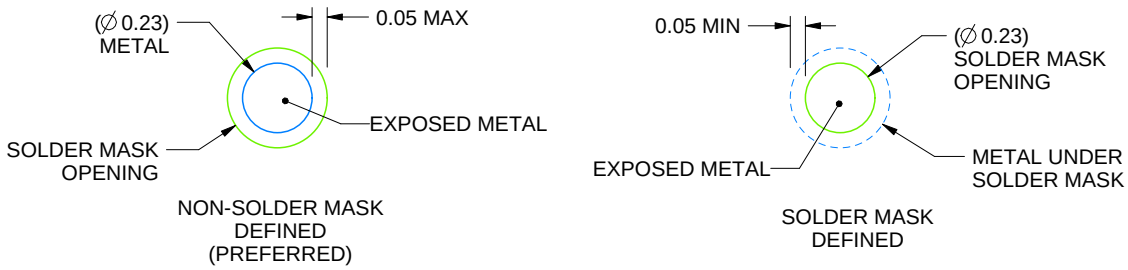
NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:25X



SOLDER MASK DETAILS
NOT TO SCALE

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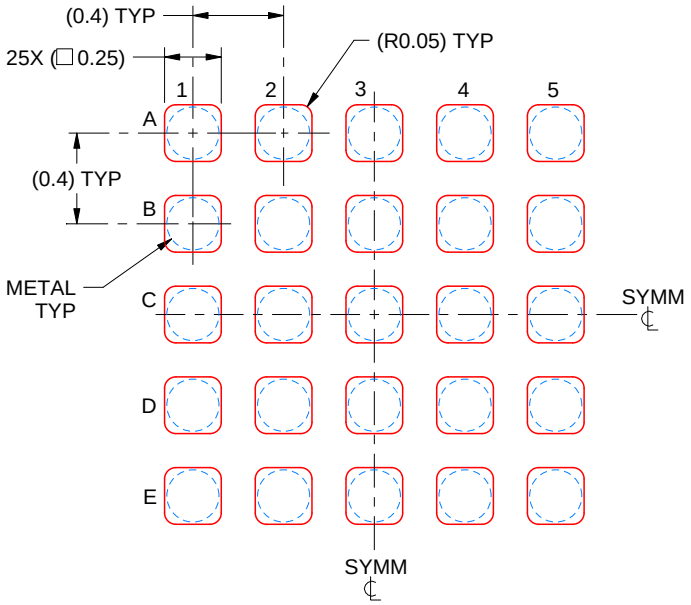
NOTES: (continued)

3. Final dimensions may vary due to manufacturing tolerance considerations and also routing constraints.
For more information, see Texas Instruments literature number SNVA009 (www.ti.com/lit/snva009).



SOLDER PASTE EXAMPLE
BASED ON 0.1 mm THICK STENCIL
SCALE:30X

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NOTES: (continued)

4. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release.

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